



**KTH Signals  
Sensors and Systems**

# Novel RF MEMS Switch and Packaging Concepts

**Joachim Oberhammer**

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*The front cover shows SEM-pictures of different top and bottom parts of MEMS switches fabricated by the author. The switches are designed to mechanically switch electrical signals in the frequency range from DC up to a few tens of GHz. The total dimensions of the parts shown in the pictures are about 1 mm. The size of the smallest features is 2  $\mu\text{m}$ , and the layer thickness is from 40 nm up to 18.2  $\mu\text{m}$ .*

*The pictures shown on pages xii, 4 and 78 are of minor scientific value and show motives from our world miniaturized in silicon by a circa 75  $\mu\text{m}$  deep plasma etching process. The dimensions of the structures are indicated by the scaling bar in the lower parts of the pictures.*

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## Abstract

Radio-frequency microelectromechanical systems (RF MEMS) are highly miniaturized devices intended to switch, modulate, filter or tune electrical signals from DC to microwave frequencies. The micromachining techniques used to fabricate these components are based on the standard clean-room manufacturing processes for high-volume integrated semiconductor circuits. RF MEMS switches are characterized by their high isolation, low insertion loss, large bandwidth and by their unparalleled signal linearity. They are relatively simple to control, are very small and have almost zero power consumption. Despite these benefits, RF MEMS switches are not yet seen in commercial products because of reliability issues, limits in signal power handling and questions in packaging and integration. Also, the actuation voltages are typically too high for electronics applications and require additional drive circuitry.

This thesis presents a novel MEMS switch concept based on an S-shaped film actuator, which consists of a thin and flexible membrane rolling between a top and a bottom electrode. The special design makes it possible to have high RF isolation due to the large contact distance in the off-state, while maintaining low operation voltages due to the zipper-like movement of the electrostatic dual-actuator. The switch comprises two separately fabricated parts which allows simple integration even with RF circuits incompatible with certain MEMS fabrication processes. The two parts are assembled by chip or wafer bonding which results in an encapsulated, ready-to-dice package. The thesis discusses the concept of the switch and reports on the successful fabrication and evaluation of prototype devices.

Furthermore, this thesis presents research results in wafer-level packaging of (RF) MEMS devices by full-wafer bonding with an adhesive intermediate layer, which is structured before bonding to create defined cavities for housing MEMS devices. This technique has the advantage of simple, robust and low temperature fabrication, and is highly tolerant to surface non-uniformities and particles in the bonding interface. It allows cavities with a height of up to many tens of micrometers to be created directly in the bonding interface. In contrast to conventional wafer-level packaging methods with individual chip-capping, the encapsulation is done using a single wafer-bonding step. The thesis investigates the process parameters for patterned adhesive wafer bonding with benzocyclobutene, describes the fabrication of glass lid packages based on this technique, and introduces a method to create through-wafer electrical interconnections in glass substrates by a two-step etch technique, involving powder-blasting and chemical etching. Also, it discusses a technique of improving the hermetic properties of adhesive bonded structures by additional passivation layers. Finally, it presents a method to substantially improve the bond strength of patterned adhesive bonding by using the solid/liquid phase combination of a patterned polymer layer with a contact-printed thin adhesive film.

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Royal Institute of Technology, SE-100 44 Stockholm, Sweden*

*You are like a manuscript, not a completed book. A book has a beginning, a middle, and an end. It is printed by machines in a press and bound as a permanent unit. But you do not know from where you have come or where you will go. You are in search of the beginning and ending pages of life in order to make your manuscript complete.*

*Sri Swami Rama  
1925–1996, Indian Sage  
in "Path of Fire and Light: Vol. 2"*

*To D. F.*



*Your manuscript is both good and original, but the part that is good is not original and the part that is original is not good.*

*Samuel Johnson  
1709–1784, British Author*

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*Freedom is not worth having if it does not include the freedom to make mistakes.*

*Mahatma Gandhi  
1869–1948, Indian Politician*

## List of papers

The presented thesis is based on the following *international reviewed journal papers*:

1. *Selective Wafer-Level Adhesive Bonding with Benzocyclobutene for Fabrication of Cavities*  
**Joachim Oberhammer**, Frank Niklaus, and Göran Stemme  
*Sensors and Actuators A: Physical*, vol. 105, no. 3, pp. 297–304, July 2003.
2. *Sealing of Adhesive Bonded Devices on Wafer Level*  
**Joachim Oberhammer**, Frank Niklaus, and Göran Stemme  
*Sensors and Actuators A: Physical*, vol. 110, no. 1–3, pp. 407–412, Feb. 2004.
3. *Low-Voltage High-Isolation DC-to-RF MEMS Switch Based on an S-shaped Film Actuator*  
**Joachim Oberhammer** and Göran Stemme  
*IEEE Transactions on Electron Devices*, vol. 51, no. 1, pp. 149–155, Jan. 2004.
4. *Design and Fabrication Aspects of an S-shaped Film Actuator Based DC to RF MEMS Switch*  
**Joachim Oberhammer** and Göran Stemme  
*IEEE/ASME Journal of Microelectromechanical Systems*, vol. 13, no. 3, pp. 421–428, June 2004.
5. *Low-Cost Glass-Lid Packaging by Adhesive Full-Wafer Bonding with Two-Step Etched Electrical Feedthroughs*  
**Joachim Oberhammer** and Göran Stemme  
Submitted for journal publication.
6. *BCB Contact Printing for Patterned Adhesive Full-Wafer Bonded 0-Level Packages*  
**Joachim Oberhammer** and Göran Stemme  
*IEEE/ASME Journal of Microelectromechanical Systems*, accepted for publication, tentative print issue December 2004.
7. *A 2-bit Reconfigurable Meander Slot Antenna with High-Displacement Low-Voltage RF MEMS Switches*  
Marc Mowler, **Joachim Oberhammer**, Björn Lindmark, and Göran Stemme  
Manuscript for journal publication.

The *contribution of Joachim Oberhammer* to the different publications:

- 1 major part of design, all fabrication, all experiments, major part of writing
- 2 part of design, major part of fabrication, all experiments, all writing
- 3 major part of design, all fabrication, all experiments, major part of writing
- 4 major part of design, all fabrication, all experiments, major part of writing
- 5 major part of design, all fabrication, major part of experiments and writing
- 6 all design, all fabrication, major part of experiments, major part of writing
- 7 part of design, part of fabrication, part of writing

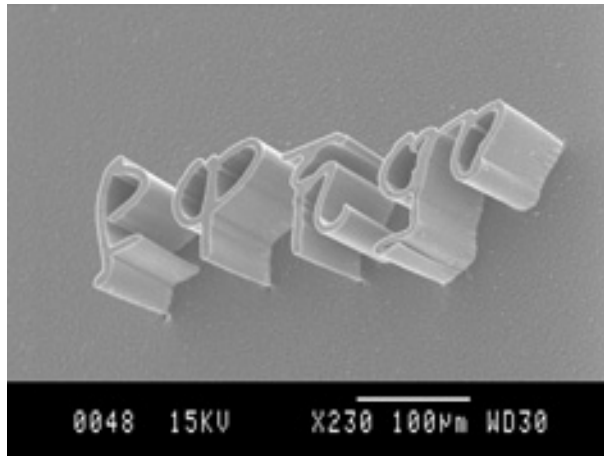
Also, the work has been presented at the following *international reviewed conferences*:

8. *Selective Wafer-Level Adhesive Bonding with Benzocyclobutene*  
**Joachim Oberhammer**, Frank Niklaus, and Göran Stemme  
*Proceedings Micromechanics Europe MME 2001*, Cork, Ireland, Sept. 16–18, 2001, pp. 54–57, *poster presentation*.
9. *Sealing of Adhesive Bonded Devices on Wafer Level*  
**Joachim Oberhammer**, Frank Niklaus, and Göran Stemme  
*Proceedings Eurosensors XVI*, Prague, Czech Republic, Sept. 15–18, 2002, pp. 298–299, *poster presentation*.
10. *Incrementally Etched Electrical Feedthroughs for Wafer-Level Transfer of Glass Lid Packages*  
**Joachim Oberhammer** and Göran Stemme  
*Proceedings of IEEE Transducers 2003*, Boston, MA, USA, June 8–12, 2003, pp. 1832–1835, *oral presentation*.
11. *RF Characterization of Low-Voltage High-Isolation MEMS Series Switch Based on a S-shaped Film Actuator*  
**Joachim Oberhammer** and Göran Stemme  
*SBMO/IEEE MTT-S International Microwave and Optoelectronics Conference IMOC 2003*, Foz de Iguazú, Brazil, Sept. 20–23, 2003, pp. 537–540, *oral presentation*.
12. *Contact Printing for Improved Bond-Strength of Patterned Adhesive Full-Wafer Bonded 0-Level Packaging*  
**Joachim Oberhammer** and Göran Stemme  
*IEEE Micro Electro Mechanical Systems 2004*, Maastricht, The Netherlands, Jan. 25–29, 2004, pp. 713–716, *poster presentation*.
13. *S-shaped Film Actuator for Low-Voltage High-Isolation MEMS Metal Contact Switches*  
**Joachim Oberhammer** and Göran Stemme  
*IEEE Micro Electro Mechanical Systems 2004*, Maastricht, The Netherlands, Jan. 25–29, 2004, pp. 637–640, *oral presentation*.

It has also been presented at the following *international workshops*:

14. *Adhesive Wafer Bonding for Packaging Applications*  
Frank Niklaus, **Joachim Oberhammer**, and Göran Stemme  
*Workshop on MEMS Sensor Packaging 2003*, Copenhagen, Denmark, Mar. 20–21, 2003, *oral presentation by Frank Niklaus*.
15. *A 2-bit Reconfigurable Meander Slot Antenna with High-Displacement Low-Voltage RF MEMS Switches*  
Marc Mowler, **Joachim Oberhammer**, Björn Lindmark, and Göran Stemme  
*MEMSWave — Workshop on MEMS for Millimeterwave Communication 2004*, Uppsala, Sweden, June 30–July 2, 2004, pp. D20–D24, *oral presentation by Marc Mowler*.

The RF MEMS switch the journal papers 3 and 4 are based on was also brought closer to the public by the full-page article *Nytt koncept för RF-switchar* (*New concept for RF-switches*) published in the weekly Scandinavian electronics magazine *Elektronik i Norden*, no. 19, November 28, 2003, p. 46, written by Jonas Karlsson.



The first name of the author in Tigrinia (JU WA CHI M), one of the three official languages of Eritrea. The letters are etched about  $75\text{ }\mu\text{m}$  deep into a silicon wafer by an inductive coupled plasma tool. The line width is between  $3$  and  $10\text{ }\mu\text{m}$ , and the structures have a height to width aspect ratio of up to 25:1.

*I would have written a shorter letter but  
I didn't have the time.*

*Blaise Pascal  
1623–1662, French Scientist*

# 1 Introduction

## 1.1 General introduction

Besides the need for computational circuits in our information age, the success of microelectronics is based on its ability for miniaturization, integration and high volume production, leading to very low chip prices which allows them to be used in the low end mass produced devices surrounding us in everyday life.

Microelectromechanical systems (MEMS), are integrated microdevices combining electrical components with active and passive interface functions to the physical surroundings, interacting with motion, sound, light, radio waves, gases, liquids, the chemical environment, thermic radiation, etc. Thus, if the traditional microelectronic chip is the "brain" or the information processing unit, the MEMS part adds the "senses" to the microsystem. The variety of new functionalities opens a new world for microdevices, whose impact on society is predicted to become similar to the impact of microelectronics up to today. The high potential of the enabling technology MEMS lies — similar to microelectronics, since MEMS fabrication is based on traditional, highly developed and optimized semiconductor manufacturing processes, infrastructure and logistics — in the possibilities of miniaturization with advantageous scaling properties for improved device or system performance, a high level of circuit integration to improve performance or to reduce barriers to incorporation into larger systems, and batch fabrication for very large volume production. The latter results in unbeatable product uniformity and low prices per unit. The complexity of the mechanical parts and interfaces of a MEMS device, reliability issues and the non-standardized packaging procedure, as well as the fact that the production volume and the degree of miniaturization are decisive for the price of the final device and thus for its market potential, have so far only led to a few successful commercialized high-volume MEMS products. Examples are accelerometers, pressure sensors and recently also gyros in automotive applications<sup>1</sup>, inkjet print heads and actuators for hard disk read/write heads, as illustrated in Table 1. However, there are niche-markets where the required device performance dominates over the cost of a low-volume product. Typical examples for such MEMS devices are pressure sensors for minimal invasive diagnostics and surgery, besides high-end military and space applications. In recent years, semiconductor equipment manufacturers have been putting increasing efforts into machines dedicated for MEMS production. Due to this tendency, in line with fabrication and material costs pushed down by the advancing IC industry<sup>2</sup>, we hopefully soon will

---

<sup>1</sup>Already in 2000 a BMW 740i had over 70 MEMS devices aboard [1].

<sup>2</sup>With the IC industries moving to 8 and 12 inch production lines, 4 and 6 inch factories and equipment are available cost-effectively for MEMS. To give another example: the costs of SOI wafers, a favorable substrate for MEMS sensor applications, are expected to sink drastically to about 3 times of the costs of conventional silicon wafers, since SOI wafers are predicted to account for half of CMOS sub-180 nm production on 300 mm wafers in about 5 years [2].

**Table 1.** Commercialization of selected MEMS devices (based on [3], 2000).

Product	Discovery	Evolution	Cost reduction, application expansion	Full commer- cialization
pressure sensors <sup>a</sup>	1954–1960	1960–1975	1975–1990	1990–present
accelerometers <sup>b</sup>	1974–1985	1985–1990	1990–1998	1998
gas sensors	1986–1994	1994–1998	1998–2005	2005
valves	1980–1988	1988–1996	1996–2002	2002
nozzles	1972–1984	1984–1990	1990–1998	1998
photonics/displays	1980–1986	1986–1998	1998–2004	2004
bio/chem. sensors	1980–1994	1994–1999	1999–2004	2004
rotation sensors <sup>c</sup>	1982–1990	1990–1996	1996–2002	2002
RF switches	1994–1998	1998–2001	2001–2005	2005
micro relays	1977–1982	1993–1998	1998–2006	2006

<sup>a</sup>Silicon strain gauges have been commercially available since 1958 [1], and in 2003, SensoNor, Norway, now an Infineon Technologies company, fabricated about 20 million tire pressure sensors.

<sup>b</sup>The first commercial accelerometer was brought on the market by Analog Devices in 1990 [1] and the total sales of accelerometers fabricated by Analog Devices in the year 2003 was over 50 million [4].

<sup>c</sup>Robert Bosch GmbH, Germany, fabricated over 4 million gyros in 2003 [4].

see more MEMS volume products also in less exclusive market segments.

RF MEMS are micromachined devices interacting with electrical signals up to the radio frequency range. The components of main interest are micromachined switches, mechanically tunable capacitors and three-dimensional inductors, which offer superior performance compared to their electronic counterparts. Of all of the emerging subfields of MEMS with ongoing research and development activities, RF MEMS is believed to have the largest market potential since the high volume consumer market of personal communication devices already exists and literally cries out for new technologies expanding its possibilities. Technology reliability, integrability, the demanded short development times and the risk of taking a completely new type of component into a high volume product on a very competitive market have so far inhibited the leap from research labs to commercial production lines. High-end applications such as automated measurement and test equipment, automotive safety and communication systems, military and space applications are believed to be the initial application fields of RF MEMS components.

This thesis focuses on two topics of RF MEMS switches: firstly, it presents a new electrostatically actuated microswitch design which, due to its concept, addresses and offers solutions in many critical aspects such as low voltage design, large traveling distance of the switching contacts which is important for achieving high isolation, the device packaging and in the integration with RF circuits. Secondly, this thesis presents new ideas in the field of near-hermetic wafer-level packaging by full-wafer bonding with a patterned adhesive layer. Issues such as in-plane and vertical interconnections are discussed, and solutions are provided to improve the hermeticity of polymer packaging and the bond strength of patterned adhesive bonded wafers.

## 1.2 Objectives and methodology of this thesis

A thesis, following a statement by Christopher Clack, University College London, UK, is the *acquisition and dissemination of new knowledge*. I would like to add that writing a thesis is also a most welcome possibility to reflect over the knowledge acquired during the preceding years, and that it is a learning process about how to prepare this knowledge for distributing and presenting it to the interested community. It also makes it possible to give a personal accent to the research topics, thus letting the people participate on the experiences accumulated by being exposed to all the little problems a researcher encounters during his daily work.

From a MEMS device perspective, the presented thesis adds knowledge in the fields of RF MEMS switches and MEMS wafer-level packaging by showing new concepts and ideas which are proved by successful fabrication, experiments and evaluation.

From a micromachining point of view, the thesis provides new knowledge mainly in the field of adhesive bonding and in the fabrication of thin-film microactuators.

From a scientific perspective, finally, the value of this thesis in terms of novelty and relevance for the field is attested by the acceptance of the appended international journal papers and the referred international conference proceedings through an established scientific reviewing process.

## 1.3 Structure of this thesis

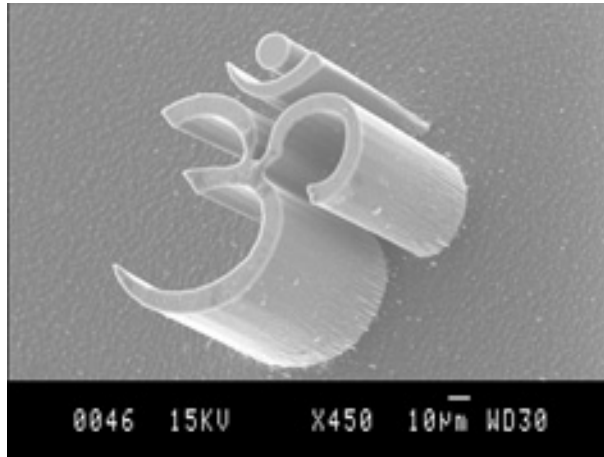
According to the procedures at the Royal Institute of Technology, this thesis consists of a brief introductory part and of a part with the appended journal papers.

The intention of the introductory part is not to recite the known knowledge in the field, but rather to give an overview of the motivation and the activities in the field, and to provide references for further study of the related topics (Section 2). However, a few selected topics are discussed in a more extensive way in Section 2.2. The references, collected at the end of the introductory part, are chosen by the preference of the author in terms of their relevance according to the intention stated above, and are not intended to give a complete overview of the research activities in the field.

Furthermore, the introductory part of this thesis presents the research topics of the scientific work this thesis is based on, discussing their concepts and relevance and comparing them to related existing work (Sections 3, 4). Finally, the introductory part closes with a detailed summary of the appended papers (Section 5), drawing conclusions (Section 6) and giving a brief outlook over the possible continuation of the work described in this thesis (Section 7).

The second part consists of the reprints of the journal papers this thesis is based on. The conference proceedings listed on pages x and xi are not included in this thesis but their content is also discussed in the introductory part.

Finally, I kindly ask for tolerance of the many footnotes spread all over the book. I know about the arguments against footnotes but I still think the way they are used in this book is a good compromise between providing interesting detailed information and not distracting the reader from the main flow of reading, which I have tried to keep on a less detailed level.



*All men dream: but not equally.  
Those who dream by night in the dusty recesses  
of their minds  
wake in the day to find that all was vanity:  
but the dreamers of the day are dangerous men,  
for they may act their dream with open eyes,  
to make it possible.*

*Thomas Edward Lawrence  
1888-1935, British Soldier, Writer,  
in "The Seven Pillars of Wisdom"*



*You can't really be strong until you can  
see a funny side to things.*

*Ken Kesey  
1935–2001, American Writer,  
in "One flow over the cuckoo's nest"*

## 2 Background

### 2.1 RF MEMS switches

#### 2.1.1 RF MEMS

The first MEMS switch was fabricated already in 1979 by K. E. Petersen [5], showing the possibilities of the emerging micromachining technology. In 1991, L. E. Larson presented a MEMS switch dedicated to switching RF signals up to a few tens of GHz [6]. Despite the immaturity of the device, its electrical signal characteristics were so remarkable that it brought many research institutes and universities on the track of RF MEMS [7–14]. In the middle of the nineties of the last century, one publication followed another on new RF MEMS device concepts, and by the end of the nineties the research included more and more device reliability issues, indicating the increasing maturity of the technology. The device of main interest, due to its excellent performance, was and still is the micromachined switch. Other RF MEMS components are mechanical [15–19], dielectric [20] and cavity [21, 22] resonators for frequency-selecting applications, tunable capacitors [23–33], three-dimensional inductors [34–39], as well as micromachined antennas [40–42] and transmission lines [43–46]. RF circuits based on MEMS components are wide-band phase-shifters realized with MEMS switches [47–50] or micromachined tunable capacitors [51, 52], reconfigurable antennas [53–56], voltage controlled oscillators (VCO, [8, 57, 58]) and impedance tuning circuits [59, 60], just to give a few examples.

In a total, all these devices are characterized by their outstanding performance — low insertion loss and high signal linearity for the switches; high frequency selectivity and very small chip size of filters based on mechanical resonators; high  $Q$ -value and high self resonance frequency (SRF) of micromachined inductors; high tuning range, low series resistance and high linearity of mechanically tunable capacitors [7, 10] — but on the other hand also by their fabrication complexity, reliability issues [61–64], the difficulty of integrating them with traditional electronic circuits, by their special packaging requirements [9, 11, 65–67], and by device specific problems, such as the high impedance of micromechanical resonators and the relatively low power handling ability of MEMS switches.

Wireless MEMS is a term often used synonymously to RF MEMS and indicates the market potential of MEMS components in consumer products with telecommunication functionality, demanding low size, low weight, low power consumption, reconfigurability and good signal properties to fulfill new telecommunication standards [8, 10, 14, 68, 69]. These requirements make RF MEMS devices very suitable for replacing bulky passive off-chip components, which currently inhibit further miniaturization of wireless equipment since they consume most of the circuit board size and, in contrast to RF MEMS devices, cannot be integrated on-chip without sacrific-

ing performance [11]. Wireless applications are a low cost and high volume market where introducing devices of a new mechanical type with uncertain reliability entails high risk. Especially in this field it will be very difficult for MEMS switches to compete with semiconductor switches, which are available in high volumes for 0.3–0.6 USD/circuit [70]. Also, the down-turn of the telecommunication sector since 2001 with an investment stop in emerging technologies was a barrier in the ascension of RF MEMS. Many companies specialized in RF MEMS development crashed because their cash reserve was not sufficient to support cash outlay operations until volume orders could be delivered and paid<sup>3</sup>.

Microwave acoustic devices [71] such as film bulk acoustic resonators (FBAR, [72]), solidly mounted resonators (SMR, [73]) and often also surface acoustic wave (SAW) devices [74] are considered RF MEMS components since they involve micromachining processes in their fabrication. In contrast to the other RF MEMS components mentioned above, these acoustic signal processing devices are already in production in high volumes. SAW filters have been used in consumer products since the 1970s and their world-wide production, according to an estimation published in March 2002, is in the region of 3 billion devices per year, mainly used in mobile radio systems [71]. Also FBAR based devices have been on the market in high volumes for a couple of years, and by July 2002, Agilent Technologies alone was shipping more than 1 million FBAR duplexers per month. The success of these microwave acoustic devices anticipates the market potential and the market's need of RF MEMS devices to even further push technological limits and fulfill customer demands.

*The beauty of MEMS switches is their near-ideal behavior and the relative ease of their circuit designs.*

*Gabriel M. Rebeiz [70]*

### 2.1.2 Why miniaturized *mechanical* switches?

By experience, the most vulnerable and the life-time determining components of an electrical device are its mechanical parts. Also, the reliability and the performance of electrical/electronic components has been increasing during the last decades at an incredible pace. Especially electronic switches have been the object of extensive research and development activities since they are the very basic components of our information age characteristic computational devices. Why then would one seriously think about using a miniaturized *mechanical* component for switching electrical signals in technological high-end applications?

The development of digital switches (transistors) in logic devices has proceeded at an incredible speed over the last decades in terms of components per chip, cost per function, clock rates, power consumption, compactness and functionality [86, 87]. However, the limits of digitally controllable analog signal switches have not been advancing that fast, and electronic switches based on PIN diodes and field

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<sup>3</sup>A recent example is PHS MEMS located in Grenoble, France, probably the until now largest European MEMS company specialized in RF MEMS, founded in 1998 and declared to be in liquidation in February 2004, with 95 dismissed employees.

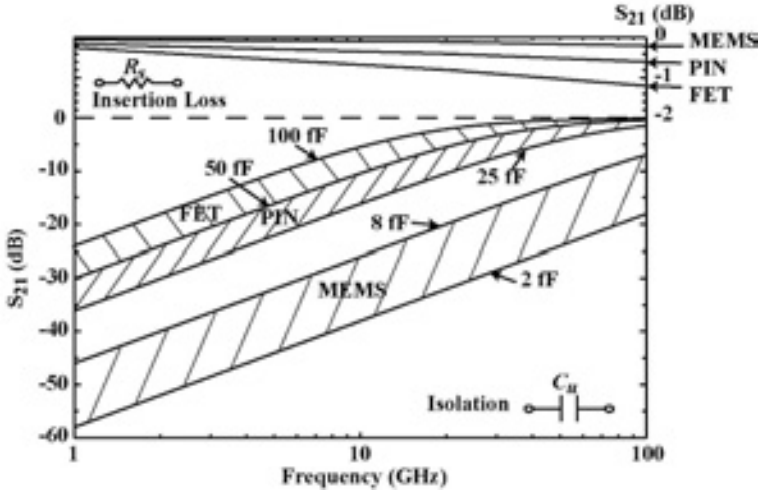
**Table 2.** Performance comparison of switches based on PIN-diode, FET, MEMS or conventional electromechanical relay (EMR) technology, designed for switching RF signals.

Parameter	Unit	PIN diode switch [7, 70, 75–80]	FET switch [7, 70, 75, 81–83]	EMR <sup>a</sup> [84, 85]	MEMS switch [7, 70, 75]
actuation voltage	V	±3–5	3–5	3–24	20–80
actuation current	mA	3–20	0	150–15	0
power consumption	mW	5–100	0.05–0.1	<400	0.05–0.1 <sup>b</sup>
switching time	μs	0.01–0.1	0.001–0.1	>1000	1–300
off-state capacitance	fF	18–80	70–140	–	1–6
series resistance	Ω	2–4	4–6	<0.1	0.5–2
cutoff frequency	THz	1–4	0.5–2	0.005	20–80
RF isolation (1–10 GHz)	dB	>35	15–25	>40	>40
RF isolation (>10 GHz)	dB	20–35	<20	–	25–40
insertion loss (1–10 GHz)	dB	0.3–0.7	0.4–2	<0.3	0.05–0.2
insertion loss (>10 GHz)	dB	0.7–2	>2	–	0.1–0.2
isolation voltage	–	medium	low	high	low
power handling (1 GHz)	W <sub>CW</sub>	<10	<5	10	<0.5
signal linearity (IP3)	dBm	27–45	27–53	>60	66–80
size	–	small	very small	large	small
cost — SPDT type [84]	USD	0.9–8	0.5–4.5	0.85–12	8–20
life cycles [84]	–	>10 <sup>9</sup>	>10 <sup>9</sup>	0.5–5×10 <sup>6</sup>	>10 <sup>8</sup>

<sup>a</sup>PCB mounted relays for high-frequency switching applications, such as the OMRON G6 series<sup>b</sup>includes drive circuitry**Table 3.** Advantages and disadvantages of MEMS switches.

Advantages	Disadvantages
very high DC and RF isolation	high actuation voltages needed <sup>a</sup>
very low insertion loss	low switching speed
high signal linearity	limited power handling
almost zero power consumption <sup>a</sup>	lifetime uncertain
very large bandwidth <sup>b</sup>	problematic integration with RF circuits
miniaturization	special packaging necessary
simple control circuits	reliability uncertain
high volume production possible	price uncertain
very resistant to external influences <sup>c</sup>	

<sup>a</sup>for electrostatically actuated switches<sup>b</sup>without sacrificing performance<sup>c</sup>temperature, radiation, mechanical shock



**Figure 1.** Typical RF performance of MEMS metal-contact switches compared to PIN-diode and FET based switches, extracted from a simplified equivalent circuit model [88].

effect transistors (FET) can hardly meet the performance requirements of today's communication systems, especially concerning the isolation, the insertion loss (on resistance) and the signal linearity. Even though also semiconductor-based switches have been improving over the last decade [76–78, 81–83], their RF signal performance still decreases drastically with frequency above 1 GHz where they are limited either in power handling or they show a very large insertion loss, poor isolation and high signal distortion [7]. However, MEMS switches perform very well over an extremely large bandwidth with very uniform characteristics, even above 100 GHz [75]. That is the advantage of having a purely mechanical element opening and closing or short-circuiting an almost unimpaired transmission line. Table 2 compares the performance of PIN-diode and FET based electronic switches and high-frequency electromechanical relays (EMR) with RF MEMS switches. The superior RF signal handling properties of MEMS metal contact switches are also demonstrated in Figure 1.

Unfortunately, MEMS switches also have some obvious disadvantages, which are compared to their advantages in Table 3. This problematic has inhibited the leap from the lab to the fab, since so far no system and component manufacturer has dared to take the risk of embedding a device uncertain in its reliability and lifetime into a volume product.

It is very interesting to observe that miniaturized mechanical relays, as small as  $0.3 \text{ cm}^3$  in last generation telecom applications [89], still out-perform their electronic counterparts, solid-state relays (SSR), especially in isolation, on-resistance and even in price. This drives a significant commercial need for a device which bridges the technical and economical gaps between electromechanical "macro-relays" and solid-state devices, in particular to address the need for extreme miniaturization while preserving the benefits of metal contact switching. MEMS fabrication technology offers the benefits of economies of scale, precision, and device matching capabilities that are un-

**Table 4.** Categorizing of MEMS devices in terms of their mechanical complexity<sup>a</sup>.

	Category 1	Category 2	Category 3	Category 4
moving parts	—	×	×	×
touching surfaces	—	—	×	×
impacting surfaces <sup>b</sup>	—	—	—	×
examples	inkjet heads, microneedles	accelerometers, gyros	valves	metal contact switches
complexity	low	—————→		high
reliability	less critical	—————→		highly critical

<sup>a</sup>Based on information from an oral presentation by G. M. Rebeiz at the Transducers 2003 conference, Boston, USA, June 8–12, 2003.

<sup>b</sup>The physical impact — force or rubbing — between the two surfaces determines the functionality of the device.

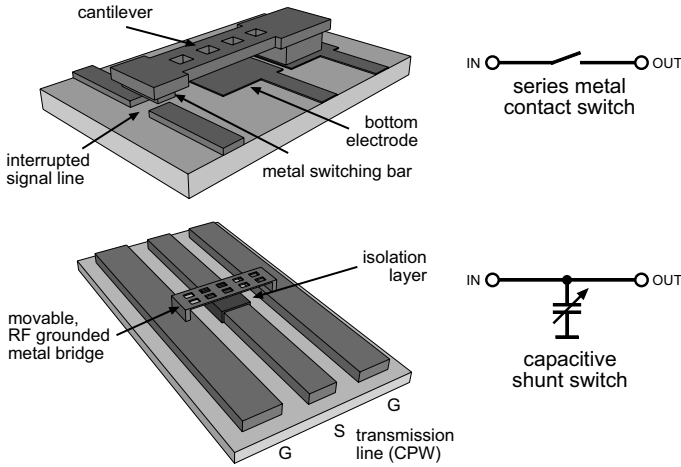
paralleled in conventional assembly-based manufacturing. Therefore, micromachining seems to be a very attractive technology for fabricating extremely miniaturized, high-volume, low-cost micromechanical relays [90].

Even from a MEMS perspective, a micromachined switch — especially a DC metal contact switch — is mechanically a rather problematic and complicated device compared to other typical MEMS devices, as illustrated in Table 4. However, because of their market potential, MEMS switches are attracting a lot of attention from universities, research institutes and industries. The need of better signal switching properties in more sophisticated telecommunication standards and in high-end applications might compensate for the technological skepticism [7, 10, 14, 91].

At the end of this introducing section I would like to acknowledge the so far most exhaustive source on RF MEMS switches and the probably best book ever written on this sub-topic of MEMS: Gabriel M. Rebeiz's "RF MEMS: Theory, Design, and Technology" [75], discussing the status of the field by September 2002. This book is so rich in detail that it seems to address all conceivable aspects of the matter. It completely meets the requirement stated as the first sentence in its preface: "All I wanted to do is to write a *deep book*" (*italics as in the original*). Many thanks to the author and his co-writers for sharing their knowledge and preparing it in such an extraordinary way!

### 2.1.3 MEMS switch types

A MEMS switch is a bistable mechanical device fabricated by micromachining techniques, allowing the free propagation of an electrical signal from an input to an output in one state and blocking the signal in the other state. Stability in one or both of the states is achieved with or without applying an external energy source, and the transition between the two states is controlled by imposing or releasing the external



**Figure 2.** The two most common electrostatically actuated MEMS switch types.

effort.

RF MEMS switches can, among others, be classified by their actuation principle, by the circuit configuration, the fabrication technology or by the intended application. Table 5 illustrates some possibilities to categorize MEMS switches.

About 80% of the switches presented so far belong to the two most common switch types, schematically illustrated in Figure 2:

- Electrostatically actuated series switches with metal contacts:** A cantilever or membrane with a switching metal contact bar vertically opens or closes the signal line. Switches of this category are normally off, and the incoming RF wave is reflected by the interrupted signal line in the off-state. In the on-state, the signal can propagate over the metal contact bar short-circuiting the gap between the input and the output line. This switch type is capable of switching DC to RF signals and is both in its fabrication and in its reliability more complicated than the next type, mainly because of its metal contacts.
- Electrostatically actuated capacitive shunt switches:** Consist of a metal bridge or membrane connected to RF ground and moving vertically above the isolated signal line. The switch is normally on, and in the down position, the bridge capacitively short-circuits the signal line to the RF ground. Thus, in the down-state (off-state), the signal propagates via the bridge to the ground and the isolation mainly depends on the parasitic inductance of the whole ground path<sup>6</sup>. This switch type is basically not suitable for low frequency signals. It is relatively simple in its fabrication, very fast with a switching time of only a few microseconds, and very small compared to the metal contact switches which need a stronger actuator to obtain the necessary contact and opening forces.

<sup>6</sup>The parasitic inductance is often used as a design feature, since it creates, together with the bridge capacitance, a series resonant circuit providing increased isolation around the resonance frequency [70, 93].

**Table 5.** Possible categories of MEMS switches.

Classified by	Categories	Comments
circuit type	○ series switch	metal contact or capacitive
	○ shunt switch	metal contact or capacitive
switching contact	○ metal contact	DC to a few tens of GHz
	○ capacitive	RF >2 GHz <sup>a</sup>
I/O configuration	○ SPST	single pole, single through
	○ SPDT, etc.	single pole, double through
actuation mechanism <sup>b</sup>	○ electrostatic	fast, low power, simple, high voltage
	○ electrothermal	slow, high power <sup>c</sup> , low voltage
	○ magnetostatic	high force, high power <sup>c</sup> , complex fabrication, large distances
	○ mercury-droplet	very limited applications
	○ piezoelectric	complicated fabrication
	○ combined mechanisms	complicated fabrication, good performance
movement	○ lateral	large size, difficult contact fabrication
	○ vertical	most common and most suitable for surface-micromachining
	○ torsional	low voltage, push-pull concept, simple, large contact distance
	○ rotary, in-plane	complicated, uncommon
moving structure	○ membrane	large actuation electrodes, low voltage
	○ bridge	low RF intrusion
	○ cantilever	single side clamped bridge
fabrication technology	○ on top of RF circuits	SOC integration, only CMOS/GaAs compatible processes
	○ CMOS technology	using the standard CMOS back-end process chain, very restricted, very low cost
	○ transfer bonded	SOC integration, complicated, very flexible designs
	○ unrestricted	SOP integration or single package
	○ non-standard substrates	fabrication on low-cost laminates, restricted clean-room compatibility
switching application	○ microrelay	to switch currents >10 mA
	○ DC..RF signal	for automated test equipment (ATE) e.g.
	○ RF signal	low power, low current (<1 mA) RF signals
	○ RF power	RF signal switching >10 dBm

<sup>a</sup>the so-called floating metal-layer capacitive switch has good performance down to  $\approx 300$  MHz [92]<sup>b</sup>for a comparison see also Section 2.2.1<sup>c</sup>unless using a bistable latching mechanism

From an application point of view, a third type should be added, even though it is not a "true" RF MEMS switch, and, in its basic concept, is just a variation of the first switch type:

- **Microrelays**<sup>7</sup>: Highly miniaturized metal-contact relays fabricated by micro-machining techniques. The most common actuation mechanisms are electrostatic, magnetostatic or electrothermal, and the relays are designed with larger contact and restoring forces from a few to a few tens of milliNewtons to be able to handle currents over 10 mA. The contact resistance is rather small and in the range of a few hundred milliOhms or lower. The contact distance in the off-state has to be at least 10–20  $\mu\text{m}$  to provide sufficient DC isolation up to about 400 V in air. Reliability is a big issue for this type of switch, since the relatively weak MEMS actuator has to open metal contacts carrying larger currents. A longer switching time in the millisecond region is tolerated by most of the target applications<sup>8</sup>.

#### 2.1.4 Some selected MEMS switches

A variety of switch concepts with interesting design features have been published over the last decade. The following paragraphs give an overview of some MEMS switches, commenting on their special characteristics and performance. The selection criteria for the listed switches are: novelty at the time they were introduced, relevance in functionality, outstanding performance, special design features, fabrication processes or special choice of structural materials. The items are sorted chronologically by the date of the first publication, even if a later published journal paper with more extensive information than a first published conference paper is given as the reference.

It is very interesting to observe that research institutes, companies and universities are similarly engaged in switch development and also in publishing results in scientific journals and at conferences. However, it is clearly visible that companies put their efforts in metal contact switches with, if possible, good microwave performance, whereas the field of capacitive microwave shunt switches is still dominated by universities and research institutes. The latter switch type is of great interest for military applications, which could be one reason why companies might not be so eager to disclose their activities to the community.

A survey of MEMS switches should not be limited to scientific publications, especially since this field is of so much commercial interest. Patents are a rich source of information on switch designs and reveal many clever, otherwise unpublished ideas. Also, the activities of many companies such as Intel, Northrop Grumman, NEC, Nokia or Ericsson, publishing rarely or not at all on MEMS switches in scientific papers, can

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<sup>7</sup>The term "microrelay" is often used for all types of MEMS switches whose switched signal path is electrically isolated from the actuation circuit, i.e. not a so-called "in-line" switch. In this book however, the term "microrelay" is used for highly miniaturized, micromachined electromechanical relays to switch larger DC to low frequency currents, in contrast to a mere signal switch for currents below 1 mA.

<sup>8</sup>Basically, one must distinguish between the actuator response time and the actual switching time: for most applications, a very slow actuator is not crucial, but the actual movement of the contacts should not be longer than a few milliseconds to avoid degradation of the contacts due to sparking and discharging in close approximation.



be tracked down. Patents are not as suitable as good review papers for newcomers to study, but they are a better source for a person already having some knowledge in the field, since they often provide more extensive information and consist of much more details than scientific publications. It is also interesting to note that patent applications anticipate chronologically the development of MEMS switches. Just to give one example: The switch research and development published in 1997 by Zavracky/Majumder at Northeastern University, MA, USA, in cooperation with Analog Devices and now continued at Radant MEMS, goes back to basic ideas which were filed as patents by Zavracky already in 1984, long before Larson published his historical RF MEMS switch paper in 1991. Table 6 lists a few interesting, granted US patents on electrostatically actuated series switches with metal contacts, from the mid 80s of the last century up to 2004.

**Table 6.** Interesting US patents on electrostatically actuated MEMS series switches, granted by April 28, 2004. Sorted by the date of filing of the patent, with references to scientific publications if directly corresponding to the patented designs.

Patent no.	filed on	Inventor (assignee)	Description
4543457	25.01.1984	Petersen et al. (Transensory Devices)	pressure, acceleration, force digitizing switch
4674180	01.05.1984	Zavracky, Morrison (Foxbro)	historical cantilever switch patent
5121089	01.11.1990	Larson (Hughes)	rotating RF MEMS switch [6]
5619061	31.10.1994	Goldsmith et al. (Texas Instr., now Raytheon)	many switch variations (in-line, torsional, ..)
5578976	22.06.1995	Yao (Rockwell)	Rockwell cantilever switch [94]
5673785	03.10.1995	Schlaak, Schimkat (Siemens)	curled membrane switch [95]
5638946	11.01.1996	Zavracky (Northeastern Univ.)	series switch, base for a later Analog Devices (now Radant) switch [96]
6162657	06.11.1997	Schiele, Kozlowski (Fraunhofer Ges.)	curled cantilever switch [97]
6046659	15.05.1998	Loo, Hyman, et al. (Hughes, now HRL)	classical cantilever-spring design with sandwich structure [98]
6100477	17.07.1998	Randall, Kao (Texas Instruments)	recessed membrane switch
6143997	04.06.1999	Feng and Shen (Univ. of Illinois)	guided, free-standing membrane switch
6229683	30.06.1999	Goodwin-Johansson (MCNC)	curled cantilever; electrodes surrounding the contacts; buried electrodes
6307452	16.09.1999	Sun (Motorola)	folded spring membrane switch with distance bumps
6307169	01.02.2000	Sun et al. (Motorola)	single anchored, membrane-type switch [99]

Table 6. continued.

Patent no.	filed on	Inventor (assignee)	Description
6373007	19.04.2000	Calcaterra et al. (US Air force)	3-metal-layer, vertically stacked shunt/series switch
6469603	19.06.2000	Ruan, Shen (Arizona State Univ.)	latching micro-magnetic relay [100], later at MICROLAB Inc.
6655011	02.10.2000	Kornrumpf et al. (General Electric)	mechanically bistable, vertically moving switch
6483395	16.03.2001	Kasai et al. (NEC)	membrane type with distributed stiffness; transfer-bonded switch
6646215	29.06.2001	Nelson (Teravacin Technologies)	cantilever switch with pull-mechanism to separate the contacts
6529093	06.07.2001	Ma (Intel)	stepped actuation electrodes switch
6531668	30.08.2001	Ma (Intel)	hollow-shaped in-line switch
6535091	06.11.2001	Bechtle et al. (Sarnoff)	torsional switch
6720851	01.04.2002	Hallbjörner et al. (Ericsson)	switch with mechanically tunable membrane stiffness
6628183	10.05.2002	Kang, Cho (Samsung)	center-anchored minimal invasive, ultra-miniaturized series switch
6657525	31.05.2002	Dickens et al. (Northrop Grumman)	stiff membrane; folded, thin springs; distance keepers
6624720	15.08.2002	Allison, Lee (Raytheon)	4-pole switching device based on four cantilever switches
6639494	18.12.2002	Bluzer (Northrop Grumman)	bridge switch with three symmetric arms

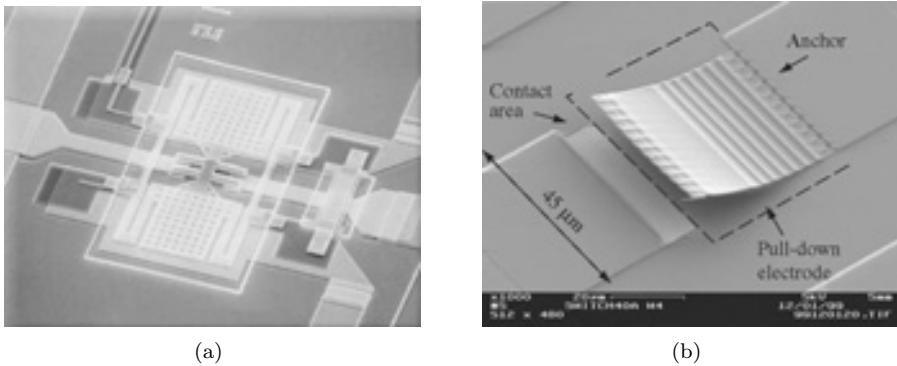
## Series switches

The ancestor of all RF MEMS switches was published in 1991 by Hughes Research Laboratories, CA, USA. The rotating microwave transmission line switch initiated world-wide RF MEMS activities since it proved to have excellent RF signal performance. Its insertion loss  $S_{21,ON}^9$  was less than 0.5 dB and its isolation  $S_{21,OFF}$  was greater than 35 dB from DC to 45 GHz. However, the required actuation voltages were above 100 V, and the switch had difficulties in establishing proper contact between the rotor and the switch contacts [6].

In 1994, the University of Neuchâtel presented a switch fabricated with traditional LPCVD surface micromachining processes. The switch could be operated at 50–75 V and was integrated with MOSFETs on the same substrate. However, the contact resistance of the doped polysilicon structure was in the order of 10 k $\Omega$  [101].

In 1995, people at Texas Instruments, USA, presented an electrostatically actuated switch based on a modified micromechanical membrane structure initially developed

<sup>9</sup>In accordance with common practice, all data of logarithmic  $S$ -parameters is given in absolute (positive) values throughout this thesis. Mathematically it would be more correct to set a negative sign, since all of the  $S$ -parameters of a passive network can only be within the area of the unit-circle, i.e.  $|S_{xx,lin}| \leq 1$ .



**Figure 3.** (a) Rockwell Science Center membrane based metal contact switch (size about  $250 \times 250 \mu\text{m}$  [104]), and (b) MIT Lincoln Labs in-line switch [105].

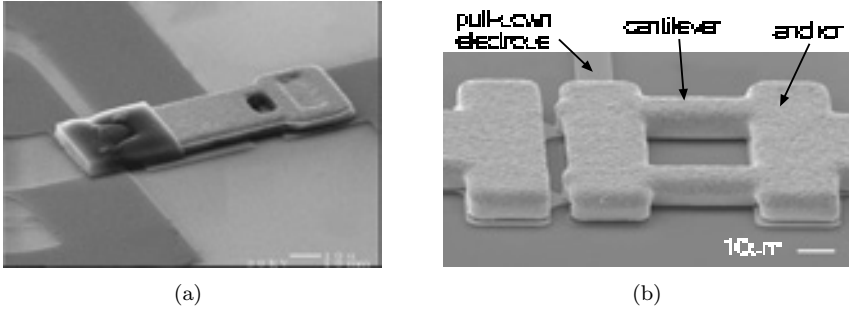
for an optical mirror array<sup>10</sup>. The membrane could be actuated at voltages between 30 and 50 V. The switch was intended for a 4-bit time delay phase shifter [103].

Also in 1995, the Rockwell Science Center, CA, USA showed a  $\text{SiO}_2$  cantilever switch with gold contacts, electrostatically actuated at 28 V and fabricated on a polyimide sacrificial layer. With a  $3 \mu\text{m}$  contact distance in the off-state, the switch had quite good DC and RF characteristics ( $S_{21,ON} < 0.12 \text{ dB}$ ,  $S_{21,OFF} > 50 \text{ dB}$  up to 4 GHz [94]). Another series switch recently developed at the center is shown in Figure 3(a).

In 1997, the Northeastern University, Boston, MA, USA, in cooperation with Analog Devices, published an electrostatically actuated (30–400 V) switch consisting of an electroplated nickel cantilever on top of a copper sacrificial layer, illustrated in Figure 4(a) [96]. Based on this design but later electroplated in gold, in 1999 they presented an extremely small in-line switch with a size of only  $30 \times 75 \mu\text{m}^2$ , shown in Figure 4(b). The off-state gap of the two parallel contacts with an area of only  $5 \mu\text{m}^2$  each was only about  $0.6 \mu\text{m}$  and the actuation voltage of the very stiff structure was above 60 V. Despite the extremely small contact distance, the isolation was quite good (40 dB at 4 GHz) and the contact resistance was  $1\text{--}1.5 \Omega$  with two parallel contacts, resulting in an insertion loss of about 0.15 dB [106]. The technology was licensed to Radant MEMS Inc., founded in May 2000, where the research and development work is being continued [107]. In 2002, a device very similar to the first design was demonstrated by NMRC, Cork, Ireland, most probably also in cooperation with the local MEMS division of Analog Devices in Limerick, Ireland, using sputtered titanium as the structural material and polyimide as the sacrificial layer [108].

In 1999, OMRON, Japan, one of the world's largest manufacturers of mechanical relays, introduced an electrostatically actuated (19.2 V) mono-crystalline silicon membrane switch, whose structural elements were etched in a SOI-wafer and transferred to a silicon target substrate by anodic bonding. The switch, probably still

<sup>10</sup>The digital micromirror array, developed by Texas Instruments since about 1983 and finally brought on the market in 1996, is a small success story of a highly complex MEMS volume product. By the end of April 2004, Texas Instruments announced the shipment of its three millionth digital light processing (DLP) device [102].



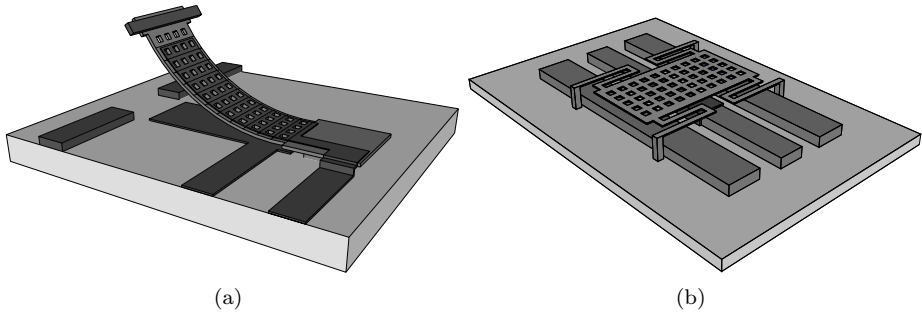
**Figure 4.** (a) 4-terminal microswitch by Analog Devices (now Radant MEMS [96]), and (b) in-line series switch by the same company [106].

one of the most maturely developed devices of its kind, was presented in a glass-frit package and had good performance from DC up to a few GHz [109].

Switch integration by wafer-scale transfer-bonding was also used for an electrostatically actuated switch at the University of California, Berkeley, USA, in 2000. A polysilicon/silicon nitride membrane, fabricated on a silicon dioxide sacrificial layer on the donor wafer, was transferred to the target wafer by gold to gold thermo-compression bonding. The contact distance could not be controlled accurately and varied from 1.5 to 10  $\mu\text{m}$ , resulting in actuation voltages of 25–120 V [110].

In 2000, the Korea Advanced Institute of Science and Technology (KAIST), presented an interesting switch concept for very low actuation voltage without sacrificing device performance. The switch consisted of a SiN leverage beam fixated by torsional springs and was endowed with a push-pull concept allowing active opening and closing. Actuation voltages as low as 5 V were reported, and the leverage effect of the beam resulted in a relatively large contact distance despite an electrode distance of only 1  $\mu\text{m}$ , providing fair RF performance ( $S_{21,ON} \approx 1$  dB,  $S_{21,OFF} \approx 40$  dB at 1 GHz and 30 dB at 4 GHz). Without the pull mechanism, the same switch achieved an isolation of only 17 dB at 4 GHz [111]. Independent of the proposed design at KAIST, the push-pull concept with torsional springs was adapted for switches at the University of Hawaii, Manoa, USA, published in 1999 [112], and at the University of California, Berkeley, USA, in 2000, for a design variation of the switch mentioned in the previous paragraph [110]. Also Infineon Technologies, Germany, picked up the idea in 2001 for a double switch operating at voltages below 10 V and fabricated in a BiCMOS compatible process [113]. Without being published otherwise, a patent on a torsional hinged SPDT switch was filed by people at TRW Inc., CA, USA, in 1999 [114], and inventors at Sarnoff Corp., NJ, USA, filed a patent on a relatively complex torsional switch structure in 2001 [115].

In 2000, the MIT Lincoln Labs in Lexington, MA, USA, presented a very compact and simple in-line switch design based on a curled electrostatic actuator operating at 35–80 V. The cantilever, shown in Figure 3(b), was composed of a  $\text{SiO}_2/\text{Al}/\text{SiO}_2$  sandwich structure with a controlled out-of-plane bending resulting in a 10–15  $\mu\text{m}$  contact distance between the platinum contacts [105]. The switch was also realized in a capacitive series switch configuration with an on-off capacitance ratio of over



**Figure 5.** (a) Curled cantilever switch with a large contact distance and relatively low actuation voltage due to the small effective electrode distance resulting in a zipper-like movement of the structure when closing; (b) typical membrane switch. The large electrode area and the soft-spring suspension lead to a low actuation voltage, but poor RF performance and decreased reliability.

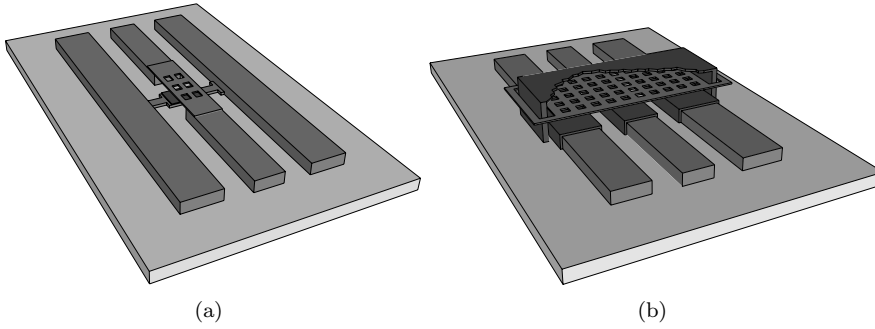
300 and was intended to be used in arrays for reconfigurable distributed microwave components [116].

In 2001, Samsung's internal research institute, SAIT, Korea, published an electrostatically actuated switch of very low actuation voltage. Its SiN membrane was fabricated on a photoresist sacrificial layer and the switch achieved very high RF isolation despite actuation voltages of only 3–5 V ( $S_{21,ON}=0.9$  dB,  $S_{21,OFF}=46$  dB at 2 GHz). However, no data was presented on the restoring force of the relatively weak structure and stiction problems might be expected<sup>11</sup> [117].

A system-on-a-package (SOP) integrated solution was presented by Motorola, AZ, USA, in 2002. The electrostatically actuated cantilever switch (30–60 V) with very good RF performance ( $S_{21,ON}=0.3$  dB,  $S_{21,OFF}=50$  dB at 2 GHz) was integrated together with a separate charge pump electronic chip into one hermetic package. The total power consumption of this "true" microsystem was 200  $\mu$ W at an actuation voltage of 3 V, thus meeting the low voltage requirements for wireless equipment. The Motorola approach is very interesting since it shows that the company's RF MEMS experts believe in a "high voltage" switch concept with better reliability due to a larger restoring force, even for targeting low voltage applications, made possible by an integrated electronic "high voltage" drive circuitry [99].

One of the other giants in the semiconductor industry, ST Microelectronics, France/Italy, is following a completely different track both in cracking the low-voltage problem and in the switch integration. In 2003, STMicroelectronics disclosed information on a combined electrothermally and electrostatically actuated SiN membrane switch jointly developed with their long-term research partner CEA-LETI, Grenoble, France. The switching contacts are closed by the thermal actuator (2 V, 20 mA, for 200  $\mu$ s), and clamped by an electrostatic actuator with 10–15 V. With an off-state contact distance of 3  $\mu$ m, the switch achieved a remarkable RF performance at 2 GHz ( $S_{21,ON}=0.18$  dB,  $S_{21,OFF}=57$  dB). The switch is integrated on chip-level (SOC) with electronics, since it is fabricated directly on top of CMOS circuits [118].

<sup>11</sup>For an estimation of the restoring force, see Section 2.2.1.



**Figure 6.** (a) Longitudinal capacitive shunt switch [93]; (b) hinged switch with completely unsupported membrane moving between the top and the bottom electrodes, with potentially very low actuation voltages (the top electrode is partially removed in the drawing to reveal the membrane).

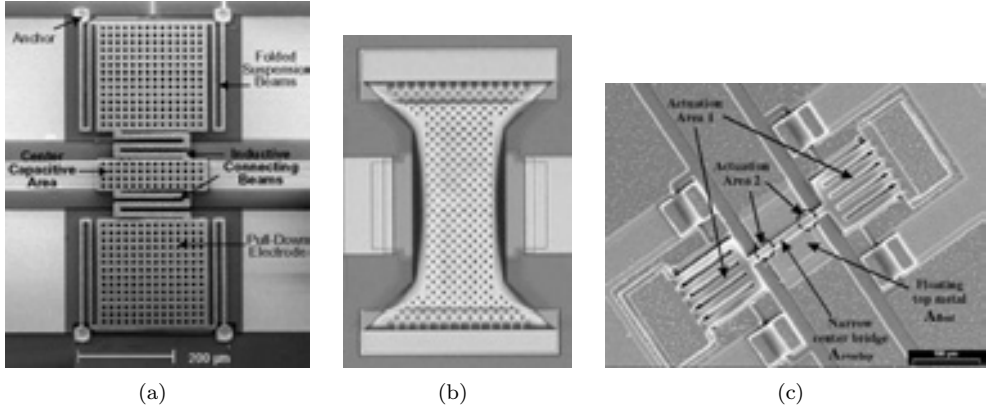
### Shunt switches

In 1998, the University of Michigan, Ann Arbor, USA, published on an electrostatically actuated (14–16 V) switch. The 2  $\mu\text{m}$  thick, electroplated gold membrane of this switch was embedded between bottom and top electrodes, the latter providing additional forces to open the switch. Aluminum was used as sacrificial layer material, and 50 nm of parylene was coated on the metal structure for isolation and anti-stiction purpose. The RF performance of the switch was quite good ( $S_{21,ON}^{12}=0.2$  dB at 20 GHz,  $S_{21,OFF} > 30$  dB for frequencies  $< 40$  GHz [119]). A low spring-constant switch recently developed by the RF MEMS team at the university is shown in Figure 7(a). In this switch, the inductance of the signal path in the down state is determined by the special geometry of the beams supporting the membrane, providing improved isolation because of resonance behavior in the target frequency range.

Also in 1998, at Raytheon, Dallas, USA, people formerly working on MEMS switches at Texas Instruments presented a capacitive shunt switch based on a 0.4  $\mu\text{m}$  thick aluminum membrane supported by 4  $\mu\text{m}$  tall, sputtered aluminum posts embedded in a polyimide sacrificial layer. Using silicon nitride as the dielectric material for the isolation layer, the switch achieved an on-off capacitance ratio of 80–110, which resulted in an insertion loss of 0.25 dB and an isolation of 35 dB at 35 GHz. The actuation voltage of the switch shown in Figure 7(b) was between 30 and 50 V [120]. Their first paper was followed by a series of publications addressing dielectric charging problems and lifetime characterization [121].

Researchers at the University of Illinois, USA, came up with a very interesting idea to lower the actuation voltage of a metal contact shunt switch. Their switch,

<sup>12</sup>The terms "on" and "off" are not used consistently for capacitive shunt switches in literature. In this book, they are used from a system point of view and refer to the signal propagation in the "black box MEMS-switch" and not to the mechanical actuation. Thus, "on" means that the signal can freely propagate from the input to the output and that the switch is not actuated (membrane up), and "off" means that the input signal is disconnected from the output whereas the switch is actuated (membrane down). MEMS as a technology is the slave of its application: the switch is "on" when the actuator is "off" and vice versa.



**Figure 7.** (a) Low spring-constant capacitive shunt switch by the University of Michigan [70]; (b) the classical Raytheon capacitive shunt switch (size about  $280 \times 120 \mu\text{m}^2$  [120]); (c) floating metal layer capacitive shunt switch by IMEC [125].

presented in 2000, consisted of a  $1 \mu\text{m}$  thick evaporated, unsupported gold membrane, fabricated between 2 polyimide sacrificial layers, and kept in place by hinges allowing free movement between the bottom and the top electrodes. Since no restoring spring elements are needed for this concept, actuation voltages as low as 9 V could be achieved. The DC to RF performance was also remarkable ( $S_{21,ON} < 0.1 \text{ dB}$  up to 40 GHz,  $S_{21,OFF} = 50 \text{ dB}$  at 1 GHz and 25 dB at 40 GHz [122]).

In 2000, the University of Michigan, Ann Arbor, USA, very active in the field of RF MEMS switching circuits, presented a cross-configuration of 4 electrostatically actuated capacitive shunt switches tuned for the Ka-band. This special arrangement, despite a small off-state contact distance of  $1.5\text{--}2.5 \mu\text{m}$  and actuation voltages as low as 15–25 V to move the  $2\text{--}2.5 \mu\text{m}$  thick gold bridge, resulted in reduced RF reflections and in high isolation over a wide frequency range ( $S_{21,ON} = 0.3\text{--}0.6 \text{ dB}$ ,  $S_{21,OFF} = 45\text{--}50 \text{ dB}$  at 22–38 GHz [123]).

Also in 2000, the LG corporate Institute of Technology, Seoul, Korea, presented a capacitive switch using a 190 nm thick strontium titanate oxide ( $\text{SrTiO}_3$ ) isolation layer with very high relative permeability, resulting in an on-off capacitance ratio of 600. The gold or copper switch membrane was electroplated on top of a polyimide or photoresist sacrificial layer. The overall RF performance of the switch was very impressive ( $S_{21,ON} = 0.08 \text{ dB}$  at 10 GHz,  $S_{21,OFF} = 42 \text{ dB}$  at 5 GHz [124]).

An interesting design variation was introduced in 2001 by Robert Bosch GmbH, Germany. The membrane of their longitudinal capacitive shunt switch is part of the signal line and, when lowered by applying approximately 25 V, is capacitively short-circuited by a ground potential connection between the two ground lines of the coplanar waveguide, as illustrated in Figure 6(a). The inductance of the signal path of that connection, together with the off-state capacitance, builds a series resonance circuit which is optimized in different designs for 24 or 77 GHz ( $S_{21,ON} < 0.3 \text{ dB}$ ,  $S_{21,OFF} > 33 \text{ dB}$ ). These frequencies suggest that the target application is automotive radar [93].

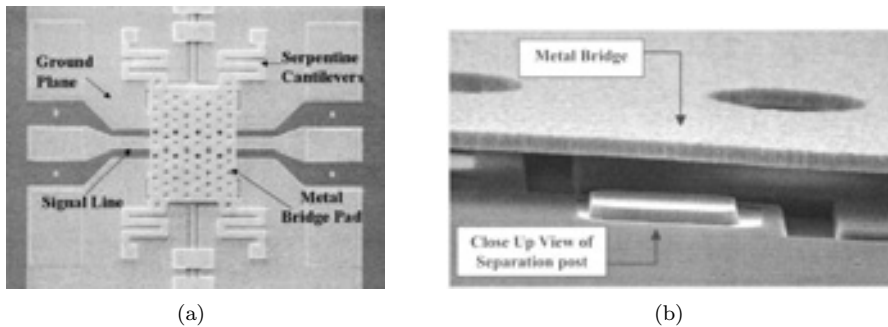
In 2002, the University of Michigan, Ann Arbor, USA, published an electrostatically actuated metal contact shunt switch using the classical membrane design of capacitive shunt switches. A  $0.8\text{ }\mu\text{m}$  thick gold membrane, fabricated on a  $\text{SiO}_2$  sacrificial layer, short-circuits the signal line when actuated at about 50 V [126].

Also in 2002, IMEC, Belgium, Europe's largest independent microelectronics research center, presented a very clever switch design. It has a 100 nm thick aluminum layer deposited on top of a 200 nm thick tantalum pentoxide ( $\text{Ta}_2\text{O}_5$ ) isolation layer, as shown in Figure 7(c). The additional metal plane acts as an electrically floating layer, which, when touched by a very narrow aluminum bridge, takes on RF ground potential and thus capacitively short-circuits the signal line. This concept has the advantages that 1) the area of the moving bridge does not determine the short-circuiting capacitance and can therefore be very narrow, with very little influence on the RF signal propagation in the on-state, and 2) since it is deposited and not mechanically movable, the floating metal layer is in very intimate contact with the isolation layer, resulting in an optimal, very high off-state capacitance as compared to the classical design where the capacitance is decreased by the surface-roughness between the touching layers and due to local bending of the membrane. The achieved on-off capacitance ratio of this switch design exceeds 600, which is remarkable for its small size and the materials involved [92]. The concept also eliminates the stiction problem caused by charge trapping in the thin dielectric layer, which is one of the major failure mechanism of capacitive shunt switches [75, Section 7.1].

Stiction caused by charge trapping does not occur in so-called all-metal switches, which do not involve any isolation layer. In 2003, the University of Illinois, USA, presented a metal-contact shunt switch using minute separation posts underneath the metal membrane as illustrated in Figures 8(a) and (b), increasing the overall switch reliability [127].

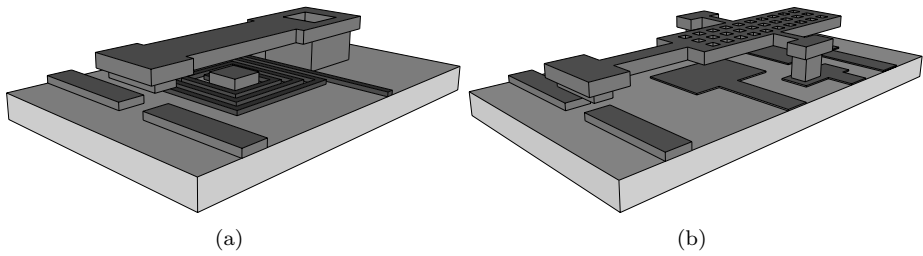
### Vertically actuated microrelays

As stated above, these devices are basically just design variations of metal contact series switches. Instead of targeting DC to RF signal switching applications, they



**Figure 8.** All-metal shunt switch by the University of Illinois: (a) overview of the membrane switch (size about  $250 \times 150\text{ }\mu\text{m}^2$ ), and (b) close-up SEM image of a separation post (dimensions not available [127]).





**Figure 9.** (a) Typical magnetostatic microrelay with a ferromagnetic armature closed by the magnetic field of a coil; (b) switch with leverage beam mounted on torsion springs: push-pull functionality, large contact distance and low actuation voltage.

are designed for larger signal currents from DC to low frequencies. Therefore, they are more susceptible to contact welding which, in addition to a larger contact distance required for better DC isolation and a larger contact force for low resistance, necessitates a stronger actuator design leading to a larger and slower device.

In 1997, the Fraunhofer Institute, Munich, Germany, presented an electrostatically actuated (20–90 V)  $\text{SiO}_2/\text{Au}/\text{SiO}_2$  cantilever microrelay fabricated onto a  $4\text{ }\mu\text{m}$  thick polyimide sacrificial layer. The residual stress in the sandwich structure made the beam curl out of plane and thus provided a sufficient restoring force and a large distance of 10–60  $\mu\text{m}$  between the gold switching contacts [97].

In 1998, Siemens Berlin, Germany, presented an electrostatic curled actuator named the "moving-wedge" actuator, which was expected to lower the actuation voltage. The switch could be operated at 24 V and was also tested successfully to survive over  $10^6$  switching cycles carrying a remarkable load of 100 mA between the AuNi5 contacts mounted on an interesting contact spring construction [95].

In 1998, CSEM, Neuchâtel, Switzerland, in cooperation with CP Clare N. V., Belgium, published on a magnetostatically actuated microrelay fabricated on a ferromagnetic FeSi substrate. The actuator consisted of a NiFe electroplated armature processed on a top wafer which was transfer-bonded to the substrate containing a two-layer copper coil with 127 turns. The power consumption of that device was 16 mW at an actuation voltage of 1.9 V, creating a contact force of 1 mN [128]. With a size of approximately  $5.3 \times 4.1\text{ mm}^2$ , the relay was hermetically packaged by a solder-bonding transfer process, had an off-state contact distance of 22  $\mu\text{m}$ , and was successfully tested up to  $10^5$  switching cycles when carrying a 10 mA load [129].

An electrothermally actuated cantilever switch was developed at Tohoku University, Sendai, Japan, and published in 2001. Its actuation mechanism consumed 20–80 mW at an operation voltage of 5 V. A special feature of this switch is spring metal contacts increasing the effective contact area at low contact forces. Also, the switch was fully packaged in a very sophisticated way, including electrical interconnections through a covering glass wafer [130].

In 2002, the Delphi Research Labs, MI, USA, presented an electrostatically actuated microrelay based on a copper electroplated blade closing a 3  $\mu\text{m}$  contact distance when a voltage of 25 V was applied. Furthermore, the switch was provided with an electronic arc suppression circuit to enhance its lifetime.  $1.7 \times 10^6$  cycles were achieved

at a 50 mA load, and even 5100 cycles when switching a 350 mA current [131].

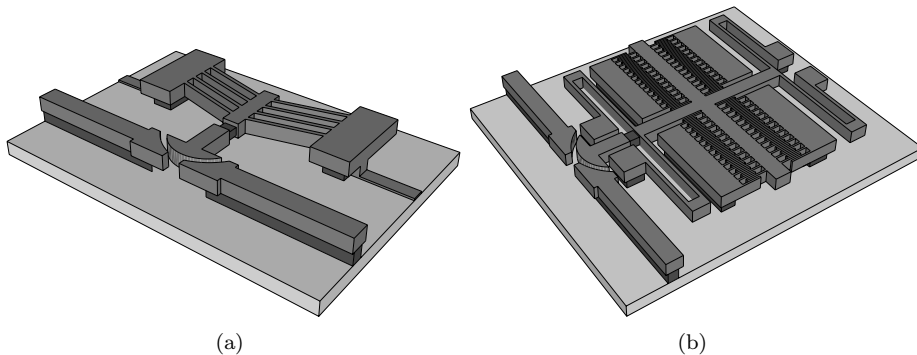
### Laterally actuated metal-contact series switches and microrelays

Lateral switches have also been the object of research activities. The major advantage of lateral actuation is that the actuator, the conductor pads, the support structures and often even the contacts can be fabricated in a single lithographic step. Also, an additional counter-actuator to actively open the switch does not necessarily impose increased fabrication complexity, and a large displacement is much simpler to design for lateral actuators than for vertically moving structures. Lateral switches with electrothermal V-shaped actuators are quite strong and an output force of up to 8 mN was reported at an actuation power of 180 mW [132]. These features make them perfectly suitable for microrelay applications. However, lateral switches are more area consuming than their vertically moving counterparts, and the deposition of the contact material on a vertical wall is much more difficult [75, 133, 134].

In 1998, the MEMS Technology Application Center (now Cronos Integrated Microsystems), NA, USA, presented an electrothermally actuated, nickel electroplated relay with a size of about 1 mm<sup>2</sup>. An actuation power of up to 300 mW created a contact force of 1–10 mN, resulting in contact resistances of 100–200 m $\Omega$  for 2  $\mu$ m thick gold contacts, and 500–700 m $\Omega$  for rhodium contacts. Switching a current of 10 mA over 1 million cycles was performed successfully [90]. In 2001, the company presented a VHF bandpass filter tunable by 6 electrothermally actuated microrelays with a current handling capability of 1 A each. The relay was found to survive 53 million cold-switched operating cycles carrying an RF power of 25 W between each switching transition [135].

An electrostatically actuated lateral switch was presented by the Fraunhofer Institute, Munich, Germany, in 1999. The switch was etched by an ICP tool into the 30  $\mu$ m thick device layer of an SOI wafer. The actuator bends a cantilever beam with the switching contact along a curved electrode at actuation voltages of 50–260 V [133].

In 2002, the Berkeley Sensor and Actuator Center at the University of California at Davis, CA, USA, published an electrothermally driven relay, consisting of LPCVD



**Figure 10.** Lateral microrelays: (a) electrothermal actuator; (b) electrostatic comb-drive actuators for both opening and closing.

polysilicon as the structural material for the actuator, the gold coated signal lines and for the contact head. The latter was thermally and electrically isolated from the actuator by a SiN plug. The thermal actuator was cascaded in an optimized way to create a total displacement of up to  $9.2\text{ }\mu\text{m}$  at a power consumption of 20–40 mW [136].

In 2003, the Massachusetts Institute of Technology, Cambridge, MA, USA, presented a lateral microrelay with a very interesting bistable electrothermal actuation mechanism. Being about 8 mm wide, the actuator was quite area consuming, and was designed to have a nominal contact force of 4 mN at a displacement of  $137\text{ }\mu\text{m}$ . Initial tests resulted in a contact resistance of  $60\text{ m}\Omega$  with  $2.5\text{ }\mu\text{m}$  thick electroplated copper contacts, making it able to carry a static load of 3 A [137].

### Switches using special fabrication technology

The switches discussed in this paragraph are probably not so maturely developed and might not have very outstanding performance, but they show new and interesting methods in fabrication technology especially suitable for switch integration and for merging MEMS and integrated circuits.

Almost all MEMS switches listed in this section are CMOS-compatible. Thus, they could be post-processed on a substrate already containing RF circuits since their highest process temperature does not exceed about  $350\text{ }^{\circ}\text{C}$ <sup>13</sup>. In a different approach, further integrating and lowering the costs, the switches are even fabricated by standard CMOS back-end technology. In 2001, the Oriental Institute of Technology, in cooperation with the National Taiwan University, both in Taipei, Taiwan, presented a lateral moving switch fabricated in a conventional  $0.6\text{ }\mu\text{m}$  single polysilicon three metal layer CMOS process. Only a single additional, mask-less free-etch process step had to be carried out to release the structures [138]. A similar approach was investigated by the IBM Watson Research Center, NY, USA, and presented in 2004. Here, vertically moving metal contact switches and mechanical resonators were fabricated by base copper damascene processes from IC interconnect technology. The switches could be activated at voltages of about 40 V and showed acceptable performance up to 2 GHz [139].

MEMS devices — and thus RF MEMS switches — are traditionally fabricated on wafers, the standardized substrate of the semiconductor industry. The substrate and fabrication costs are usually distributed over several thousands of chips, leading to very low costs per device despite the expensive clean-room technology. However, certain applications such as antennas in the lower GHz range cannot be miniaturized and require space in the order of many square-centimeters. On the other hand, their rather large process tolerances do not require semiconductor manufacturing precision. Thus, it would be too expensive to waste a semiconductor substrate with a yield of only a few devices per wafer, especially, since the antenna structures could be fabricated very cheaply using conventional printed circuit board (PCB) technology. Whenever MEMS switches are intended to reconfigure such antennas, they should not be fabricated together with the antennas on the same substrate for these economical

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<sup>13</sup>The absolutely highest temperature allowed in CMOS post-processing is  $480\text{ }^{\circ}\text{C}$ . However, also the exposure time of the circuits to these temperatures is an influencing factor. As a rule of thumb, all post-fabrication steps above  $400\text{ }^{\circ}\text{C}$  are problematic.

reasons. Therefore, the idea to fabricate the MEMS switches on a large-area substrate is close at hand. In 2003, the University of California at Irvine, USA, published on a fabrication process for directly constructing RF MEMS capacitive switches on a laminated microwave PCB. Traditional semiconductor/MEMS fabrication processes cannot be used on these substrates, and special processes such as wet metal etching, compressive molding planarization, and a low temperature, high-density, inductively coupled plasma chemical vapor deposition (HDICP-CVD<sup>14</sup>) were used to fabricate the aluminum-membrane switches on a standard, low-cost FR4 substrate. The switches could be actuated at 35 V and had acceptable RF performance in the frequency range from 5 to 30 GHz [140].

In 2001, the University of Colorado, USA, published on flexible circuit based RF MEMS switches, fabricated by using the typical flexible circuit processes and materials. The switch membranes consisted of low-cost Kapton-E polyimide with copper cladding, bonded at 130 °C to a patterned benzocyclobutene (BCB) spacer layer on RT/Duroid 6002 substrate with a copper transmission line. With a chip size of about 1 cm<sup>2</sup> and a contact distance of 40–50 μm, the switches could be actuated at about 70–90 V [141].

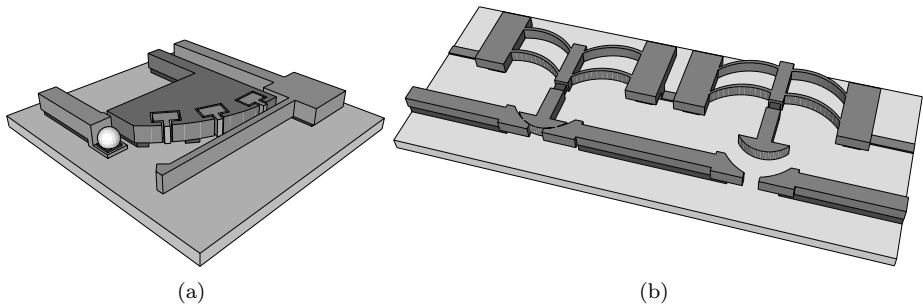
### **Other interesting switch types**

Acceleration switches are devices closing a contact by a mass moving due to external acceleration. These switches are not typical RF MEMS switches, but research and development in this area has been going on for more than a decade and the engineering solutions of these devices, especially in integration and packaging, are very mature. Compared to conventional piezoresistor or capacitive read-out accelerometers, these switches have the advantages of high immunity to electromagnetic influence (EMI), simple signal preconditioning, and very simple fabrication and integration with ICs. All these advantages make them perfectly suitable for their high-volume target market, even though the price of analog electronics and analog-to-digital converters (ADC) has been dropping rapidly over the last decade. In 1996, KAIST, Korea, presented a pre-stressed bimorph beam switch with a 7 μg proof mass. The switch was tunable in its acceleration sensitivity by a voltage of 0–76 V [142]. Siemens AG, later Infineon Technologies, Germany, has also been quite active in this field since the middle of the 1990s. They have been working on laterally moving, CMOS compatible, electroplated structures which can be fabricated directly on top of integrated circuits [143]. Furthermore, they presented an array of acceleration switches with different acceleration thresholds, thus a "quasi-analog" accelerometer [144]. In 2002, the Tohoku University, Sendai, Japan, in cooperation with Sensor company, Hanyu, Japan, presented a bulk micromachined acceleration switch whose on-state duration, determined by the mass, squeeze-film damping and by a clever mechanism to extend the hold time, is designed to convert the acceleration magnitude [145].

The next paragraphs introduce a few switches with interesting and certainly inspiring actuation mechanisms. The different designs show the possibilities of the creative and unlimited MEMS world when solving engineering problems.

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<sup>14</sup>for deposition of silicon nitride below 100 °C



**Figure 11.** Interesting switch concepts: (a) mercury drop switch with cantilever beam bending along a curved electrostatic electrode [146]; (b) mechanical bistable lateral switch shown in the on and in the off-state [134].

Mercury-wetted macro-relays have been used since 1940 to solve the reliability and life-time problems of solid-solid contact relays [147]. At the University of California, Los Angeles, CA, USA, this idea was adapted for a micromachined relay based on mercury microdrop contacts in 1996. A suspended microheater in a deionized water reservoir grows a vapor bubble whose pressure induces the movement of a mercury droplet short-circuiting two metal contacts [148]. In another embodiment, a mercury drop evaporated on a gold pad is contacted by a laterally moving, electrostatically actuated polysilicon cantilever, as illustrated in Figure 11(a). However, the contact resistance of the mercury contact relays was unsatisfactory and above  $1\text{ k}\Omega$  [146].

Mechanical stability in both the on and the off-state is a highly ranked design feature especially for electrothermally actuated switches. One attempt to achieve this goal was presented by the New Jersey Institute of Technology, Newark, NJ, USA, in 1998. Their switch consists of a two-segment multimorph cantilever. The bistable relay could be operated with thermal power pulses of  $10\text{--}12\text{ mW}$  for  $0.5\text{ ms}$  and the lateral displacement of the contacts was about  $30\text{ }\mu\text{m}$  [149].

In 2002, Ford, MI, USA, in cooperation with the Brigham Young University, Provo, UT, USA, published an electrothermally actuated micromachined relay with a size of  $1.92\text{ mm}^2$ . The switch, schematically drawn in Figure 11(b), is based on an interesting bistable snapping mechanism with a very large contact displacement of  $88\text{ }\mu\text{m}$ . Even though optimized for the chosen geometry, only a relatively low contact force of less than  $25\text{ }\mu\text{N}$  could be reported [134].

In 2001, the company MICROLAB Inc., Chandler, AZ, USA, proposed an interesting zero-power magnetostatic switch concept based on a micro magnetic latching mechanism. Only a short positive or negative current pulse ( $80\text{--}120\text{ mA}$  at  $5\text{--}6\text{ V}$  for  $0.4\text{ s}$ ) is needed to reverse the magnetic polarity of a permalloy cantilever suspended by torsional springs in an external magnetic field. Depending on the magnetic polarization in the cantilever, a positive or negative torque is created, bringing the switch into two stable positions. The contact forces achieved with this design were relatively small ( $40\text{ }\mu\text{N}$ ) and the main disadvantage of the whole concept is the need of a permanent magnet mounted under the substrate [100].

2.1.5 Applications of MEMS switches

MEMS switches are intended to switch the propagation path of an analog signal with a very high signal purity. This function might be used to connect/disconnect an electrical potential, or to select signals, for instance the incoming and the outgoing signal of an antenna, or to reconfigure a sub-system. Typical applications of the latter type are switching of filter banks, tuning of filters by switching capacitors or inductors [135], switching of delay lines in phase-shifters [47–50], impedance matching by switching stub-lines [59] or capacitors [60], and configuration of antenna patterns or frequencies by switching antennas or parts of an antenna [53–56].

Depending on the frequency specifications, either capacitive or metal contact switches are preferred for certain applications.

To switch DC to RF signals, typically required in high-quality test and measurement equipment, the most demanded properties are an extremely large bandwidth from DC to a few tens of GHz with very good signal properties over the whole bandwidth. The switching speed and the power consumption are less important. Metal contact switches are good candidates for these applications.

Microrelays are intended for applications where DC to low frequency currents have to be switched. Here it is important to galvanically decouple the input from the output in the off-state, and to have a small series resistance for a low voltage drop and low power dissipation in the on-state.

Capacitive shunt switches are usually designed for frequencies above 1 GHz and

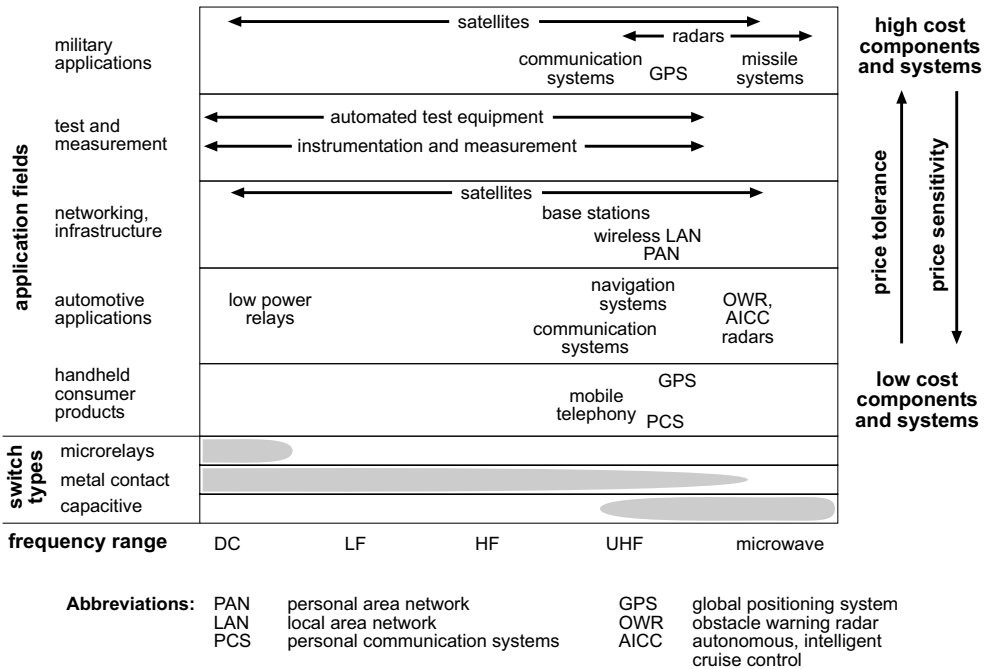


Figure 12. Main application fields of MEMS switches.

small bandwidths. Especially when targeting wireless equipment, low power consumption and small size are required.

Military and space applications also favor MEMS components because of their robustness against external influences such as radiation, temperatures and strong electromagnetic fields, and because of their high shock resistance.

Switch reliability and a large number of switching cycles of the order of tens of billions are equally important for most applications.

The diagram in Figure 12 gives an overview of different MEMS switch applications, displayed over the signal frequency. The diagram also gives information on the price sensitivity of the specific applications. It can be expected that MEMS switches are first going to be introduced in less cost critical areas before they might be considered for low-cost, high volume markets.



*It is better merely to live one's life,  
realizing one's potential,  
rather than wishing  
for sanctification.*

*He who lives in filial piety and love  
has no need of ethical teaching.*

*When cunning and profit are renounced,  
stealing and fraud will disappear.  
But ethics and kindness, and even wisdom,  
are insufficient in themselves.*

*Better by far to see the simplicity  
of raw silk's beauty  
and the uncarved block;  
to be one with onself,  
and with one's brother.*

*It is better by far  
to be one with the Tao,  
developing selflessness,  
tempering desire,  
removing the wish,  
but being compassionate.*

*Lao Tzu  
Tao Te Ching, Chapter 19  
(translation by Stan Rosenthal)*



*"I refuse to prove that I exist," says God,  
"for proof denies faith, and without faith,  
I am nothing."*

*Douglas Noel Adams  
1952–2001, British Science Fiction Writer,  
in "The Hitch Hiker's Guide to the Galaxy"*

## 2.2 Design considerations of electrostatically actuated metal-contact switches

A quick search in the INSPEC database of scientific publications in April 2004, using the search terms "MEMS switch"  $\vee$  "microswitch"  $\vee$  "microrelay", results in approximately 1600 records. Most publications report on new switch designs with characterization results, discussing their general advantages and talking about their applications. Details are often not disclosed to protect intellectual property and to stay ahead of competitors. Only a few sources specifically address design related issues, discuss the influence of the variety of design parameters, and compare different concepts with each other [75, 89, 123, 150–153].

This section addresses some selected, typical and in my opinion important design problems of electrostatically actuated metal-contact switches. It reports on the influences of various design parameters on each other and tries to provide a feeling for designing a MEMS switch. It also covers the dominant reliability problems of contact switches.

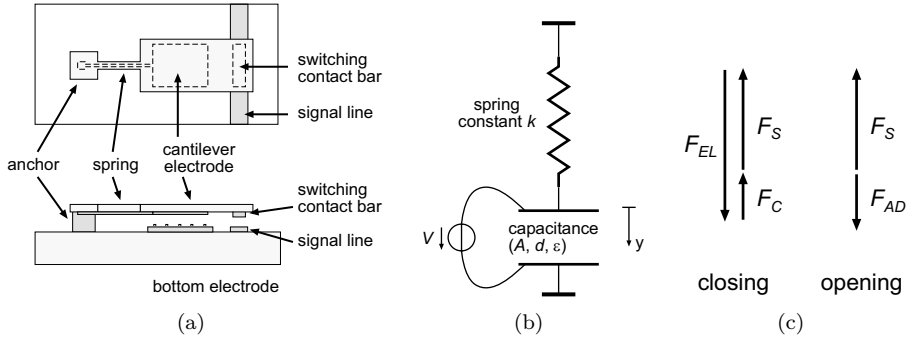
The high art of managing complex MEMS devices is still their fabrication and the knowledge about the mechanical behavior of multi-layer structures. Material issues such as the properties of low-stress free-standing films are not of minor importance but are not discussed in this section whose focus is on a device and system level, and I would like to refer to the fabrication chapters of standard (RF) MEMS source-books [75, 154, 155]. However, material related issues of electrical contacts under low mechanical forces are a crucial point for microrelays and metal-contact microswitches, and are therefore briefly addressed in Section 2.2.3.

The packaging and the integration of RF MEMS switches, two of the major obstacles on their way to commercialization, are discussed in Section 2.3.5.

### 2.2.1 Actuation voltage, displacement, contact and restoring forces

These four switch parameters are tightly coupled to each other, which makes good switch design very difficult. Figure 13(a) is a schematic drawing of a simple electrostatically actuated cantilever switch, and Figure 13(b) shows a simplified quasi-static electromechanical equivalent circuit model of such a cantilever-spring or membrane-spring based switch. These are the most common MEMS series switch types and are discussed in this section. The active forces, when establishing the contact and when opening it, are illustrated in Figure 13(c). The overall goal of any switch design is to balance these forces to achieve the desired contact performance, expressed in the contact resistance and in the contact reliability.

Figure 14 shows how the design parameters actuation electrode area  $A$ , distance between the electrodes  $d$  (not to be confused with the switching contact distance),

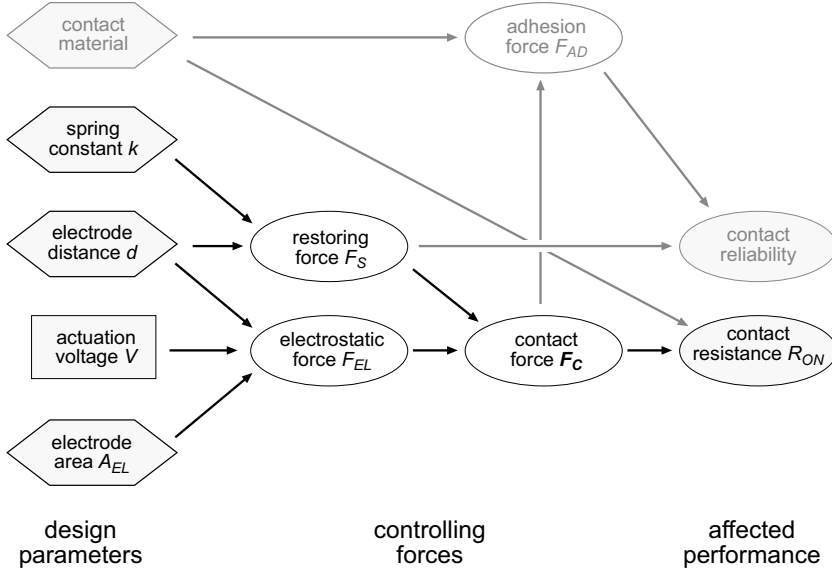


**Figure 13.** (a) Top and side view of a cantilever based switch; (b) simplified mechanical equivalent circuit model; (c) active forces when establishing and when opening the contact (forces as defined in the text and in Figure 14).

actuation voltage  $V$  and spring constant  $k$  control the active forces and the contact performance of the switch model. The choice of contact material has also an important influence, which will be discussed in Section 2.2.3. The distance between the electrodes is closely connected to the contact displacement, which determines the DC and RF isolation of the switch, see Section 2.2.2. The isolation is not directly considered in the discussion of this section, but it should be kept in mind that certain very low electrode distances practically do not make sense for the discussed model, because the switch would not provide acceptable isolation.

Basically, a high contact force  $F_c$  results in a low contact resistance, and a high restoring force is desired to prevent contact stiction caused by microwelding. A high restoring spring force  $F_s$  can be achieved by a relatively stiff structure and a large displacement, both requiring an increased electrostatic force  $F_{el}$  to pull down the structure. Also, increasing the restoring force demands an even stronger electrostatic actuator to maintain the desired contact force (see Figure 13(c)). The electrostatic force can only be increased by the actuation voltage or by a larger actuation electrode area, resulting in a larger switch size. To complicate it even more, the effective restoring force is the spring force minus the adhesion force  $F_{ad}$ . The adhesion force is very difficult to predict or to control, since the contact physics in microswitches is still not very well understood [153, 156].

The actuation voltages of current electrostatic switch designs are between 15 and 80 V to achieve sufficient switching performance [70, 75]. These voltages are by far not compatible with the voltage levels used in electronic products. Thermal and magnetostatic actuators can be designed in a way resulting in actuation voltages below 5 V, but their high power consumption, unless using a bistable latching mechanism, is even less acceptable especially when targeting wireless applications. Also, switches based on electrostatic actuation are the most promising designs in terms of reliability and wafer-scale manufacturing possibilities [70]. Another strong advantage of electrostatic actuation mechanisms is the fact that the maximum force is created in the end position with touching contacts. The last paragraph of this section lists the advantages and disadvantages of common actuation mechanisms for MEMS switches with almost-zero power consumption in both the on and the off-states.



**Figure 14.** Illustration of the connection between the design parameters and the active forces in the switch model and how they affect its performance in terms of contact resistance and reliability.

The pull-down voltage required by a switch as illustrated in Figure 13(a) can be calculated approximately with the simplified mechanical model shown in Figure 13(b). The electrostatic force and the restoring spring force are given by [7]

$$F_{el} = \frac{1}{2} \epsilon_0 \epsilon_r \frac{A}{(d - y)^2} V^2 \quad (1)$$

$$F_s = -ky \quad (2)$$

with  $\epsilon_r$  the effective relative permeability,  $A$  the electrode area,  $d$  the initial distance between the electrodes,  $V$  the actuation voltage,  $k$  the spring constant reflecting the mechanical stiffness, and  $y$  the idealized parallel deflection of the beam. The upper electrode snaps against the lower one if the electrostatic force is larger than the spring force. That is the case for when the distance between the electrodes falls below a critical distance, and can be explained by the positive feedback mechanism in the electrostatic actuation. The attracting electrostatic force grows with the square of  $\frac{1}{d}$ , whereas the restoring spring force only grows linearly with the distance. The critical distance is  $y = \frac{1}{3}d$  and is independent of the design geometry<sup>15</sup> [75, Section 2.6].

<sup>15</sup>This fact also limits the theoretical maximum tuning range of a simple, parallel-plate based MEMS tunable capacitor to  $(C_{max} - C_{min}) / C_{min} = 50\%$  [7]. There are, however, active control methods to extend the stable operation range even for the simple parallel-plate tunable capacitor [157].

**Table 7.** Actuator geometry and performance comparison of a few switches where the model of Figure 13 is applicable (all data, unless calculated, from the given literature references and [75, Chapter 5]).

Company, institute	Spring constant $k$ in $\text{N}\cdot\text{m}^{-1}$	Estimated electrode distance <sup>a</sup> $d$ in $\mu\text{m}$	Electrode size $A$ in $\mu\text{m}^2$	Actuation voltage $V$ in V	Contact force $F_c$ in $\mu\text{N}$	Estimated restoring force <sup>b</sup> $F_s$ in $\mu\text{N}$
Rockwell [104]	15	3.8	$2\times 75\times 75$	50–60	– <sup>c</sup>	38
Motorola [99]	35–40	2.6	$100\times 80$	40–60	– <sup>c</sup>	64
HRL [98]	4–8	3.9	$100\times 100$	30–40	400	16
Analog Devices [106]	100	0.8	$15\times 25$	60–80	100	53
OMRON [158]	400	3.5	$2\times 800\times 1600$	13–20	5000	$933^d$
Univ. of Michigan [159]	20–50	2.2	$2\times 50\times 100$	30–40	– <sup>c</sup>	51
Samsung [117]	1	4.0	$2\times 300\times 150$	3–5	– <sup>c</sup>	2.7

<sup>a</sup>calculated from  $A$ ,  $k$ , and  $V$ :  $d = \frac{3}{2} \sqrt[3]{\frac{\epsilon A V^2}{k}}$

<sup>b</sup>calculated by  $F_s = ky$ , approximating  $y = \frac{2}{3}d$

<sup>c</sup>not published; an approximated calculation is not applicable

<sup>d</sup>the value reported in the literature is 1000  $\mu\text{N}$

Thus, the pull-down voltage of the cantilever switch can be estimated by

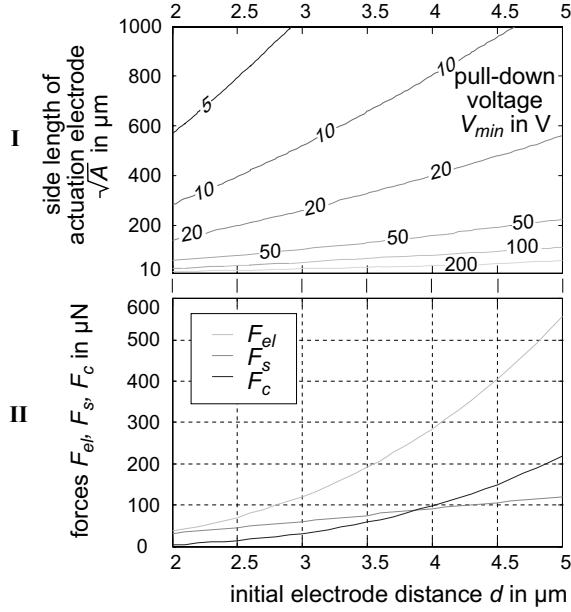
$$V_{min} \geq \sqrt{\left(\frac{2}{3}\right)^3 \frac{k d^3}{\epsilon_0 \epsilon_r A}} \quad (3)$$

It can be seen that the minimum required actuation voltage is proportional to  $d^{\frac{3}{2}}$  and indirectly proportional to  $\sqrt{A}$ .

Table 7 gives an overview of the actuator geometries, the actuation voltage and the contact and restoring forces of a few selected switches where the discussed model is applicable. These examples show that a stiff structure results in a larger restoring force, but requires either larger electrode areas or a higher actuation voltage unless the electrode distance is very small (which basically decreases the signal isolation because the contact distance is then also smaller for the discussed model).

Let us participate at a little game between the forces  $F_{el}$ ,  $F_s$  and  $F_c$  on the four-dimensional board of  $A$ ,  $d$ ,  $V$ ,  $k$ .

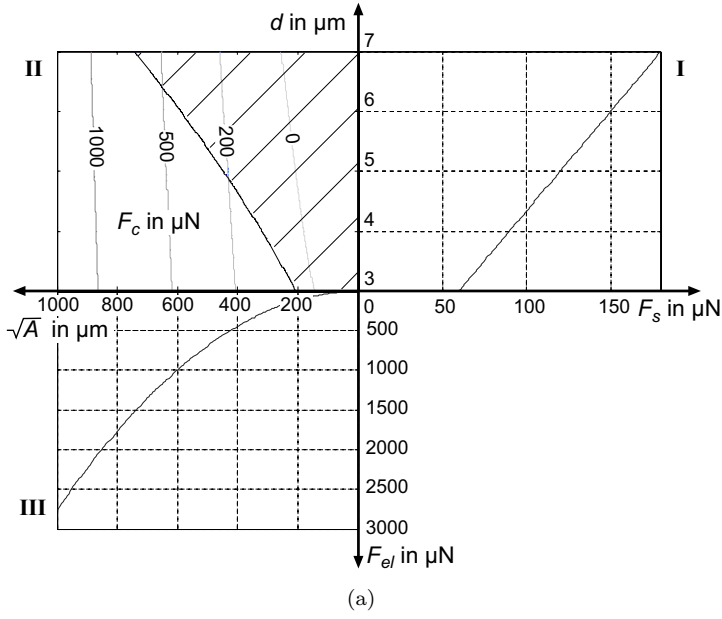
Figure 15, upper part, shows a contour plot of the pull-down voltage  $V_{min}$  depending on the initial electrode distance  $d$  and on the side length of the actuation electrodes  $\sqrt{A}$ , for a spring constant of  $30 \text{ N}\cdot\text{m}^{-1}$ , representing a typical MEMS switch. Most vertical switch designs have a spring constant of  $15\text{--}40 \text{ N}\cdot\text{m}^{-1}$ , resulting in a restoring spring force of  $30\text{--}120 \mu\text{N}$  for a typical electrode distance of  $3 \mu\text{m}$  [75, Section 2.9]. The lower graph in Figure 15 shows the corresponding electrostatic force, the restoring spring force and the contact force in the down-state, assuming a to-



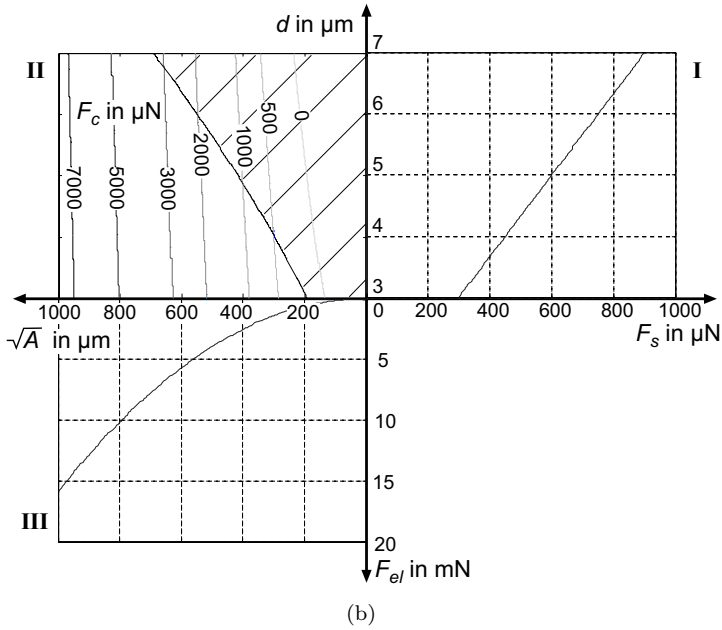
**Figure 15.** Simulation results of the slightly corrected (see text) switch equivalent model from Figure 13(b), for a spring constant of  $30 \text{ N}\cdot\text{m}^{-1}$ . Upper graph: dependence of the pull-down voltage on the initial electrode distance and on the side length of the actuation electrode area; lower graph: electrostatic, spring and contact forces depending on the initial electrode distance (see text). The two diagrams are connected by the common X-axis.

tal deflection down to a remaining electrode distance  $d_{\text{cont}}$  of  $1 \mu\text{m}$ . The resulting contact force is calculated by  $F_c = (F_{el} - F_s) \cdot 50\%$ , whereas the weighting factor of 50% reflects the fact that the contact force is only about 40–90% of the pull-down force, with the remainder part contributing to beam flexure or being lost by touching electrodes [75, Section 2.9]. The two graphs have the initial electrode distance  $d$  as the common axis, which allows to find the acting forces of each point in the pull-down voltage plot. This does not mean that the electrostatic force is area-independent but dependent on the initial electrode distance, as might be simulated by the lower graph, since the product  $A \times V^2$  is constant in the upper graph along any vertical line with  $d=\text{constant}$ . In other words, a larger electrode area allows a lower actuation voltage to create the same forces, which is reflected in the graphs.

In Figure 16, the voltage is kept constant and the active forces are displayed depending on the electrode size and on the initial gap distance (quadrant II). Quadrant I shows the related restoring spring force and quadrant III the electrostatic force. Also here, the graphs of the different quadrants are connected by their common axes. The textured area in the second quadrant shows the "forbidden" design area: switches with a geometry chosen out of this area do not snap down for the given actuation voltage. Sub-figure (a) represents a weaker switch design with a spring constant of  $30 \text{ N}\cdot\text{m}^{-1}$  and an actuation voltage of  $25 \text{ V}$ , and Sub-figure (b) displays a stronger switch design with a spring constant of  $150 \text{ N}\cdot\text{m}^{-1}$  and an actuation voltage of  $60 \text{ V}$ . For both



(a)



(b)

**Figure 16.** Active forces depending on design geometry: (a) weak switch design with a spring constant of  $k=30 \text{ N}\cdot\text{m}^{-1}$ , an actuation voltage of  $V=25 \text{ V}$ , and a down-state electrode distance of  $d_{cont}=1 \mu\text{m}$ ; (b) stronger design with  $V=60 \text{ V}$ ,  $k=150 \text{ N}\cdot\text{m}^{-1}$ , and  $d_{cont}=1 \mu\text{m}$ .

designs the remaining electrode distance in the down state is assumed to 1  $\mu\text{m}$ . The "stronger" design provides quite large contact and restoring forces suitable for switching currents with high reliability, especially when using a large electrode area. An example of this type of switch is the OMRON switch [109, 158, 160]. A contrasting design philosophy was realized in the Samsung metal contact switch, actuated at very low voltages with average sized actuation electrodes. This is achieved by an extremely low spring constant at the price of a very small restoring force and therefore probably also low contact reliability [117]. For a direct comparison see also Table 7.

The discussed model does not take into account any isolation layer between the two electrodes. Such an isolation layer does not substantially decrease the necessary pull-down voltage, since most of the electrostatic energy is stored in the free space and not in the dielectric layer, but would increase the contact force. This is only relevant for very weak switch designs (to switch low power signals), which require a low on-resistance but tolerate a relatively low restoring force. For more reliable switch designs, the electrostatic actuator has to be quite strong to overcome the required stiffness of the structure. A strong actuator already results in a large contact force and does not necessarily have to be increased by an additional dielectric layer (see Figure 16(b)). Not to give a wrong impression: the main function of a dielectric layer is still the electrical isolation between the electrodes. In general, isolation layers should be avoided by using a so-called all-metal switch design because of the tendency to electrostatic stiction, as discussed in Section 2.2.4.

All these reflections and calculations on the model show that it is not so easy to create an effective switch design with low actuation voltages (for electronic compatibility), small electrode areas (for small chip size and low cost per switch) and large contact distance (for high isolation), by maintaining a reasonable restoring force (for contact reliability) and contact force (for low resistance). The RF MEMS community is aware of this problem and many different electrostatically actuated switch concepts seeking clever design solutions have been published over the last decade:

- **Curled cantilever/membrane:**

Out of plane bending structure due to a residual stress gradient or a stress difference in a multi-layer structure, providing low effective electrode distance ("zipper-like" or so-called touch-mode actuation [161]) on the fixed side but with large deflection on the freestanding (contact) side [95, 97, 105], see Figures 3(b) and 5(a) on pages 15 and 17, respectively.

**Advantages:** low actuation voltage and large contact distance

**Disadvantages:** stress level difficult to control; increased susceptibility to electrode stiction because of large areas in very close contact

- **Curved or stepped electrodes:**

The "zipper" effect mentioned above is achieved by a curve-shaped undeformable counter-electrode; typically used in lateral actuators because of their simple fabrication [133, 146]; possibility for vertical actuators: stepped electrodes [162].

**Advantages:** low actuation voltage, increased contact distance

**Disadvantage:** in the case of a vertical actuator increased fabrication complexity

- **Push-pull concepts to provide an active restoring force:**

- **Push-pull mechanism with additional top electrode:** Membrane or cantilever with electrode moving between the two fixed top and bottom electrodes [119].

**Advantages:** active opening and closing, allows flexible structure with low actuation voltage

**Disadvantages:** increased fabrication complexity, not necessarily larger contact distance since the electrostatic forces of the basic parallel-plate configurations are very small over a larger distance

- **Torsional suspended actuators with leverage effect:** Push-pull electrodes on both sides of the rotational axes of a plane suspended on one of its symmetry axes, tilting the plane when actuated to one side or to the other. The switching contact is mounted on the far outside edge or on a leverage beam to further extend the displacement [110–113], see Figure 9(b) on page 21.

**Advantages:** active closing and active opening, very low actuation voltage, large contact distance

**Disadvantages:** design of the suspension stiff enough against vertical movement but with very low rotational spring constant

- **Pull-mechanisms to separate the contacts:** Fulcrum contact structure providing a rotational axis in the down-state to open the contacts together with a part of the pull-down electrode [163].

**Advantages:** active restoring force with very little additional fabrication complexity, low actuation voltage

**Disadvantages:** does not improve the open contact distance

- **Free-moving membrane actuator:**

Completely freestanding membrane moving vertically along hinges between a top and a bottom electrode [122], see Figure 6(b).

**Advantages:** theoretically no restoring spring energy → very low actuation voltage

**Disadvantages:** wedging membrane on the hinges (friction and stiction problems), complicated fabrication, low opening force at a reasonable contact distance

Despite all of these possibilities, the cantilever-spring or membrane-spring concept as discussed in this section still seems to be the most favored one, mainly because of its robustness in fabrication, its simplicity in usage and because of its predictability. The probably most mature switch designs, the OMRON, the Motorola, the Analog Devices/Radant, the Rockwell Science Center and the HRL switches, are based on this concept. The first three of them have solved the problems of the membrane-spring concept in different ways that are interesting to compare (see also Table 7 on page 32):



**OMRON switch:** very stiff spring for high restoring force; very large electrode areas for relatively low actuation voltage

**Motorola switch:** stiff spring for high restoring force; small electrode areas requiring large actuation voltage, accomplished by a more complex electronic drive circuitry

**Analog Devices switch:** very stiff spring for high restoring force; very small contact gap and small overlapping contact areas, resulting in relatively low actuation voltages and requiring only small electrode areas

It is also very interesting to observe that one of the Analog Devices/Radant switch designs, the Motorola, the HRL and a recently presented high-power switch from the University of Michigan [164] show a bewildering likeness in their design. This tendency seems to anticipate a "natural selection process" in MEMS metal contact switches, which hopefully does not mean that there will not be any further substantial improvements in the performance of MEMS switches.

The switch presented in this thesis is a combination of the push-pull concept with top electrode and a double-curved actuator using the low-voltage "zipper"-effect both with the bottom and the top electrodes (see Section 3).

For completeness, the following list gives an overview of the actuation voltages, displacements and active forces in electrostatically actuated switches compared to alternative actuation mechanisms with zero power consumption in both the on and the off-state:

- **Electrostatic actuators:**

**Advantages:** simple in fabrication and operation, very well investigated, maximum force in the contact position, strong contact force for small chip sizes, pull-in hysteresis<sup>16</sup>

**Disadvantages:** requires relatively high actuation voltages, the displacement is basically limited to a few micrometers

- **Electrothermal, lateral moving actuators with mechanical latching mechanisms** [134, 137]:

**Advantages:** large displacement, low actuation voltage, fabrication relatively simple

**Disadvantages:** very large size, maximum force during the transition and not in the end positions, mechanical latching provides relatively small contact forces

- **Electrothermal with electrostatic clamping** [118]:

**Advantages:** low voltage, large displacement, maximum force in the contact position, large contact force, small size

---

<sup>16</sup>Especially for microrelays, a very important feature of electrostatic actuators: the hold-voltage is lower than the actuation voltage, which defines the switching state very well even at unstable control voltages.

**Disadvantages:** complex fabrication

- **Magnetostatic, latching by switching magnetic polarization** [100]:

**Advantages:** low actuation voltage, restoring force to a certain extend controllable by the actuation current

**Disadvantages:** low displacement, low contact force, low restoring force, complex fabrication involving the mounting of a permanent magnet

- **Other, less commonly used actuation mechanisms:**

- **Electrothermal, vertically moving, with mechanical latching mechanism** [149]: low actuation voltage, large vertical and lateral, but uncontrollable displacement, low contact force
- **Piezoelectric:** so far only used for a tunable capacitor based on a PZT layer [31], requires rather thick piezoelectric layers for a displacement of a couple of micrometers with an acceptable contact force, difficult fabrication and process incompatibility<sup>17</sup>

Other possible combinations of these actuation and latching mechanisms are not considered here, since, to the knowledge of the author, they have so far not been reported in the literature.

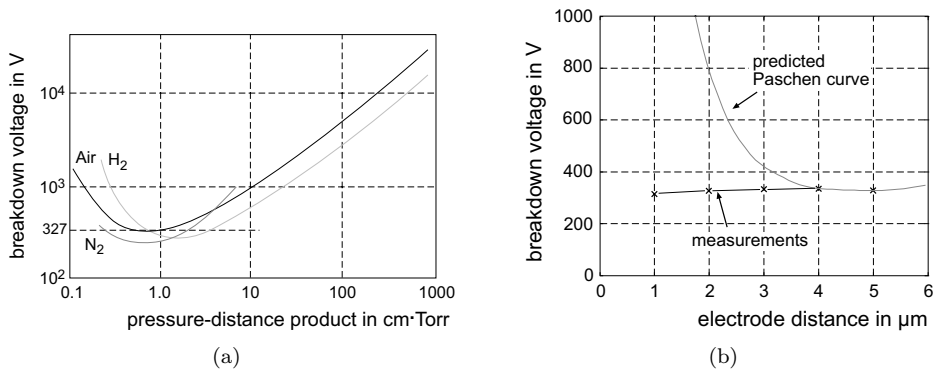
### 2.2.2 Contact distance vs. isolation

#### DC isolation

The DC voltage isolation of a switch is given by the maximum voltage which can be applied in the off-state between the input and the output without having a breakdown between the separated contacts. By experience, such a breakdown occurs when a certain field strength is reached in the isolation gas (for dry air, atmospheric pressure, room-temperature, about  $3 \text{ kV}\cdot\text{cm}^{-1} \cong 30 \text{ V}\cdot\mu\text{m}^{-1}$ ). Therefore, the isolation or so-called stand-off voltage basically depends on the atmosphere inside the switch package and on the contact distance.

For dimensions of the order of a few tens of micrometer the breakdown mechanisms are different. In large scale systems, electrical breakdown arises due to electrons accelerated in the electric field colliding successively with molecules which are ionized and thus result in more electrons which are subsequently accelerated by the field and cause further collisions (Townsend or avalanche effect [165]). For dimensions of the same order as the mean free electron path to collision or for very low pressures with decreased probability of collision, ionizing breakdown cannot occur. For very small pressure-distance products, the breakdown voltage theoretically even increases after having passed a theoretical minimum, as illustrated in Figure 17(a). In dry air, the minimum voltage at which a breakdown is possible is 327 V, corresponding to a field strength of  $66 \text{ V}\cdot\mu\text{m}^{-1}$  at atmospheric pressure. This minimum was confirmed

<sup>17</sup>It is assumed that some companies are currently working on PZT-based switches [75, Section 5.22].



**Figure 17.** The Paschen curve: (a) theoretical breakdown voltage for dry air, nitrogen and hydrogen against the pressure-distance product [165]; (b) measured breakdown voltage on the left side of the Paschen minimum in dry air [75, 166].

by measurements reporting on a breakdown voltage of  $64 \text{ V} \cdot \mu\text{m}^{-1}$  [167]. In further experiments, the predicted theoretical increase of the breakdown voltage for lower pressures/dimensions on the left side of the minimum could not be verified, as shown in Figure 17(b). This can be explained by extremely high local field strengths due to surface roughness which plays an increasing role for very low gap distances [166]. Thus, the breakdown occurs by field emission depending on the surface properties. Irregular and unstable discharging phenomena can also occur at larger distances if the field strength exceeds  $28 \text{ V} \cdot \mu\text{m}^{-1}$ , without leading to a breakdown [167]. Discharging does not destroy the electrodes but, in the long term, a stable glow discharge degrades them and should be avoided in normal operation modes.

From a practical point of view, a MEMS in-line switch with deposited metal contacts provides reliable isolation for 300 V at gap distances of 10–20  $\mu\text{m}$  and about 150 V for contact distances of 2–4  $\mu\text{m}$ .

## RF isolation

The RF isolation of an electrical switch is defined by the ratio of the output voltage to the input voltage in the off-state, which basically is the so-called  $S_{21}$ -parameter for a passive two-port network<sup>18</sup>. Since this ratio can vary over many orders of magnitude, a logarithmic scale is commonly used.

One might assume that the isolation in the microwave band is mainly determined by the coupling capacitance between the overlapping switching contact areas. When comparing the measured isolation with the calculated value from the capacitance between the contacts, it can be seen that the effective coupling capacitance must be much larger. Table 8 compares the calculated contact capacitances to the effective coupling capacitances reported in the literature for a few selected switches. For "standard" membrane and cantilever switch designs, the measured coupling capacitance by far even exceeds the single contact capacitance. The reason for this discrepancy is

<sup>18</sup>For a reversible network, the  $S$ -parameter matrix is symmetric and  $S_{21}$  is equal to  $S_{12}$ .

**Table 8.** Comparison of the overlapping contact area, contact distance, off-state capacitance and RF isolation of different metal-contact series switches (all data, unless calculated, from the given literature references and [75, Chapter 5]).

Company, institute	Reported in the literature				Calculated	
	Overlapping contact area	Contact distance	Total off- capacitance	Isolation (4 GHz)	Parallel plate capacitance <sup>a</sup>	Ratio
	$A$ $\mu\text{m}^2$	$d$ $\mu\text{m}$	$C_{off}$ fF	$S_{21}$ dB	$C_A$ fF	$\frac{C_{off}}{C_A}$
Rockwell [104]	150	2–2.5	1.75–2	50	0.5–0.7	$\approx 3$
Motorola [99]	300	2–3	2	44	0.9–1.3	$\approx 1.8$
HRL [98]	400	1.5–2	3	45	1.8–2.4	$\approx 1.5$
Analog Devices <sup>b</sup> [106]	10	0.6–1	4	40	0.1–0.2	$\approx 25$
MIT-Lincoln Labs <sup>b</sup> [105]	360	2–15	4–6	40	0.2–1.6	4–20
OMRON [158]	500 <sup>c</sup>	3	5	35	1.5	3.3
Univ. of Michigan [159]	1200 <sup>c</sup>	1.5–2	6–8	35	5–7	$\approx 1.2$
KTH <sup>d</sup>	3500	14.2	4.2	40	2.2	1.9

<sup>a</sup>for one contact only; calculated from  $A$  and  $d$ ; fringe capacitances not taken into account; no dielectric layer

<sup>b</sup>a so-called in-line switch with a single interruption in the signal line; thus, the total off-state capacitance consists of  $C_A + C_f + C_p$

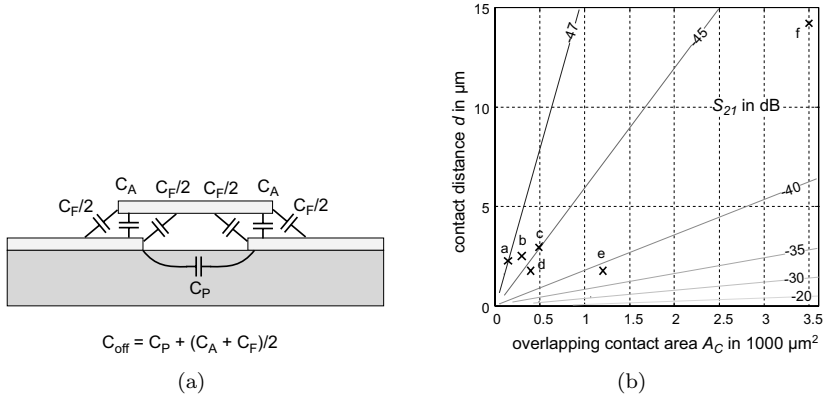
<sup>c</sup>estimation from pictures

<sup>d</sup>the switch presented in this thesis

that the total coupling capacitance is composed of

$$C_{off} = \frac{1}{2}(C_A + C_f) + C_p \quad (4)$$

with  $C_A$  the parallel-plate capacitance of the open contacts,  $C_f$  their fringe-field capacitance, and  $C_p$  the parasitic capacitance between the open ends of the transmission line, as shown in Figure 18(a). The fringe capacitance is proportional to the ratio of the contact distance to the overlapping contact length, and can even exceed the parallel-plate capacitance itself. The coupling capacitance between the open ends of the transmission line mainly depends on the gap between the two signal lines and on the signal line width. For a typical switch design on silicon, this coupling capacitance is 2–4 fF, and it is about 50% lower for GaAs-substrate based switches [75, Section 4.9]. The contour plot in Figure 18(b) is an attempt to estimate the anticipated isolation from the overlapping contact area and the off-state contact distance at 4 GHz. The isolation, calculated for a 50  $\Omega$  system, is based on the approximation  $C_{off} \approx \frac{1}{2}C_A(1 + f_{fringe}) + C_p$ , assuming a fringe-factor  $f_{fringe}$  of 100% and a transmission line capacitance of 3 pF. These parameters were derived by finding the best matching parameters based on the design and measurement data of the Rockwell, Motorola, OMRON, HRL, University of Michigan and KTH switches from



**Figure 18.** (a) Simplified off-state model of a series switch with contact area capacitance  $C_A$ , fringe capacitance  $C_f$ , and parasitic substrate coupling capacitance  $C_p$ ; (b) contour plot of the approximate RF isolation at 4 GHz, depending on the contact distance and on the overlapping contact area, calculated as described in the text, with marks corresponding to some of the switches from Table 8: a) Rockwell, b) Motorola, c) OMRON, d) HRL, e) Univ. of Michigan, and f) KTH switch.

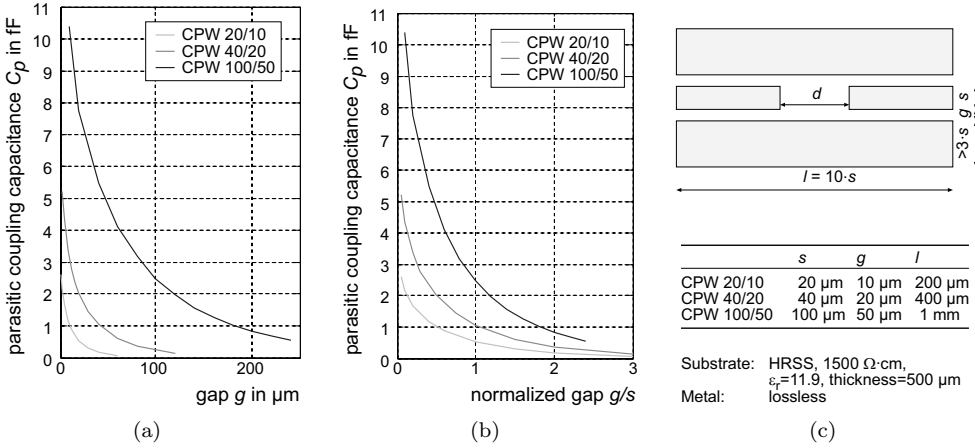
Table 8. The Analog Devices switch was not considered for the curve-fitting, since it deviates from the other switch designs by its small contact area and contact distance, resulting in extremely high parasitic capacitances compared to the mere contact capacitance, as calculated in the table. The Lincoln Labs switch was not taken into account either, due to the reported large variation of the contact distance.

The simulation data presented in Figure 19 tries to verify the estimated parasitic coupling capacitance for the transmission line. The effective capacitance between the two ends of the signal line is extracted from simulation results of the RF isolation, depending on the line width and the gap between the open ends<sup>19</sup>. One is tempted to assume that different design geometries could simply be characterized by the ratio of the gap distance to the signal line width, which is not the case, as shown in Figure 19(b). The simulation was done with the high-frequency planar solver Sonnet 9.51, and the estimated capacitances are slightly lower than the results of comparable simulations reported earlier in [75, Section 4.9]. The thickness of the metal lines is not taken into account in this discussion since it plays only a minor role because it is much smaller than the lateral dimensions of the transmission line.

<sup>19</sup>The parasitic coupling capacitance is extracted from  $S_{21}$  by using the following correlation:

$$Z_p = 2 \cdot Z_0 \cdot \frac{1 - \hat{S}_{21}}{\hat{S}_{21}} \quad \text{and} \quad C_p = -\frac{1}{\omega \cdot \text{Im}(Z_p)}$$

$\hat{S}_{21}$  is the insertion loss of the interrupted line normalized by the insertion loss  $S_{21}$  of the through line,  $Z_0$  the characteristic impedance of the transmission line, and  $\omega$  the angular frequency. Even though the insertion loss of the through line is very small compared to the open line, the normalization is necessary to eliminate the phase shift. The extraction of the coupling capacitance makes sense only if the complex coupling impedance is dominated by a capacitive component, i.e.  $\text{Im}(Z_p) < 0$  and  $|\text{Im}(Z_p)| \gg \text{Re}(Z_p)$ , and, if the calculated  $C_p$  is nearly constant over the investigated frequency band. Otherwise this simple equivalent circuit model would not be appropriate.



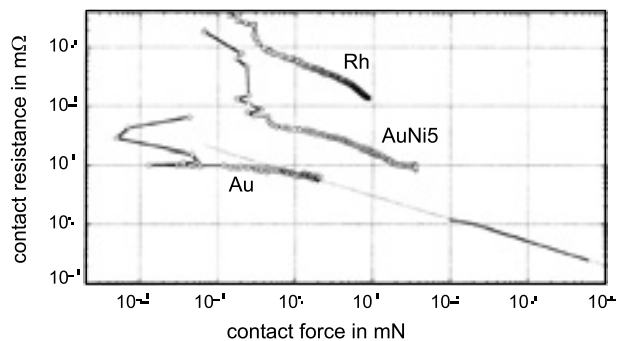
**Figure 19.** Simulated parasitic coupling capacitance against line separation for different interrupted CPW geometries on a high resistivity silicon substrate: (a)  $C_p$  against gap  $g$ ; (b)  $C_p$  against normalized gap  $g/s$ ; (c) geometry of the simulation model.

### 2.2.3 Metal contacts and their reliability

The metal contacts are the most crucial part of a MEMS switch since they determine the on-state resistance and the current handling capability, and are the source of the most dominant failure mechanisms in the type of switches under discussion. Besides literature addressing metal contact issues in conventional relays [168–176], a variety of publications report on the contact physics, contact forces and contact reliability in MEMS switches and relays [89, 156, 167, 177–182]. The main issues determining contact performance in microswitches are discussed in the following paragraphs.

#### Contact force, resistance, materials, and effective contact area

The contact force of typical MEMS switches is between 10  $\mu\text{N}$  and 10 mN, compared to conventional relays with forces from 100 mN upward [89]. The dependence of the contact resistance on the contact pressure has been thoroughly investigated for different contact materials [89, 156, 167, 177–183]. According to the literature, a stable contact resistance of 80–200 m $\Omega$  can be achieved at about 100  $\mu\text{N}$  for gold, which is the most common contact material, dropping to 10–20 m $\Omega$  at 1 mN [177]. For AuNi5, a force of at least 300  $\mu\text{N}$  is necessary for a stable contact resistance below 100 m $\Omega$  [156]. Rhodium, another but less common choice, gives a stable contact resistance of about 1  $\Omega$  at a minimum force of 600  $\mu\text{N}$ , as displayed in Figure 20 [179]. Other well-suited materials are sputtered or plated rhenium and gold-palladium alloys with a contact resistance of 0.5–3  $\Omega$  for 0.2–2 mN contact force [75, Section 7.3], or hard-gold AuCuCd with a stable contact resistance of 50–80 m $\Omega$  at a contact force of 200  $\mu\text{N}$  [178]. Platinum was also used for contact material alloys in a MEMS switch design, but it was not commented on the composition and performance [105]. The contact resistance of gold is about 10 times lower than that of AuNi alloys,



**Figure 20.** Contact resistance versus contact force for gold, rhodium and 5% alloyed nickel in gold [179].

which again is lower than that of rhodium by a factor of 10. Besides the different resistivities, this can be explained by the hardness of the materials, resulting in a larger effective contact area for a low hardness at a given contact force, due to elastic deformation of the metal. The effective contact area is then one of the main factors influencing the contact resistance. A larger contact area results in a smaller resistance, as expected. Smaller areas are also characterized by non-linear behavior, i.e. the resistance increases with increasing current, which can be explained by local annealing effects caused by increased localized heating due to the smaller thermal conductivity of smaller contact areas. In general, the better heat dissipation of larger effective contact areas is *the* critical design criterion to maintain low contact resistance, high power handling capability, and a minimum of surface adhesion wear [177].

Materials with a native oxide such as aluminum, copper, nickel or silicon, are not the first choice since they require a substantial contact force for an acceptable contact resistance. Due to its hardness, silicon, even though very suitable since it often is already used as the structural material of the device, results in very high contact resistances of the order of 20 kΩ [101].

In all, gold and gold alloys in a hermetic atmosphere (nitride or air at atmospheric pressure) are still the contact materials of first choice for microrelays because of their low hardness and low resistivity, both resulting in a low contact resistance, their relatively high melting temperature for a soft material, and their resistance to absorption of surface contaminants [75, 181]. However, the final choice of the contact material also depends on the specific application.

### Contact contamination

Clean contacts are extremely important for microrelays compared to conventional relays since the very low contact forces are not able to break absorbed contaminant films. Sample preparation with tetrachloroethylene followed by alcohol and deionized water was found to be a sufficient cleaning method [177]. Another successful tested procedure is cleaning in isopropanol with 40 subsequent switching cycles at 40 mA,

called "Schaltreinigung" [156]. In both cases, the samples were afterward kept and tested in a nitrogen atmosphere. Besides residues left over from fabrication steps with organic materials, surface contamination can also occur by absorption or condensation of gases, either during the packaging procedure or during operation, if the device is not sealed hermetically. From a material point of view, gold is very suitable since it strongly resists surface layer formation [75, Section 7.4].

## **Metal deposition process**

The deposition processes of the contact material affects its contact performance. Sputtered gold is much harder than electroplated gold and therefore less susceptible to surface damage. Also, higher gold deposition temperatures (hot sputtering at 200 °C, for example) result in a dense and compact grain size and show less surface change due to material transfer and annealing at higher currents than gold deposited at lower temperatures [177].

## **Adhesion force and microwelding**

The estimation of the adhesion force is very important to design the necessary restoring force of the switch actuator. The adhesion force is the least understood parameter of metal contacts in microrelays, and strongly deviating values were reported for different switch designs. Besides the contact material, the switching history influences the remaining adhesion force after releasing the actuation mechanism. The adhesion force depends on the hardness of the contact material, and was reported to be between 0.3 and 2.7 mN for Au and between 0 and 0.3 mN for AuNi5 in an experimental setup [156]. For a practical switch design (the Motorola switch discussed in the Sections 2.1.4, 2.2.1 and 2.2.2) the stiction force was found to be about 130  $\mu$ N. The reader might find it interesting to compare these values to the relatively low estimated restoring forces of some MEMS switches listed in Table 7 on page 32.

Heat dissipation in the contacts is a very important parameter since increased contact temperature softens the material, resulting in a larger effective contact area with larger adhesion. At even higher temperatures caused by local current densities, a soft material with low melting point deforms plastically under the contact pressure, and static microwelding might occur, which can be assumed to be the major failure mechanism at larger switching currents [181]. The very sudden decrease in life time with the switched current is demonstrated in Table 9.

The probability of failure due to contact adhesion depends also on the switching history, which is not only characterized by the switching conditions and the number of cycles, but also by the dwell time of the closed contacts. It is well known for relays in the macro-world that contacts have an increasing tendency to stick to each other if the closed contact is maintained for a very long time, especially when applying an electrical load. Unfortunately, no reliability data has been published for MEMS switches about this type of aging with closed contacts.



**Table 9.** Selection of life-time studies of metal contact switches.

Institute, company	Contact description	Life-time test conditions and results
Analog Devices [96]	gold, $R_{ON}=50\text{ m}\Omega$	hot switching, in air: $10^5$ – $10^7$ cycles up to 1 mA, a few hundred cycles at 10 mA; in nitrogen: $10^9$ cycles cold switched, $2\times 10^8$ cycles hot switched at 10 mA (criterion for failure: $R_{ON} > 100\text{ }\Omega$ )
Delphi Labs [131]	copper, $R_{ON}=40$ – $60\text{ m}\Omega$	with arc-suppressing circuit: $1.7\times 10^6$ cycles at 50 mA, $>2000$ cycles at 350 mA; without arc-suppressing circuit: 20 cycles at 350 mA
CP Clare [129]	gold	$3\times 10^7$ cycles at 5 mA, $10^5$ cycles at 10 mA, $>1000$ cycles at 200 mA
Siemens [95]	AuNi5, contact spring	hot-switched: $6\times 10^6$ cycles at 100 mA

### Other contact degrading mechanisms

Besides less harmful discharging phenomena between contacts in close approximation (see Section 2.2.2), sparking is a major contact degrading mechanism in macro-world metal contact relays. Arcs during the switch transition are a discharging problem caused by parasitic inductances which can create a relatively high voltage over the just opened contacts, trying to continue the current flow in the line, according to  $V_L = L \frac{di}{dt}$ , with  $L$  the total parasitic inductance and  $\frac{di}{dt}$  the time dependent change of the current due to the contact separation. The parasitic inductances in microswitches (especially in microswitch measurement setups) as well as the switched currents are usually quite low and probably play a minor roll in the overall contact degradation, also because of the very fast switch transition times. An exception is the switching of inductive loads with MEMS switches [131]. However, sparking was also identified as a failure mechanism in low current MEMS switches, but the underlying mechanisms are not yet definitely identified [181].

Due to the failure mechanisms mentioned above, a metal contact microswitch dies in a short-circuit, to use the language of semiconductor electronics. This fact is a major drawback from an application point of view. Table 9 lists the results of life-time tests of a few MEMS switches, without discussing the actuation mechanisms and the restoring forces. More life-time test results can be found in [75, Section 7.5]. In general, the life-time is given by the number of switching cycles before the device fails due to mechanical (actuator) or electrical problems (contact stiction or a drastic increase of the contact resistance). Unfortunately, the tests are often not carried out under comparable conditions, which sometimes are not even specified:

- **Cold-switching:** Only a small measurement current (100  $\mu\text{A}$  e.g.) is switched. The main failure mechanisms (microwelding, stiction due to softening) do not occur and very high cycle-numbers can be achieved.

- **Switching without signal current during the switch transition:** The switch is exposed to a current load in the on-state of the switch. Before opening the contacts, however, the load is removed. The switching contacts are more susceptible to stiction and to static microwelding at very large currents. Relatively high cycle-numbers are reported.
- **Hot-switching with load during switch transition:** The whole palette of other failure-mechanisms due to contact degradation can occur. The number of switching cycles is relatively low and depends on the switch design, drastically decreasing at signal currents above 10–20 mA.

A variation of hot-switching is the use of an external electronic switch in parallel to the MEMS switch, opening the line after the MEMS switch has opened, and closing it before the MEMS switch [131]. This technique results in a very low on-resistance (the MEMS switch resistance dominates) and provides substantially improved reliability since no real hot-switching is carried out, but the isolation and other performance parameters are determined by the electronic switch.

#### 2.2.4 Adhesion between actuation electrodes

Besides contact stiction as the main failure mechanism of metal-contact switches, as discussed in the previous section, unwanted adhesion can also occur between touching electrodes during operation, which is also a very important failure mechanism of MEMS switches. This type of failure might have different causes in electrostatic actuators, listed in the order of their importance for MEMS switches:

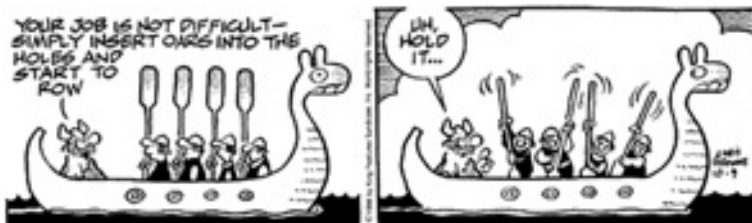
1. **Electrostatic stiction:** Caused by charge injection and charge trapping in dielectric layers, mainly dependent on the field strength, if above  $2 \text{ MV} \cdot \text{cm}^{-1}$  [184], and on the dwelling time in the active state [63].
2. **Hydrogen bridging:** Hydrophilic surfaces such as the native oxides of all silicon based dielectrics (nitrides and oxides) are highly hydrophilic and contain absorbed water layers. When two of these surfaces are brought into close contact, hydrogen bonds may form and result in quite high adhesion energy [161, 185].
3. **Capillary condensation between contacting surfaces:** Liquid condensate can drastically increase the effective contact area of otherwise rough surfaces and can cause stiction [185].
4. **Van der Waals forces between solid bodies:** Caused by mutual electric interaction of the induced dipoles in two touching bodies, strongly dependent on surface roughness [185].

Stiction occurring during processing is not discussed since that causes a yield and not a reliability problem.

Possible means of preventing the mentioned effects are:

- **Hermetic packaging:** Drastically reduces causes 2 and 3, since they depend on the relative humidity of the working environment [161].

- **Increased surface roughness:** Reduces mainly effect 4, which is a main concern only for surface roughnesses below a few nanometers typical of traditional LPCVD based surface micromachining [185, 186].
- **Anti-stiction coating:** Hydrophobic self assembled monolayer (SAM) coatings prevent the formation of hydrated layers and thus effects 2 and 3 [161, 185, 186]. However, anti-stiction coating of the total structure is not possible for metal-contact switches.
- **Distance keepers:** Small separation posts drastically reduce the contact area and thus the probability of stiction for all of the effects described above. As a further significant advantage, isolation layers between electrodes are unnecessary, leading to a so-called all-metal switch. Separation posts are the most efficient method against effect 1 [127, 185]. The slightly more complex fabrication is compensated by the various advantages: no electrostatic stiction, no adhesion problems between dielectric and metal layers, and less stress in the structure (see also Figure 8 on page 20).
- **Conscious choice of materials:**
  - Silicon dioxide has a much lower trap density than silicon nitride and is therefore less susceptible to electrostatic stiction [187], reducing effect 1. However, dioxide takes up more moisture than nitride [188] and might be worse overall if the stiction causes 2 and 3 are not eliminated by other methods such as hermetic packaging. Another rule: LPCVD layers have a lower trap density than PECVD layers [75, Section 7.2].
  - Noble metals without native oxides should preferably be used instead of aluminum or silicon to prevent problem 2.
- **Low electrostatic actuation voltage design:** Charge injection is exponential with voltage and a reduction in the actuation voltage by 6 V results in a 10-fold increase in the lifetime of a capacitive MEMS switch [121], thus, reducing effect 1.
- **Bipolar actuation voltage:** The polarity of the voltage does not affect the electrostatic force. Negative voltage pulses when releasing the actuation voltage reduce trapped charges and result in a vast improvement in the switch reliability [75, Section 7.2]. Charge trapping is also avoided by using AC actuation voltage [161]. Both methods reduce problem 1 at the cost of increased complexity of the electronic drive circuitry.



**Der Panther***Im Jardin des Plantes, Paris*

Sein Blick ist vom Vorübergehn der Stäbe  
 so müd geworden, daß er nichts mehr hält.  
 Ihm ist, als ob es tausend Stäbe gäbe  
 und hinter tausend Stäben keine Welt.

Der weiche Gang geschmeidig starker Schritte,  
 der sich im allerkleinsten Kreise dreht,  
 ist wie ein Tanz von Kraft um eine Mitte,  
 in der betäubt ein großer Wille steht.

Nur manchmal schiebt der Vorhang der Pupille  
 sich lautlos auf—. Dann geht ein Bild hinein,  
 geht durch der Glieder angespannte Stille—  
 und hört im Herzen auf zu sein.

*Rainer Maria Rilke,  
 aus "Neue Gedichte", 1902*

**The Panther***In the Jardin des Plantes, Paris*

His gaze is from the passing of bars  
 so exhausted, that it doesn't hold a thing anymore.  
 For him, it's as if there were thousands of bars  
 and behind the thousands of bars no world.

The sure stride of lithe, powerful steps,  
 that around the smallest of circles turns,  
 is like a dance of pure energy about a center,  
 in which a great will stands numbed.

Only occasionally, without a sound, do the covers  
 of the eyes slide open—. An image rushes in,  
 goes through the tensed silence of the frame—  
 only to vanish, forever, in the heart.

*Rainer Maria Rilke,  
 in "New Poems", 1902  
 (translation by Cliff Crego)*

*Once we are destined to live out our lives in the prison of our mind, our duty is to furnish it well.*

*Peter Ustinov  
1921–2004, British Actor,  
Writer, Director*

## 2.3 MEMS wafer-level packaging

### 2.3.1 MEMS packaging

The main functions of a microsystem package are the following:

- environmental protection: chemicals, temperature, electrical and magnetic fields
- routing of electrical interconnections
- interfacing of sensor and actuator elements with their environment
- hermetic sealing (if required)
- heat dissipation
- mechanical support and mechanical protection
- mechanical stress relaxation
- integration of multi chips into one single package
- standardized package for automated device handling on a system assembly level
- providing test interfaces
- handling during tests after different fabrication steps

Except a few points mentioned above, these requirements are not much different from conventional electronic packaging. However, the packaging is regarded as a very critical block in the successful commercialization of microsystems. The packaging costs contribute with a major factor to the total device fabrication costs, varying from 20% for a simple plastic encapsulated pressure sensor to 95% for special pressure sensors designed to withstand an extremely harsh environment [189, Chapter 1.4]. The main reasons for these high costs, as compared to microelectronics packaging, are the following:

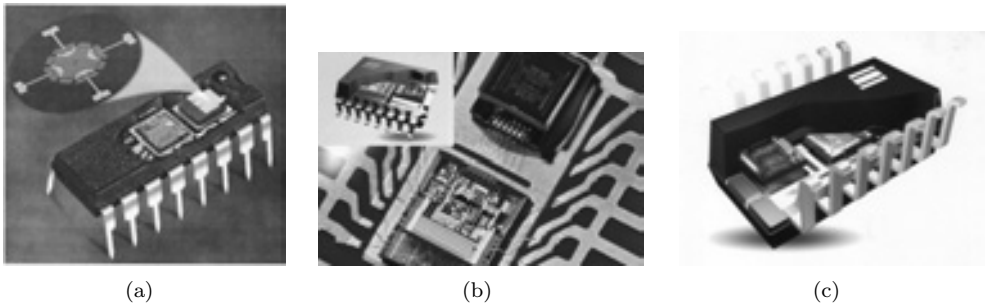
- Microsystems consist of more complex three-dimensional structures.
- Moving mechanical parts often require an encapsulated cavity with a controlled atmosphere (hermetic package).
- Each MEMS design is unique and, compared to electronic devices where the packaging is less dependent on the chip functionality, the package has to be individually developed for each microsystem.

- Standard packaging procedures and processes are often not suitable for MEMS devices.
- Microsystems typically interact with their environment. Thus, the chips are more directly exposed to harsher conditions with higher demands on the overall packaging, and the working media has to be interfaced by the microsystem.
- Especially the fabrication of three-dimensional structures with high aspect ratios results in larger geometrical variations than the typical fabrication of semiconductors. Thus, the package design has to be more tolerant to large product non-uniformity.
- The uniqueness of MEMS devices also requires unique test methods after different fabrication and packaging steps. Special test interfaces have to be provided by the package.
- The reliability of electronic packages is very well investigated and also much better understood because the failure mechanisms are less dependent on the circuit function. In contrast, each type of MEMS device has its own particular failure mechanisms which makes it difficult to provide a suitable and reliable package.
- Up to now, many (low volume) MEMS components are packaged manually with inappropriate tools in an expensive and time consuming process.
- For economical reasons, a MEMS package should be compatible with low-cost printed circuit board (PCB) assembly technology designed for semiconductor chips. However, the complex device functions of many microsystems do not allow a standard package.

The whole packaging procedure of a MEMS device typically consists of different steps. For instance, the automotive sensors shown in Figure 21 are packaged by the following procedure:

1. Encapsulation of the mechanical structure (here: by wafer bonding).
2. Mounting of the MEMS device, together with an integrated circuit chip, on a leadframe flag with standardized size and contact pads for DIP or SOP packages.
3. Wire bonding of the chips to the frame, and between the chips.
4. Passivation of the chips and the bonding wires by a layer of silicone gel.
5. Molding of the frame with the mounted chips in an epoxy compound, giving the final shape of the plastic package.

MEMS packaging, especially for low volume products, is often done completely on chip level. Thus, the wafer for the device fabrication is diced before the packaging procedure, which is carried out individually on each chip. Wafer-level packaging or so-called zero-level packaging are terms used when some parts of the packaging, typically including the sealing, are done before the wafer dicing. These packaging steps



**Figure 21.** Examples of packaged automotive sensors: (a) accelerometer by Motorola in DIP package [193]; (b) gyro sensor by SensoNor in SOP package; (c) tire pressure sensor by SensoNor in SOP package [194].

are carried out on all devices at the same time. In connection with the higher packaging densities enabled by wafer-level packaging techniques, the highly parallel process steps lower the overall packaging costs. Furthermore, the devices are already packaged before the critical dicing step which involves a high risk of contamination and requires the full protection of mechanically moving parts. Typical wafer-level packaging techniques are thin and thick film deposition processes for device sealing [189, 190] and wafer bonding [191, 192].

Figure 21(a) shows an accelerometer by Motorola [193] put into an industry standard 16-pin dual-in-line package (DIP), together with an electronics chip. This approach of integrating the MEMS device with the electronics circuitry in the package and not on a chip-level is called system-on-a-package (SOP). The MEMS-chip is encapsulated hermetically by full-wafer glass-frit bonding, which is, besides anodic bonding, the most common hermetic sealing method by wafer bonding (see Section 2.3.2).

Figure 21(b) shows a gyro sensor fabricated by SensoNor, Horten, Norway, also following the system-in-a-package idea, for a small-outline-package (SOP<sup>20</sup>). The MEMS chip is hermetically sealed by anodic wafer bonding. A very similar fabrication and packaging procedure is utilized for a tire pressure sensor by the same company, shown in Figure 21(c). Besides the pressure sensor, this device also includes a radial acceleration sensor, a temperature sensor and a battery voltage monitoring function [194].

### 2.3.2 Wafer bonding techniques

Wafer bonding is the joining of one substrate to another substrate. Typically, wafers made of the same material or wafers of different materials can be bonded together. The origins of wafer bonding are back in the 1980s and the main commercial application is still in the area of creating silicon-on-insulator (SOI) substrates for RF electronics [192].

However, wafer bonding attracted attention as a sealing and packaging method for bulk-etched cavities containing micromechanical sensor elements. It actually can

<sup>20</sup>Unfortunately, the abbreviation SOP is commonly used both for system-on-a-package and for small-outline-package.

**Table 10.** An overview of the most common wafer bonding techniques.

Bonding method	Temperature °C	Hermeticity	Reliability	Bond strength MPa	Comments
<i>Surface bonding</i>					
anodic bonding	medium, 200–500	yes	good	>30	requires flat surface, high voltages
fusion bonding	very high, 700–1400	yes	good	>20	requires surface preparation, flat surface
surface activated bonding	very low, RT <sup>a</sup> –200	yes	not certain	>10	requires flat surface, surface preparation
<i>Metallic interlayer</i>					
eutectic bonding	medium, 180–400	yes	not certain	20	particle insensitive
thermo-compression bonding	medium, 350–500	not certain	not certain	10	very high pressure required
solder bonding	low, 180–300	yes	not certain		particle insensitive
<i>Insulating interlayer</i>					
adhesive bonding	low, RT–300	no	not certain	10–25	very particle insensitive, very versatile
glass-frit bonding	med./high, >400	yes	good		particle insensitive, versatile

<sup>a</sup>room temperature

be regarded as *the* enabling technique for many high-volume, low-cost MEMS sensors such as the ones shown in Figure 21.

The main applications of wafer bonding in MEMS, besides functional or protective sealing [195–198], are vertical stacking of devices, which is also interesting in semiconductor fabrication for stacking memory chips [199], and combining structures fabricated on two different substrates, if the fabrication of the structures is not compatible or too complex for one single substrate. This technique of merging two wafers with subsequent removal of one of the two substrates is called transfer-bonding [200].

The most common wafer materials used in (RF) MEMS fabrication are silicon with different doping levels and crystal orientations as well as pyrex and quartz glass. Silicon carbide and gallium arsenide are getting increased attention. Ceramic substrates such as aluminum nitride (AlN) and alumina (Al<sub>2</sub>O<sub>3</sub>) are the most commonly used substrates in MEMS chip packaging, but are rarely used directly as a fabrication substrate. On a system level (multichip modules), typical electronic substrates such as low temperature co-fired ceramics (LTCC) are also used in MEMS packaging. Ceramic substrates allow hermetic MEMS packages [201].

Table 10 and the following list give an overview of the wafer bonding techniques used in MEMS fabrication.



*Anodic bonding* [195, 202, 203] provides reliable sealing and is suitable for high vacuum packaging down to a few milliTorr. This technique can be used for different substrate material combinations: glass-to-glass, glass-to-silicon, glass-to-metal, silicon-to-silicon. Accelerometers fabricated in very high volumes are typically sealed by anodic bonding of a pyrex glass to a silicon wafer. Anodic bonding is carried out at relatively low temperatures of 200–400 °C, but involves high voltages of 200–1700 V, which might damage microelectronic circuits. Another threat to electronics is sodium ion diffusion from pyrex glass into the silicon wafer. Anodic bonding is very particle-sensitive, and a surface roughness of less than a few nanometers is required [204]. Another problem of anodic bonding, as well as of the other "surface" bonding techniques, is the difficulty of creating in-plane electrical interconnections through the seal. However, a few methods of creating feedthroughs were reported in the literature:

- Very thin electrical feedthroughs with a thickness in the order of 50 to 100 nm can be manufactured with a special layout design without decreasing the hermeticity of the package [205].
- Interconnection lines with a thickness of up to 500 nm were demonstrated by etching recesses in the glass wafer, which were subsequently filled by metal evaporation. The metal overlaps the corners of the recesses to achieve a seal even on the slopes of the etched glass [206].
- Anodic bonding is possible with polished surfaces. Thus, in-plane interconnection lines can be covered by a polysilicon layer which is polished afterward [207].
- Interconnections can be created by doping resistors which are buried in the substrate and covered with an epitaxially deposited silicon layer, as done for the high volume automotive sensors fabricated by SensoNor, Norway.

*Fusion or direct bonding* [208–210] is also a widely used technique for hermetic packaging. The bonding surfaces have to be extremely flat and clean, and the bonding is primarily accomplished by the induced chemical forces between the surfaces which have to be activated by hydration or by a hydrophilic process. The bonding occurs "spontaneously" at oxidising ambient, and has to be followed by an annealing step [189, Chapter 2.6]. For the annealing, very high temperatures of 700–1400 °C are necessary to yield a durable bond. Direct bonding without annealing step can also be done with polymer materials, typically with PMMA or PDMS, involving a surface treatment in a plasma cleaner but yielding in a less strong bond [211].

*Surface activated bonding (SAB)* [212, 213] is basically also a direct bonding technique requiring very flat surfaces and special surface preparation. Here, the bonding energy originates from a very strong adhesive force acting between two atomically clean solid surfaces under contact [214], while in most of the other bonding techniques the bonding energy is provided by the processing temperature in combination with the imposed bonding pressure. The cleaning procedure typically consists of a dry etching process such as ion beam bombardment or radical irradiation, carried out in ultrahigh vacuum. Bonding by the SAB method is completed even at room

temperature and can be applied to a variety of different materials (Al, Cu, Ag, Au, Sn, Ti, Ni, SiC, Si<sub>3</sub>N<sub>4</sub>, Al<sub>2</sub>O<sub>3</sub>, AlN, diamond and silicon [213]).

*Eutectic bonding* [215,216] is diffusion based joining of two metals. The bonding occurs at the so-called eutectic point, which is the lowest fusion temperature of an alloy with a melting point lower than that of any other combination of the same components, and much lower than the melting points of the involved metals. At temperatures above this eutectic point, rapid diffusion of one material into the other occurs, which results in the formation of an eutectic alloy. Eutectic bonding is assisted by pressurizing the surfaces. Typical material combinations are Au 97%-Si 2.83% with an eutectic temperature of 363 °C (the melting temperatures of silicon and gold are 1410 °C and 1064.4 °C, respectively), or solder forming of Sn 62% with Pb 38% at 183 °C (melting temperatures: Sn: 231.9 °C, Pb: 327.4 °C). Silicon and gold are the preferred materials in MEMS because of processing simplicity and compatibility, but it is difficult to obtain high-quality sealing at temperatures below 400 °C [213].

*Glass-frit bonding* [217], also called low-temperature melting glass bonding, is an adhesive bonding method with an inorganic intermediate bonding layer. The glass-frit is deposited on one or both of the wafers. Then the wafers are brought into intimate contact and heated. The glass-frit reflows and forms a rather strong bond, suitable for hermetic packaging. Typical glass curing temperatures are between 400 and 1100 °C. This method is very insensitive to particles and can be applied on various wafer materials. It results in a hermetic bond but is not suitable for sealing high-vacuum cavities, because the glass-frit releases trapped gases.

*Solder bonding* [218,219] is similar to eutectic bonding. Here, the metal solder is applied as a compound material to the bonding surfaces. The surfaces are brought into contact and heated to the melting temperature of the solder. The solder reflows which causes intimate contact between the two parts, leading to a permanent and hermetic bond. Fluxes in the solder, used to remove oxides from the metal surfaces, should be avoided because they evaporate during the heating and might condense on critical mechanical parts inside the package. Similar to eutectic bonding, solder bonding is less sensitive to particles than anodic or fusion bonding. To create higher temperatures of up to 800 °C in the bonding interface without harming the structures to be packaged, localized heating by embedded resistors was found to be suitable for MEMS sealing, both for solder and for eutectic bonding [220].

*Thermocompression bonding* [221–223], often referred to as solid-state welding, is a bonding method involving a very high pressure between the two surfaces to be bonded. The pressure causes plastic deformation, which results in intimate contact between the opposing surfaces. This process is often assisted by heat, either by "global" heating [221], ultrasonic energy [224], or localized microwave heating [225]. Common material combinations are gold-to-gold, aluminum-to-gold, aluminum-to-aluminum and aluminum-to-glass. Typical bonding pressures are about 300 MPa, and temperatures of 350–500 °C are necessary to achieve an acceptable result.

In general, data in the literature about the bonding pressure has to be read very carefully whenever structured wafers are bonded. For practical reasons, the bonding pressure is typically referred to as the applied pressure between the two chucks in the bonding tool. However, the effective pressure  $p_{eff}$  between the touching structures, which determines the conditions for forming the bond, depends not only on the chuck pressure  $p_{chuck}$  but also on the ratio of the total chuck area  $A_C$  to the effective contact area  $A_{eff}$ :

$$p_{eff} = p_{chuck} \frac{A_C}{A_{eff}} \quad (5)$$

Keeping this in mind explains the different values mentioned in the literature, varying by a few orders of magnitude. For instance, the necessary bonding pressure for gold thermocompression bonding is stated to be about 300 MPa in [222], but a chuck pressure of only 4 MPa achieved very good results for the same process in [223]. The ratio of about 100 between these values can be explained by the mask layout used in [223], consisting of ring-shaped structures intended to hermetically seal cavities for MEMS devices.

### 2.3.3 Polymer adhesive wafer bonding

Wafer bonding with a polymer intermediate bonding layer [222, 226–228] is increasingly attracting attention because of its many interesting advantages:

- low bonding temperature from room temperature up to 300 °C
- simple, low-cost processing (spin-on polymers)
- high bond strength
- highly tolerant to particles and surface non-uniformities of several  $\mu\text{m}$  due to the good planarization and wetting properties of most adhesives, which even enable thick electrical interconnection lines to be embedded in the adhesive layer
- many different substrate materials can be joined
- compatibility of many polymers with standard clean-room processing
- the bonding layer acts as a stress buffer between the two substrates due to the elastic properties of polymers
- the possibility of patterning the adhesive layer before the bonding procedure, to create three-dimensional structures and to get cavities for housing MEMS devices directly in the bonding layer
- RF interconnection lines in or through the adhesive layer have very low signal attenuation and low reflections since adhesives with low loss tangent and low dielectric constants are available
- many polymers with different properties (thermo-set, thermo-plastic, curing temperature, viscosity, etc.) are available on the market, making this technology very versatile and suitable for various applications

The main disadvantages of polymer bonding for MEMS packaging should also be revealed:

- Polymers are permeable to gases to a much higher degree than metals or ceramics [190, Chapter 14].
- The long-term stability of polymers is uncertain.
- Most polymers have limited temperature stability.
- Polymers, especially when not pre-cured, are very soft which results in larger post-bonding alignment inaccuracies because of unavoidable sheer forces between the bond chucks. For adhesive bonding, the alignment uncertainty is of the order of tens of micrometers [222], as compared to anodic bonding with a post-bonding alignment accuracy of less than  $2\text{ }\mu\text{m}$ .

Typical materials used for adhesive bonding in MEMS are epoxy-based polymers such as benzocyclobutene (BCB [229,230]) and SU-8 [231,232], epoxy-based photore-sists [229], polyimides [233] and B-stage epoxies [234].

Epoxy-based polymers, such as BCB, and the so-called liquid crystal polymers (LCP) provide a certain gas-tightness, and packages involving such polymers are often classified as "near-hermetic" packages. BCB for bonding is discussed in Section 4 and in papers 1, 2, 5 and 6, and LCP, a relatively new material in MEMS which might play a larger role in adhesive bonding in the future, is discussed in the following paragraph.

LCP are thermoplastic polymers made of aligned molecule chains with crystal-like spatial regularity, exhibiting unique electrical, physical and chemical properties [235]. Table 11 compares the properties of LCP with kapton and BCB. For oxygen, carbon dioxide, nitrogen, argon, hydrogen and helium, LCP have above-average barrier performance and the permeation of gases through LCP is not affected by humidity, not even at elevated temperatures. Also, for being a polymer, LCP films have an extremely low moisture uptake and show excellent chemical resistance [236]. From a mechanical point of view, it is remarkable that the coefficient of thermal expansion (CTE) can be controlled in a predictable way during the processing, even reaching very low values [237]. LCP can be etched almost isotropically in oxygen plasma without additional reactive ions, with an etch rate of about  $0.25\text{ }\mu\text{m}\cdot\text{min}^{-1}$  at a power of 350 W [235]. To the knowledge of the author, LCP are currently only available as sheet material and not in liquid form and are therefore not suitable for spin-coating, since they get liquid only above the melting temperature of about  $280\text{ }^{\circ}\text{C}$ .

### **2.3.4 Hermetic sealing on wafer-level**

MEMS hermetic packaging on wafer-level can be classified into two categories: 1) surface-micromachined micro-shells and 2) bonding methods, either chip to wafer or wafer to wafer. Full-wafer bonding techniques, still the most common hermetic packaging techniques in high-volume MEMS production, were already discussed in Section 2.3.2. For chip to wafer capping, the same bonding physics is used as for full-wafer bonding, but each device is encapsulated individually by flip-chip-like pick-and-place techniques. Eutectic bonding and solder reflow sealing [238] are the most commonly used

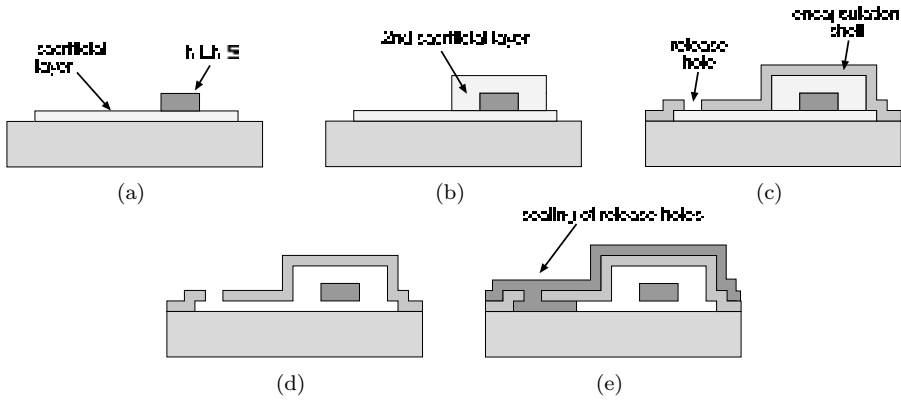
**Table 11.** Comparison of the properties of a typical liquid crystal polymer (LCP) with kapton and BCB [235].

Property	Unit	LCP	Kapton	BCB
manufacturer		Celanese/Ticona	DuPont	Dow Chemical
product specification		Vectra A-950	HN200	Cyclotene 3022
melting temperature	°C	280	400	N/A <sup>a</sup>
dielectric constant	–	2.8	3.5	2.65
loss factor, $\tan \delta$	–	$\approx 0.004$	0.002	0.0008–0.002
moisture absorption	–	$< 0.02\%$	2.8%	$< 0.12\%$
coef. of therm. expan.	ppm K <sup>-1</sup>	0–30 <sup>b</sup>	20	52
tensile strength	MPa	200	234	85
Young's modulus	GPa	9	2.5	2.0
specific gravity	kg·dm <sup>-3</sup>	1.4	1.42	1.05

<sup>a</sup>since a thermoset, and not a thermoplastic material<sup>b</sup>can be determined during the fabrication process

chip sealing techniques. Typically, hermetic packaging by full-wafer bonding as well as chip-to-wafer bonding is quite space-consuming since the sealing rings require a width of a few hundred micrometers to create an effective gas barrier. Also, the total package height is of the order of the thickness of two wafers. However, atmosphere control is better for anodic and fusion bonding than for the surface-micromachining techniques discussed in the following paragraph.

Surface-micromachining leads to very compact and thin micro-shells requiring just a bit more space than the actual devices to be packaged. The traditional sealing procedure by LPCVD processes is illustrated in Figure 22. An additional sacrificial layer (typically SiO<sub>2</sub>) is deposited onto the MEMS structures and then covered by the encapsulating shell with narrow etch holes. Both the functional and the package sacrificial layer are removed in one step (typically HF wet or vapor etching) and the release holes are closed afterward by depositing a sealing layer which penetrates into the etch holes and etch channels, but not into the encapsulated cavity [239–241]. Instead of release holes, a HF permeable polysilicon layer can be used, which afterward is covered by the sealing layer [242, 243]. The disadvantage of LPCVD surface micromachining is the high deposition temperatures, typically above 700 °C, making this sealing technique basically not suitable for pre-fabricated electronic circuitry. However, it was shown that it is possible to combine this sealing technique with a BiCMOS process, when the CMOS structures are fabricated first without metal interconnections, then the MEMS devices including the sealing procedure, and finally the CMOS back-end metal layers [244]. Besides LPCVD, metal evaporation [245] and reflow of solder in vacuum [246] were successfully used to seal LPCVD fabricated micro-shells. Also, an approach using only process steps with temperatures below 250 °C for both the micro-shell and the sealing fabrication was presented recently. Here, the encapsulation is formed by electroplating nickel onto a photoresist sacrificial layer which is etched afterward by a KOH-based photoresist remover. The etch



**Figure 22.** Schematic process flow of fabrication integrated encapsulation by surface micromachining techniques.

channels are finally sealed by reflowing lead-free and fluxless solder bumps, printed onto the chips, in vacuum [247].

An alternative to hermetic packaging on wafer-level is hermetic ceramic or metal packaging on chip level. Drawbacks of this approach are the high cost of ceramic and metal can packages and the problematic of exposing fragile micromechanical parts to contamination during the wafer dicing, the cleaning and the chip handling [248]. Thus, wafer-level encapsulation is probably the only feasible approach for a high-volume low-cost MEMS product.

If the package should keep very low pressures, the use of so-called getter materials might be considered, which absorb free gases inside the package. Such gases come from outgassing of the involved materials (especially glass), mainly if the bonding is done at elevated temperatures. Pre-heating of the components before the sealing is a very effective method to minimize material outgassing, but it is nevertheless insufficient for cavity pressures below  $10\ \mu\text{bar}$ . Typical getter materials consist of metal alloys containing Ba, Ti, Fe or Al [249, 250]. Pressures lower than  $10^{-8}$  bar were reported using so-called non-evaporable getters [251]. However, using a getter implies delicate processing steps in the fabrication. Thus, getter materials are seldom used in MEMS [252]. Getter materials are more frequently used in GaAs semiconductor circuits to absorb hydrogen and moisture, preventing so-called hydrogen poisoning. Here, typical getters consist of an active oxide and a desiccant in a flexible silicon polymer matrix. The active oxide reacts with hydrogen, converting it to  $\text{H}_2\text{O}$ , and the desiccant absorbs the moisture, thus permanently removing the harmful gases [253].

### 2.3.5 Wafer-level packaging of RF MEMS switches

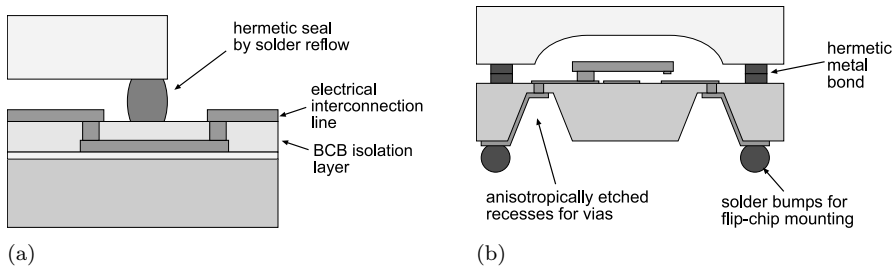
Packaging problems are thought to be one of the main reasons why RF MEMS switches are still not available on a commercial basis [7, 53]. Besides the general MEMS packaging problems, as discussed in Section 2.3.1, the packaging of RF MEMS switches is additionally confronted with the following difficulties:

- Electrical interconnection lines, penetrating through the (hermetic) package, should have low signal attenuation and low reflections. Thus, metal sealing (eutectic or solder bonding) with isolated signal lines through or underneath the sealing ring, might not be appropriate.
- Hermetic packaging is required for reliability and performance stability, especially for metal contact switches, to avoid surface absorption of contaminants and moisture uptake in isolation layers, causing stiction.
- Especially when targeting wireless applications, the chip size is very important. Thus, wafer bonding as used for many high-volume hermetic MEMS packages might not be possible without additional thinning of at least one of the substrates.
- All involved materials, including substrates, passivation layers and mounting frames, should have good RF properties.

The following paragraphs give a few examples of RF MEMS switch packages, with a special focus on RF interconnection solutions.

The typical chip-level hermetic packaging method by soldering and wire-bonding a MEMS device into a low temperature co-fired ceramic package (LTCC) and subsequent sealing with a soldered metal lid can be adapted for a RF MEMS device when optimizing the RF signal paths. Flip-chip bonding of the MEMS device to the ceramic carrier, in combination with a pin grid array (PGA) package has proved minimized signal attenuation [254].

In 2001, researchers at IMEC, Leuven, Belgium, reported on a wafer-level package for RF MEMS consisting of a glass sealing cap which is flip-chip bonded by indent solder reflow sealing onto the RF substrate containing the MEMS switches. RF feedthroughs, embedded in a BCB polymer layer below the metal sealing ring, were characterized in their RF performance up to 50 GHz [66]. The BCB layer has also the advantage that it can be used as an additional thick-film isolation layer to the substrate which decreases the substrate losses if the RF transmission lines are mainly placed on top of it. This packaging concept is shown in Figure 23(a). Also, hermeticity tests were carried out and the package was found to pass the helium leak



**Figure 23.** Hermetic packaging concepts with metal sealing rings: (a) transmission lines embedded in a thick-film isolation layer [66]; (b) RF transmission lines penetrate the wafer and not the metal seal to avoid signal attenuation [67].

test specifications by the MIL-STD 883E procedure. As the authors state in the paper and investigate in a subsequent publication [255], the leak test results have to be interpreted carefully since the packaged cavities were very small and the packaging method cannot be considered as hermetic, since it involves polymers (compare Section 4.5). However, a ceramic isolation material such as  $\text{SiO}_2$  or  $\text{SiN}$  could be used instead of BCB to improve the hermeticity, but then planarization problems have to be faced.

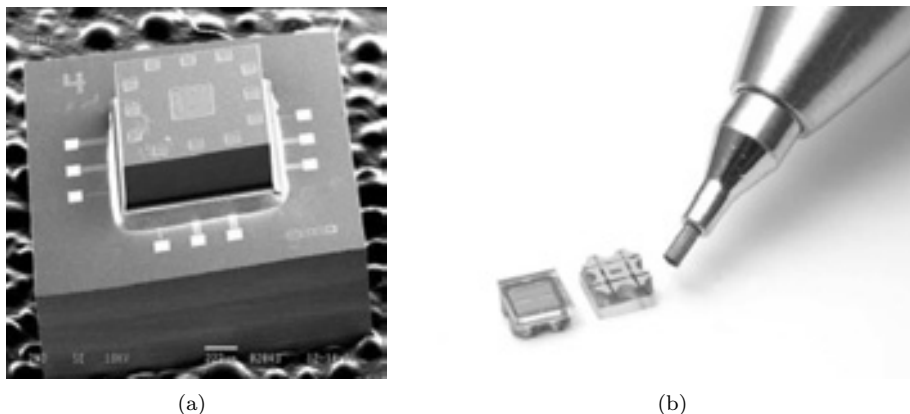
Metal sealing rings can be used even for RF applications if the RF interconnection lines do not penetrate through the sealing ring, but vertically through the wafer. Such a technique, combining a back-side anisotropic wet-etch in KOH or TMAH with metal deposition and patterning of the electrical interconnection lines by electroplated photoresist, was presented by the University of Michigan, Ann Arbor, USA, in 2001, and the RF attenuation was found to be very low [67]. Figure 23(b) shows the principle of the packaging concept. This approach, with modified fabrication sequences, was used in some following publications [256], and the idea of deposited metal lines on sloped, isotropically etched side-walls was also picked up to propose a general multilevel interconnection and packaging concept [257].

In silicon substrates, through-wafer vias allowing metal seals can also be fabricated by other techniques, such as

- photo-assisted electro-chemical etching of the deep holes in silicon and filling with Au-20wt%Sn solder by the so-called molten metal suctioned method [258]
- deep reactive ion etching (DRIE) of silicon using the Bosch process with subsequent oxidation of the walls for isolation purpose, deposition of doped polysilicon as the conducting material, and filling of the holes with low temperature oxide (LTO) and another polysilicon layer [259]
- DRIE of silicon, with a silicon nitride barrier and electroplated copper vias [260]. This technique was also used by the same authors to create a Faraday cage suppressing crosstalk [261]. A probably similar fabrication procedure with additional solder bumps was developed for a RF MEMS package [262]. DRIE in combination with electroplated copper on the sidewalls was already shown by researchers at Stanford University, CA, USA, in 1999, but without sealing the holes [263]. Electroplating of gold in DRIE vias with subsequent plating of solder bumps was used for a RF MEMS packaging procedure on ultra-thin silicon substrates of 50  $\mu\text{m}$  thickness. Here, the package was hermetically sealed by Au/Sn-Au metal bonding [264].

A very interesting device-scale micropackage was demonstrated by the National Taiwan University, Taipei, Taiwan, in 2003. A microshell is formed by electroplating nickel on an aluminum sacrificial layer in etched recesses in a glass wafer. After etching the sacrificial layer, the microcap is freed and suspended only through long tethers attached to the carrier glass, which are broken after the individual flip-chip transfer by solder bonding of the caps to a substrate wafer. This microshell fabrication and transfer technique was demonstrated to package a tunable capacitor successfully, where the electrically conducting microcap also acted as the top electrode for the three-plate tunable capacitor structure [265].





**Figure 24.** Wafer-level packaging concepts of RF MEMS switches proposed by companies: (a) Radant MEMS switch packaged with full-wafer glass-frit bonding; (b) OMRON RF switch also using glass-frit wafer bonding, with metalized through-wafer vias.

A glass-frit wafer bonding process was used by Radant MEMS to hermetically package their RF MEMS switches. Figure 24(a) shows a single packaged Radant MEMS switch [266]. Also OMRON packaged their RF MEMS switch by glass-frit wafer bonding, and interconnections to the back-side of the wafer were created by metalization of through-wafer holes, as shown in Figure 24(b). The total package is only  $1.8 \times 1.8 \times 1.0 \text{ mm}^3$  in size, and the cavity of the device is sealed with inert nitrogen gas [158].

The electrothermally actuated microrelay by Tohoku University, Sendai, Japan, is hermetically packaged by anodic bonding of a pyrex glass wafer to a SOI substrate. The electrical feedthroughs are created by DRIE of vias through the pyrex wafer, which are filled by plated nickel. Furthermore, the electrical contacts between the top and the bottom wafer are established during the bonding procedure [130].

Sand-blasted holes, subsequently filled with electroplated gold plugs, were used by the Korean Institute of Science and Technology, Seoul, Korea, in 2002, to create through-wafer vias in a  $300 \mu\text{m}$  thick pyrex glass substrate. The glass wafer containing etched cavities was bonded to a quartz substrate with B-stage epoxy to create near-hermetic packages for RF MEMS devices [234].



*Man is the only animal for whom his own existence  
is a problem which he has to solve.*

*Erich Fromm*  
1900–1980, German Social Psychologist

*The earth teaches us more about ourselves than all  
the books in the world, because it is resistant to  
us. Self-discovery comes when man measures him-  
self against an obstacle.*

*Antoine de Saint-Exupéry*  
1900–1944, French Aviator, Writer,  
in "Wind, Sand and Stars"

*The only journey is the one within.*

*Rainer Maria Rilke*  
1875–1926, Austrian Poet

*The human mind is like a drunken monkey; it wants  
to jump from one place to another, seeking externally  
the solace and peace that are already within.*

*Sri Swami Rama*  
1925–1996, Indian Sage

*Only in quiet waters things mirror themselves undis-  
torted. Only in a quiet mind is adequate perception  
of the world.*

*Hans Margolius*  
1902–1984, German Writer



*Creativity requires the courage to let go of certainties.*

*Erich Fromm  
1900–1980, Social Psychologist*

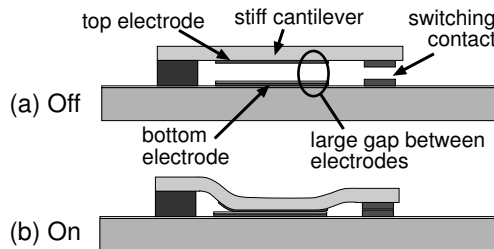
### 3 The S-shaped film actuator based switch

This section is intended to introduce the S-shaped film actuator based switch and to discuss the advantages and disadvantages of the novel concept. It also presents the prototype design and the characterization results of the first fabricated devices. More detailed information about the switch, its fabrication and the measurements can be found in the journal papers 3 and 4.

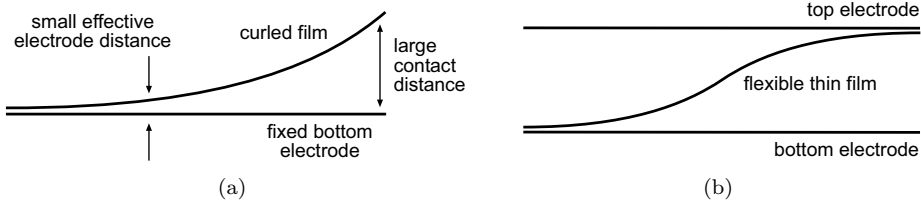
#### 3.1 The switch concept

The conventional and so far most common concept of an electrostatically actuated switch is based on a cantilever-spring system, as shown in Figure 25. This concept evolves a few problems, as already raised in Section 2.2.1:

- A large contact distance in the off-state involves also a large distance between the actuation electrodes, which requires a high actuation voltage.
- The restoring force to open the switching contacts is supplied by the spring energy stored in the deformed cantilever. A large restoring force needed to overcome the adhesion force between the contacts requires a rather stiff structure which increases the actuation voltage.
- The contact force determining the contact resistance (see Section 2.2.3) is the difference between the electrostatic force and the restoring force. Thus, a large restoring force requires an even stronger electrostatic actuator, which demands an increased electrode area or a higher actuation voltage.
- A small electrode distance in the off-state, keeping the actuation voltage and the electrode size at an acceptable level, also results in a small contact distance. That leads to a large coupling capacitance between the open contacts, resulting in poor RF isolation, especially for large overlapping contact areas.

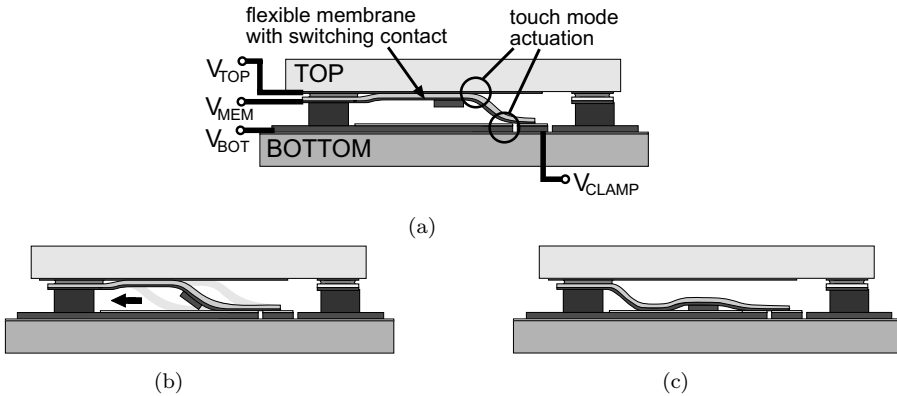


**Figure 25.** Illustration of the conventional concept of an electrostatically actuated switch.



**Figure 26.** Evolution from the curled actuator (a) to the S-shaped film actuator (b), endowed with an additional counter-actuator.

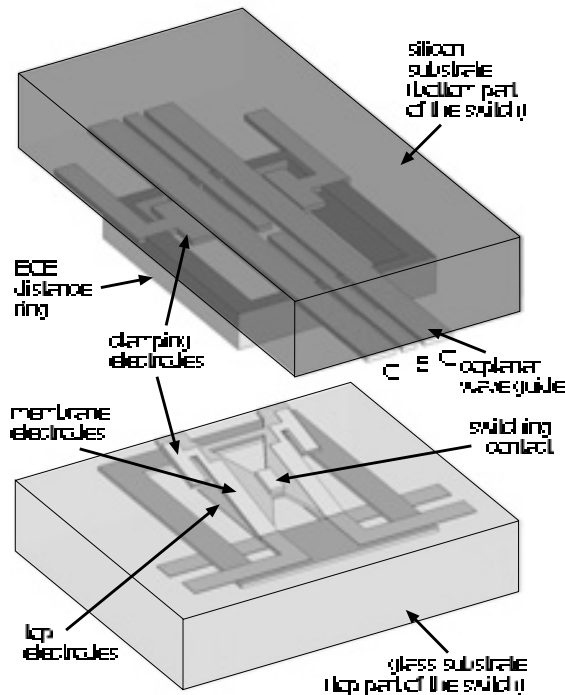
A curled actuator as illustrated in Figure 26(a), a so-called touch-mode actuator [161] moving in a zipper-like matter, improves the conventional design. It offers the possibility to have a large deflection of the cantilever tip by maintaining a small effective electrode distance for low voltage isolation. For a switch, this actuation concept allows the contact distance to exceed the sacrificial layer thickness by factors of over 10. However, the actuation voltages typically reported are not as low as expected [95, 97, 105]. This can be explained by the fact that the stress gradient in the film needed for the characteristic out-of-plane bending has to be overcome by the electrostatic force in order to close the switch. A thinner, less stiff film would also result in a curled beam and would require less energy to be actuated. Unfortunately, a thinner film also decreases the restoring force making the switch more susceptible to failure caused by contact stiction. A possible solution to this problem is a counter-actuator providing active opening capability. This can be achieved by using a second low-voltage curled actuator flipped upside down. Figure 26(b) shows the combined actuator with the characteristic S-shape of the film moving between the bottom and top electrodes. Due to the two combined zipper actuators working on a very thin and flexible membrane, the vertical movement of successive parts of the membrane anticipates a characteristic rolling rather than a pure up-and-down movement.



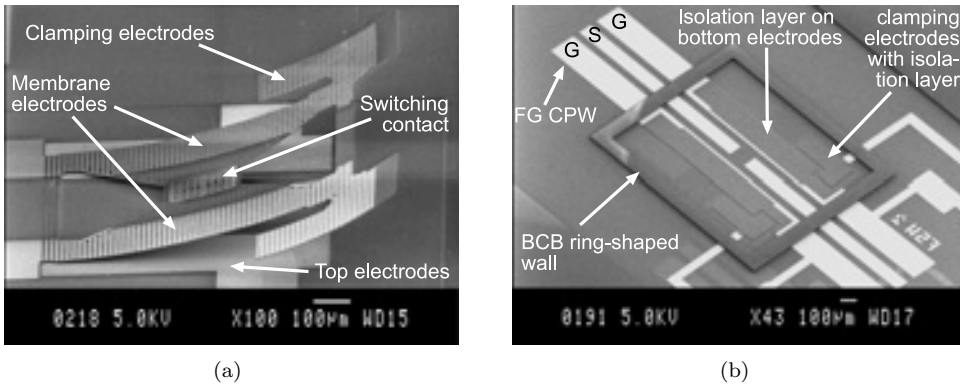
**Figure 27.** The S-shaped film actuator based switch in the off-state (a), during the transition (b) and in its on-state (c).

This S-shaped actuator was used for a gas valve with dimensions in the millimeter range in 1997 [267]. The novel switch, proposed in this thesis, is based on this concept and shown in Figure 27 in its on and off-states and during the transition between its two stable states. The switch is fabricated in two different parts which are assembled after the release-etch of the membrane, as illustrated in Figure 28. Figure 29 shows SEM pictures of the two parts of the switch and Figure 30 shows the two parts of an until now unpublished, smaller switch variant. A three-dimensional view of the top and the bottom part of the switch can be seen on page 76 and on page 77, respectively, as stereo images for parallel-eye viewing.

The characteristic S-shape of the film is created when assembling the two parts, which brings the tip of the free-etched membrane in contact with the bottom electrodes. To ensure the low-voltage zipper-like movement in both directions, the membrane tip must always be in contact with the opposite (bottom) part. For this purpose, the bottom part is endowed with additional clamping electrodes which electrostatically pull the membrane tip down to the bottom electrodes. The clamping electrodes are not necessary to establish the initial contact of the membrane tip with the bottom part, but are intended to improve the overall switch reliability, since stiction between



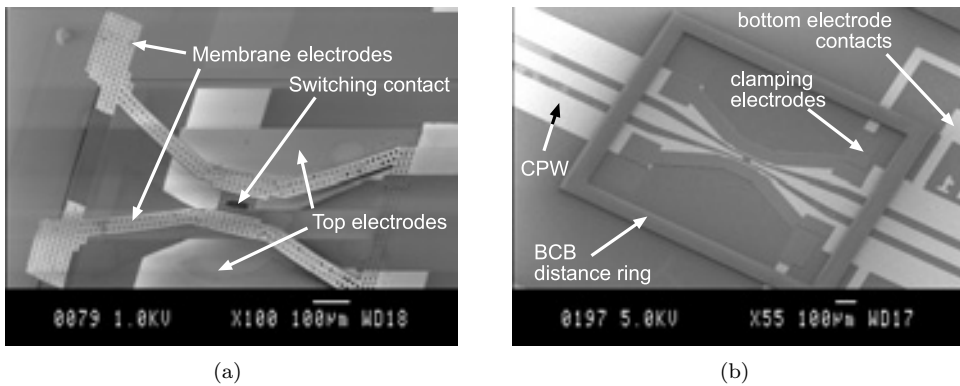
**Figure 28.** Schematic illustration of the film actuator switch with the switch shown upside-down. The silicon substrate contains the coplanar waveguide, the clamping electrodes and the BCB distance ring. The glass substrate contains the top electrodes, the silicon nitride film with the membrane electrodes and the switching contact. Details such as isolation layers and etch holes are not shown in the drawing.



**Figure 29.** SEM-pictures of the two parts of the switch before the final assembly: (a) top part with free-etched membrane; (b) bottom part with coplanar waveguide and polymer sealing ring defining the distance between the two parts.

the membrane and the top electrodes could prevent the membrane from curling down after opening the switch, which might prohibit an intended closing of the switch. Also, the clamping electrodes do not require any additional photolithography or etching step, and, since the clamping parts on the membrane are electrically connected to the membrane electrodes, not even additional interconnections or contact pads on the top part are needed.

It should be noted that the stress gradient in the membrane, which puts the tip of the membrane in contact with the bottom part of the switch, does not significantly increase the actuation voltage as in the case of a single curled actuator, since the membrane is very thin and flexible with very little stored spring energy.



**Figure 30.** SEM-pictures of the top (a) and the bottom (b) parts of a smaller, until now unpublished switch variant.

### 3.2 Discussion of the concept

The concept of the S-shaped film actuator based switch has the following advantages:

- The touch-mode actuators require very low actuation voltage to open and close the switch.
- The contact distance can basically be designed independent of the actuation voltage.
- The upper touch-mode actuator provides an active opening capability, which, to a certain extent, gives external control over the opening force and increases the overall switch reliability.
- The vertical electrostatic actuation principle provides almost-zero power consumption.
- The vertically moving actuator is fabricated by standard surface micromachining process steps.
- Due to the active opening capability, no passive restoring spring force is required, which allows a design with a very thin and flexible membrane, further lowering the actuation voltage.
- The large contact distance allows very large overlapping contact areas without compromising the RF isolation. Large nominal (or overlapping) contact areas do not directly lower the contact resistance if the contact force is not increased. However, they provide a better heat distribution from the effective contact spots heated by the dissipated power of the signal current flowing over the contact interface. Avoiding local overheating is one of the main measures increasing contact reliability for switching larger currents (see Section 2.2.3).
- The film rolls longitudinally in the direction of the signal transmission line. Thus, the ground lines of the coplanar waveguide are used as the bottom electrodes, and the membrane electrodes are placed directly above them, congruent with the bottom electrodes. Such a design results in minimum distortions of the wave propagation in the on-state, and in an almost perfect open line in the off-state.
- The contact pressure on the switching bar is created symmetrically both laterally and longitudinally in the direction of the transmission line.
- The fabrication of the part with the moving structure (top part) is carried out on a different substrate than the part with the signal line to be switched (bottom part). That means that basically there are only very few restrictions in fabrication compatibility, and the top part with the membrane can be transferred to any type of substrate containing RF circuits<sup>21</sup>. In paper 7, the upper

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<sup>21</sup>The bonding process is the only restricting fabrication step, since mis-matched coefficients of thermal expansion (CTE) of the two substrates might evolve problems if the bonding process requires a temperature typically above 100 °C.

part of a capacitive version of the switch is intended to be placed directly onto a substrate with a meander-shaped reconfigurable antenna.

- The assembly can be done either by flip-chip-like pick-and-place techniques (flexible, space efficient on the top wafer, higher yield), or by full wafer bonding if the switch density is high enough for economical reasons. Both methods result in an integrated and protected "ready-to-dice" package.
- The final vertical distance between the two parts can be controlled very accurately by the thickness of the polymer distance ring, as opposed to other switch transfer concepts reported in the literature, for instance thermal compression bonding [110].
- The actuation mechanism is suitable both for shunt and series switch configuration.

Furthermore, it should be noted that this concept can also be realized in a true single-pole-double-through (SPDT) configuration using only one single switch: The input line, electrically connected to the membrane as in the case of a so-called in-line switch, can be switched via the membrane either to a top or bottom contact on the top and bottom parts of the switch, respectively.

Besides the promising advantages of the switch concept, it also has some disadvantages:

- Switch control is more complicated since the concept requires a total of four electrodes, compared to two electrodes in a conventional switch design. However, only the potential of the membrane electrodes has to be altered to switch from one state to the other, as illustrated in Table 12. Also, the clamping electrodes were found not to be necessary for operating the prototype devices, and might be left out if not considered for reliability reasons. Thus, the switch can be operated by three control potentials which are necessary for any concept with active opening (push-pull) capability.
- The vertically stacked, three actuation electrode structure increases the fabrication complexity and the production costs.
- The two parts of the switch have to be electrically interconnected which is not trivial, depending on the overall packaging concept.
- The very thin membrane undergoes complex mechanical deformations compared to a cantilever in a conventional switch design, which might be a reliability issue,

**Table 12.** Driving potentials of the actuation electrodes during operation of the switch, as referred to Figure 27(a).

Switch state	$V_{TOP}$	$V_{MEM}$	$V_{BOT}=\text{GND}$	$V_{CLAMP}$
On	+	+	0	+
Off	+	0	0	+



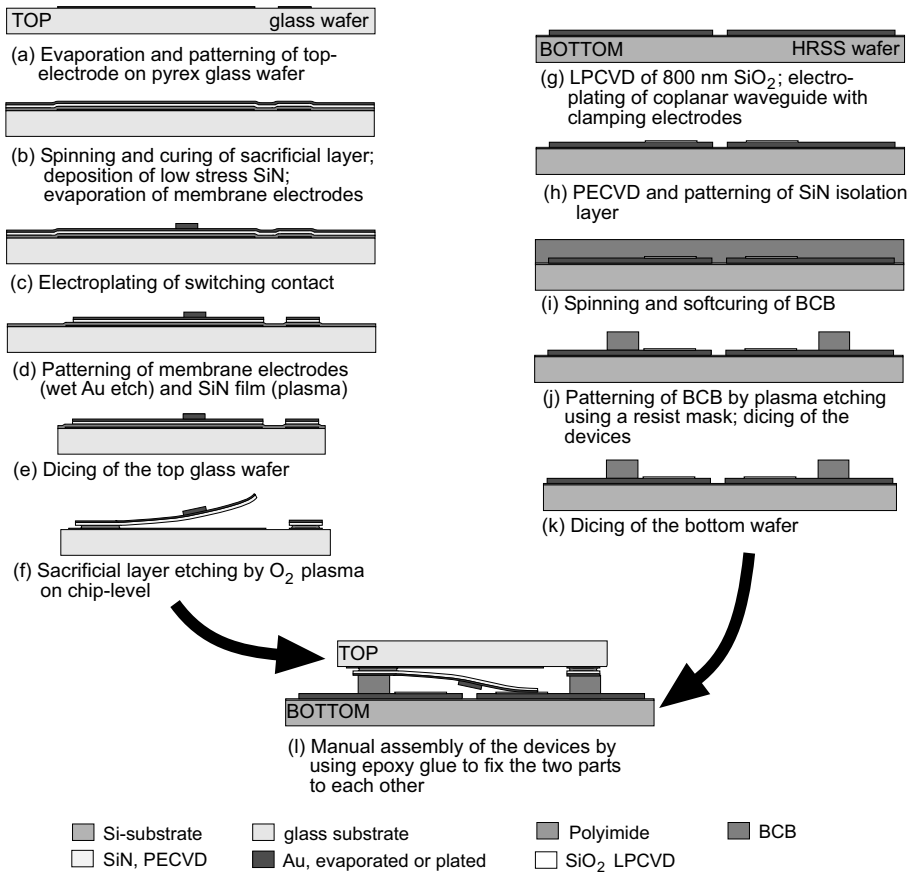
even though, by experience, MEMS contact switches rather die due to electrical contact problems than due to failure of their mechanical actuators.

- An "all-metal switch" design without electrical isolation layers between the actuation electrodes to avoid electrostatic charging problems is either more complex (metal-isolator-metal sandwich of the membrane) or mechanically more vulnerable (pure metal membrane with an isolating plug supporting the metal contact bar). In both cases, the stress imposed by the rolling movement of the film might involve mechanical reliability problems.
- Due to the flexibility of the thin membrane, the contact force is rather small if the electrodes cannot be placed very close to the contact bar (which is often not possible for DC isolation or for RF design reasons). A solution to this problem is a higher local stiffness of the membrane, just around the contacts. However, that complicates the fabrication, decreases the reliability (mechanical stress at the transition of the thin to the thicker membrane part), and results in a larger structure consisting of a conventional, stiff platform with two thin-membrane touch-mode actuators.
- The transfer-bonded switch concept involves typical problems such as alignment accuracy between the two parts/wafers.
- The contact bar "rolls" with the membrane which means that always the same end of the bar closes the signal line last and opens it first. Thus, contact degradation due to discharging and sparking always occurs on the same end.

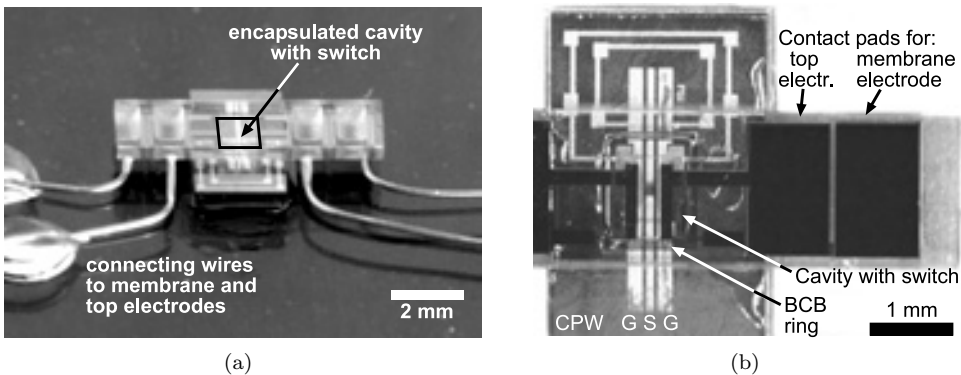
### 3.3 The first prototypes

The fabrication of the switch prototypes is done on two different substrates, as illustrated in the process flow in Figure 31. A detailed process description is given in paper 4. For easier access to the electrical contacts of the top and the bottom parts which are facing each other, the two parts were assembled manually in a cross shape, as shown in Figure 32. The overall device is quite large because of the many large contact pads, but the encapsulated cavity with the switch has a size of about  $1 \text{ mm}^2$  and is just a small fraction of the total construct, as outlined in Figure 32(a). However, also the switch itself is quite large compared to other MEMS switches, even when considering that it is already encapsulated. In the current design, the ratio of the vertical membrane displacement to the membrane length is about 1:65, imposing very little mechanical stress on the membrane. Therefore, it is assumed that the lateral dimensions of the switch can be downscaled without compromising either the actuation voltage or the contact distance, and hopefully not the reliability either.

Some of the advantages of this switch concept, as discussed in Section 3.2, are related to the possibility of having a very large contact distance and very large overlapping contact areas without sacrificing either the RF performance or the low actuation voltage. The prototypes have an off-state contact distance of  $14.2 \text{ }\mu\text{m}$ , which is about 5 to 7 times larger than in conventional designs (compare Table 8 on page 40), and with  $3500 \text{ }\mu\text{m}^2$  for each contact, the overlapping contact area is about 10 times larger than in conventional designs.



**Figure 31.** Process flow of the prototype fabrication of the S-shaped film actuator based switch.



**Figure 32.** Photographs with full view (a) and close-up view (b) of one of the manually assembled prototype switches.

**Table 13.** Comparison of the film actuator switch prototype with the OMRON RF MEMS switch [109].

	OMRON switch	Prototype of film actuator based switch
switch type	series metal contact	series metal contact
operation principle	vertically moving membrane	S-shaped rolling film actuator
actuation for closing	electrostatic forces	electrostatic forces
actuation for opening	spring forces	electrostatic forces
structural material	monocrystalline silicon	PECVD silicon nitride
signal line and contact	1.0–1.6 $\mu\text{m}$ sputtered gold [160]	2.0 $\mu\text{m}$ electroplated gold
packaged chip size	3.0×2.0×1.0 mm <sup>3</sup> <sup>a</sup>	1.7×1.4×1.0 mm <sup>3</sup> <sup>b</sup>
membrane size	1400×1700 $\mu\text{m}^2$	950×900 $\mu\text{m}^2$
membrane thickness	18–24 $\mu\text{m}$ [75]	1 $\mu\text{m}$ SiN + 0.190 $\mu\text{m}$ Au/Cr
total electrode size	2×1 000 000 $\mu\text{m}^2$	2×82 500 $\mu\text{m}^2$
switching contact area	unknown to the authors	2×3500 $\mu\text{m}^2$
open contact distance	3 $\mu\text{m}$	14.2 $\mu\text{m}$
actuation voltage	19.2 V to close	12 V to open, 15.8 V to close
contact force	5 mN at 24 V, calculated	102 $\mu\text{N}$ at 15 V, simulated
contact resistance	<50 m $\Omega$	0.65 $\Omega$
RF isolation	40–30 dB (2–10 GHz)	45–30 dB (2–15 GHz)
insertion loss	<0.5 dB	2.5 dB at 10 GHz
open-state capacitance	5 fF [75]	4.2 fF
switch transfer by	anodic bonding (>350 °C) <sup>c</sup>	patterned adhesive bonding (BCB, 280 °C) <sup>d</sup>
packaging	CSP on wafer level by glass-frit bonding (450 °C)	together with switch transfer, "ready-to-dice" package
sealing ring	250 $\mu\text{m}$ wide, 10 $\mu\text{m}$ high, glass-frit	200 $\mu\text{m}$ wide, 18.2 $\mu\text{m}$ high, BCB
package gas tightness	hermetic package	near-hermetic package

<sup>a</sup>a smaller device with a chip-scale-package (CSP) of 1.8×1.8×1.0 mm<sup>3</sup> was recently presented [158]

<sup>b</sup>size of the packaged prototype switch, excluding contact pad areas for manual probing

<sup>c</sup>thermo-compression bonding in [158]

<sup>d</sup>design target: to be done by patterned adhesive full-wafer bonding with BCB; the prototypes are assembled manually and fixed by epoxy drops

Table 13 compares characteristic design and measurement data of the switch to data of the OMRON RF MEMS switch, which is of similar dimensions and was already discussed in Sections 2.1.4, 2.2.1 and 2.2.2. The OMRON switch is currently one of the most mature MEMS switches in device performance and in its overall concept including the packaging, developed since about 1996 by the world's largest manufacturer of mechanical relays.

### 3.3.1 Prototype packaging

The switch is designed to be housed by the cavity between the two substrates, which is defined by the distance ring fabricated on the bottom wafer. Benzocyclobutene (BCB), an epoxy-based polymer from the Dow Chemical Company, is used to form the ring. The patterned polymer is hard-cured during the final assembly of the switch, leading to a relatively strong bond between the two parts. More information about adhesive bonding and patterned BCB-bonding can be found in Section 4. BCB, as all polymers, is permeable to gases and this type of packaging can therefore not be considered hermetic. Packaging with BCB is often referred to as "near-hermetic" packaging because, among polymers, BCB is one of the least permeable materials. However, it is a sufficient barrier against liquids and at least protects the switch from water during the dicing process and from dust during other post-processes outside clean-room facilities. Some advantages of BCB are its low dielectric constant and its low losses at high frequencies, making it ideally suitable as a sealing ring material to encapsulate RF devices. Also, the good planarization ability of the polymer allows the use of thick electrical interconnection lines penetrating through the wall (see Section 4). The BCB thickness is  $18.2\text{ }\mu\text{m}$  for all of the fabricated prototypes, and the width of the wall is  $100\text{ }\mu\text{m}$ .

The assembly of the switch can be done either individually on chip level (chip to chip) or on wafer level, and in the latter case either by individual pick-and-place techniques (chip to wafer) or by full wafer bonding (wafer to wafer). The sequential order of the dicing and the free-etching of the switch membrane is different for the chip-to-chip (or chip-to-wafer) assembly than for the wafer-to-wafer assembly:

**Chip-to-chip or chip-to-wafer:** 1) dicing of the top wafer; 2) sacrificial layer etching of the individual chips; 3) mounting of the chips to their counterparts, either on chip-level or on wafer-level

**Wafer-to-wafer:** 1) sacrificial layer etching on wafer-level; 2) assembly of the two parts by full-wafer bonding; 3) dicing of the chips (see also Section 4.2)

The design target is assembly by wafer-to-wafer bonding, packaging all of the switches on the wafer in one single process step. However, the prototype devices were assembled on chip level as shown in Figure 32. Also, the two parts were additionally fixed by epoxy. The electrodes on the top part, which are upside down, are accessed by electrical wires glued with conducting epoxy to the contact pads, and the electrodes on the bottom part are accessed by semiconductor measurement probes.

### 3.3.2 Prototype evaluation

Detailed information about the prototype characterization is provided in papers 3 and 4. Table 14 summarizes the most important results and the following paragraphs comment on some of them, especially on the insertion loss and on the contact force, which were both found not to be as good as expected.

The actuation voltages to open and close the switch are well below 20 V, which is remarkable for first prototype devices.

The measured insertion loss of the prototypes is far higher than expected and completely unsatisfactory for a micromechanical switch. The insertion loss cannot be

**Table 14.** Summary of the most important results of the prototype evaluation.

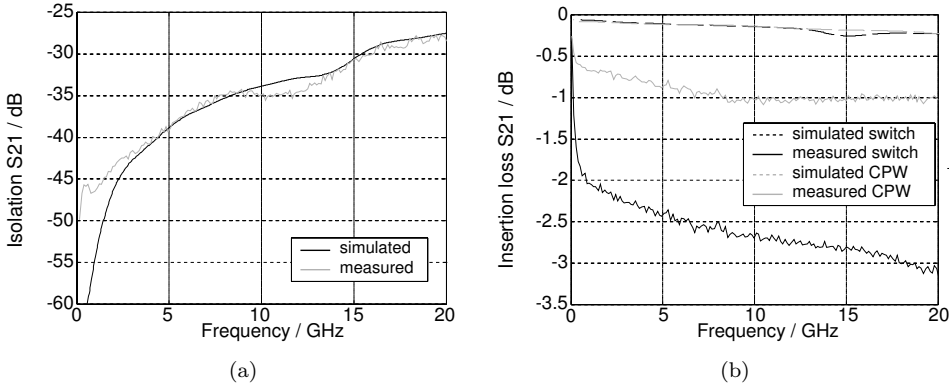
Investigated parameter	Measurement value	Comments
actuation voltage (open)	12.0 V	
actuation voltage (close)	15.8 V	
clamping voltage	19.2 V	to fully tie down the clamping electrodes
total DC resistance	650 m $\Omega$	including a 3.3 mm long signal line
contact resistance	275 $\Omega$	per contact
contact force	102 $\mu$ N	per contact, simulated with ANSYS at an actuation voltage of 15 V
RF isolation	45 dB (<2 GHz), 30 dB (15 GHz)	contact distance of 14.2 $\mu$ m, overlapping contact area of 3500 $\mu$ m <sup>2</sup> per contact
series capacitance	4.2 fF	simulated with FEMLAB
insertion loss	2.5 dB at 10 GHz	see comments in the text

explained by the reflections, which are not so high and the waveguide is relatively well matched (47.4  $\Omega$  measured, 50  $\Omega$  design target). Radiation losses can also be excluded due to the small electrical dimensions of the device, especially for frequencies in the lower GHz range. Therefore, the measured too high insertion loss is assumed mainly to be caused by the following:

- The calibration was not carried out with an on-wafer calibration set, but with a calibration kit on a low-loss ceramic substrate.
- The coplanar waveguide itself has quite a high loss. This can be concluded from the measurement data of the waveguide without switch, included in Figure 33(b). The simulated insertion loss of the waveguide is also shown in the figure, and there is quite a large discrepancy between the simulated and the measured data, as in the case of the waveguide with the switch. The substrate resistivity is 1500  $\Omega$ ·cm and probably too low for the waveguide which is separated from the substrate only by a 800 nm thick silicon dioxide layer.

However, even considering the high losses of the waveguide, the measurements show a remaining insertion loss of about 1.5 dB. The reason for this is not fully understood since the measured total DC resistance only creates an estimated insertion loss of about 0.06 dB (calculated value for a 50  $\Omega$  system).

The contact force was estimated by simulations with ANSYS to be 102  $\mu$ N per contact. This value is relatively low and corresponds to about the minimum value necessary for a stable contact resistance in the sub- $\Omega$  range, as described in the literature (see Section 2.2.3). The measured contact resistance of 275 m $\Omega$  per contact is comparable to the values reported in the literature for the simulated contact force, which confirms the simulation model. Compared to the size of the actuator, however, the contact force is relatively low. The total size of the overlapping area of the bottom and membrane electrodes is about 150 000  $\mu$ m<sup>2</sup>, which potentially could create an over 300 times larger force of about 30  $\mu$ N. Since the membrane is very flexible, which is



**Figure 33.** RF measurements of one of the prototype switches: (a) isolation; (b) insertion loss.

a design criterion, most of the force is lost due to touching electrodes. The contact force could be increased by

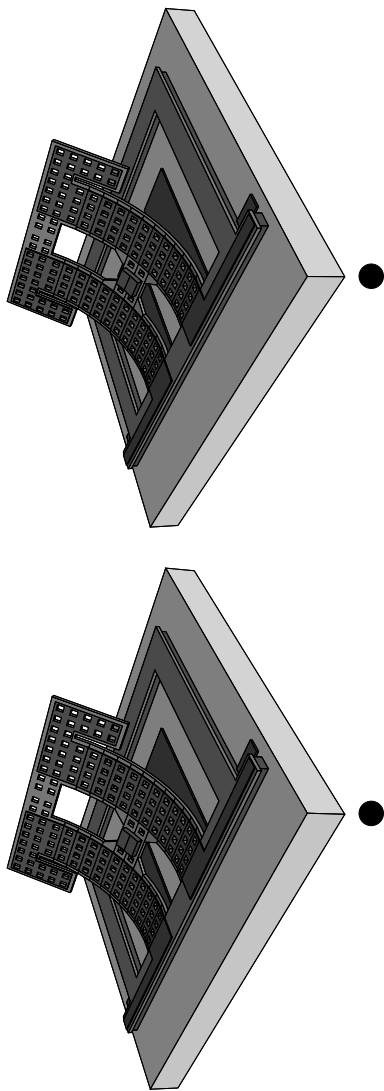
- A stiffer membrane around the contacts, which can be achieved by using a different material for that part of the membrane or by depositing a thicker membrane layer around the contacts. Both methods increase the fabrication complexity.
- Closer positioning of the actuation electrodes to the contacts, which can be achieved by changing the overall RF design with a shorter distance between the ground lines and the signal line (the current design has a relatively large gap of 50  $\mu\text{m}$ ).
- Thicker metal contacts on the membrane: creates a larger effective area of the electrostatic actuator since the touching point between the membrane electrodes and the bottom electrodes is moved farther away from the contacts.
- An isolation layer with a larger dielectric constant basically also increases the contact force. However, the isolation layer only affects the contact force if its thickness is a substantial part of the total distance between the actuation electrodes and only the electrode area not touching the isolation layer contributes to the contact force. Also, the current design uses silicon nitride with a relative dielectric constant of approximately 7.5; materials with larger permeability are more difficult to process and have an increased susceptibility to electrostatic stiction (see Section 2.2.4).

The reliability of the switch has so far not been investigated extensively. Also the process parameters, in particular those of the silicon nitride deposition, should be optimized before investigating and publishing on the reliability of the concept, and the switches should be evaluated in an inert atmosphere to be able to compare the results to other MEMS switches. However, the author made some observations during

the evaluation of the prototype switches, which was carried out in air at atmospheric pressure and room temperature:

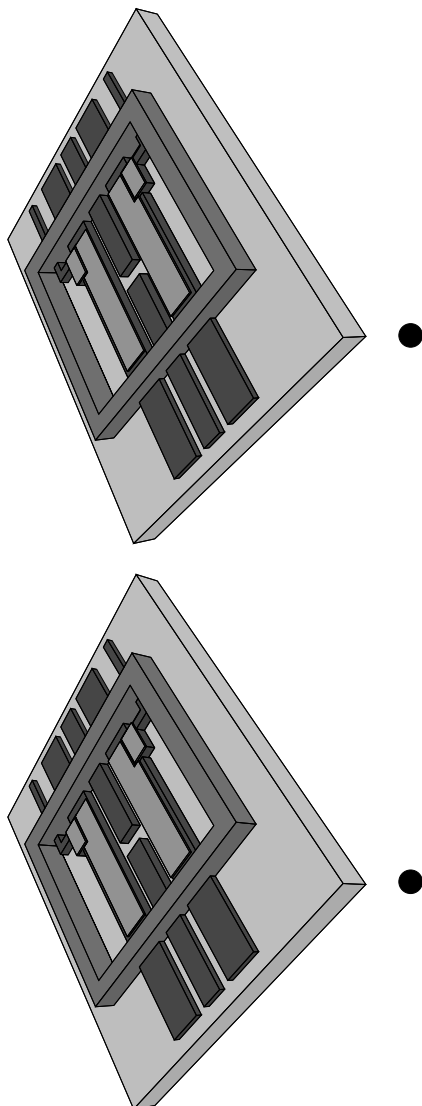
- Electrostatic stiction could not be observed. This can be explained by the relatively low actuation voltages (see Section 2.2.4) and the active opening capability of the switch. However, the actuator was not kept actively in one state for longer periods of time (hours or days) to investigate this phenomenon.
- The membrane is mechanically extremely strong. All 34 membranes on all of the processed wafers could be released successfully and the membranes can be pushed with probe tips without being damaged. During the individual assembly of the two parts of the switch, some membranes were destroyed when the tip of the out-of-plane bending membrane latched behind the thick BCB structures of the bottom wafer (thus, if the pre-alignment is done incautiously).
- The clamping electrodes were found not to be necessary since the membrane curls sufficiently out-of-plane and touches the bottom electrodes already after the assembly of the switch and each time after releasing the actuation voltage between the membrane and the top electrodes. Also, stiction between these electrodes was not found to be a problem during the prototype evaluation. However, depending on the design it might make sense to include clamping electrodes for reliability reasons (see Section 3.1).
- The membrane bilayer consisting of silicon nitride and a thin gold layer withstands thermal induced stress when exposed to temperatures of up to 280 °C needed for the proposed assembly process by full-wafer bonding.
- Once the switch is assembled it is mechanically very robust and shock-resistant.
- Irreversible breakdown occurs at about 90 to 100 V between the membrane and the bottom electrodes.
- The main reliability problem of the prototypes is the metal contacts. It was observed that either ohmic contact could not be established when closing the switch or the contacts got stuck easily. The reason for the former failure mechanism is most probably a combination of the relatively low contact forces and of surface contaminants, since the prototype switches were not assembled in clean-room environment and since the contacts can not be cleaned after releasing the membrane. The stiction problems are assumed to be caused by the soft electroplated gold and by the very large contact areas.



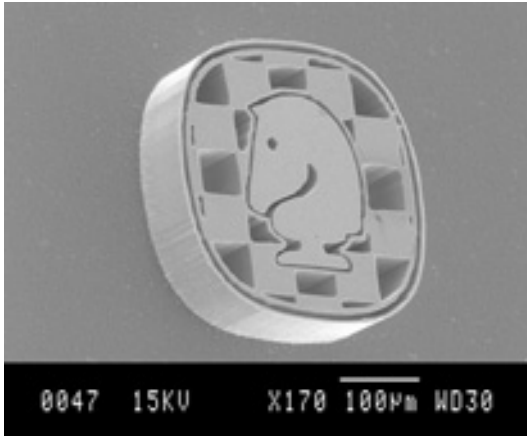


Stereo images of the top part (this page) and the bottom part (next page) of the switch.



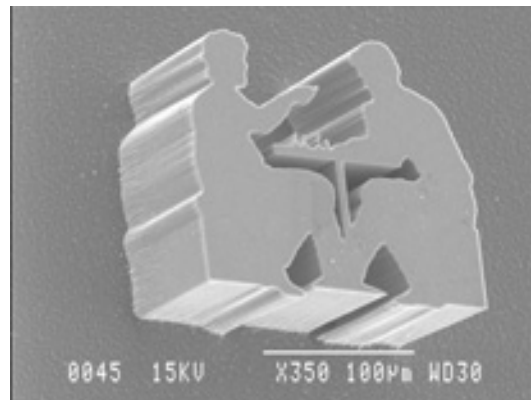


To view the stereo images, rotate the book so that the two black filled circles are on a horizontal line about 30 cm in front of you. Relax your eye muscles bringing your eyes in parallel view, until the two circles are on top of each other. If a crossed-eye view is more convenient, do not hesitate to cut out the drawings and put them beside each other in the opposite way.



*Chess teaches the humility of defeat.*

*Source Unknown*



*Chess is ruthless: you've got to be prepared to kill people.*

*Nigel Short  
1965–, British Chess Grand Master*

*A pile of rocks ceases to be a rock when somebody contemplates it with the idea of a cathedral in mind.*

*Antoine de Saint-Exupéry  
1900–1944, French Aviator, Writer,  
in "Flight to Arras"*

## 4 Packaging by patterned adhesive full-wafer bonding

### 4.1 Patterned adhesive full-wafer bonding

Some polymers used for adhesive bonding can be patterned before the bonding process, as mentioned in Section 2.3.3. Thus, the structures in the polymer create cavities directly in the bonding layer, which can be used to house MEMS devices or as channels or chambers in microfluidic applications.

The requirements for a polymer suitable for patterned adhesive bonding are the following:

- The polymer must be suitable for adhesive wafer bonding (low or no outgasing avoiding delamination in the bonding layer, sufficient physical strength in shear and tension, low viscosity).
- The polymer has to be suitable for patterning either by etching techniques, through photosensitive components of the polymer chain or by local deposition processes such as screen or contact printing.
- The patterning of the polymer should not require a fully cross-linked polymer, i.e. the polymer should still be adhesive enough after the patterning procedure to result in a strong bond.

In the work presented in this thesis, divinylsiloxane bisbenzocyclobutene (BCB) from the Dow Chemical Company is used. BCB is an epoxy-based thermo-set polymer which has been used in electronics production for dielectric layers, as underfill material, for stress-distribution layers and for opto-electronic applications since about 1993 [268–277]. The material has a low dielectric permittivity which makes it very suitable for RF applications and for high clock-rate digital circuits [278]. Some of the characteristic properties of BCB are listed in Table 15. It is available in a version which can be patterned by dry-etching [279, 280] and in a negative photosensitive version [281]. Both types are suitable for bonding, since the polymer fulfills the requirements listed above. Outgasing during the curing process is not detectable since its polymerization does not involve catalysts [229]. It can be spin-coated in a single layer with a thickness up to 40  $\mu\text{m}$  and multi-layer coating is also possible.

BCB was used as an adhesive material for full-wafer bonding to fabricate flow channels [282] and to transfer structures or films from one wafer to another [200]. In these applications the material was unpatterned and uncured before the bonding procedure. The fabrication of three-dimensional structures by bonding with photo-sensitive BCB was investigated by another research group [230] almost in parallel to

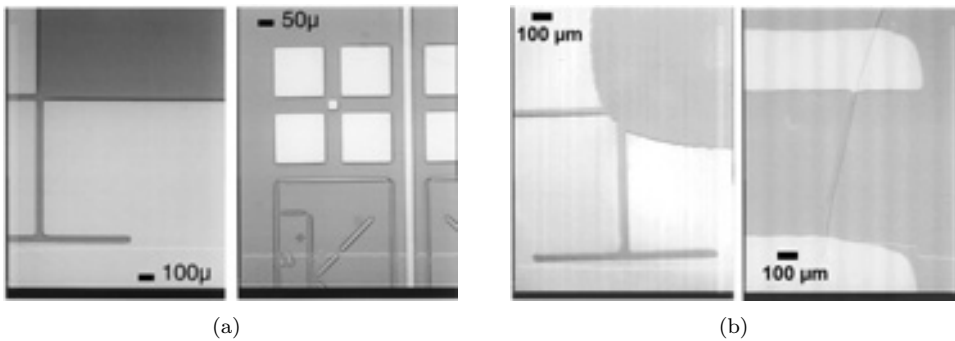
**Table 15.** Characteristic properties of the polymer benzocyclobutene.

Property	Unit	Value
hard curing temperature	°C	250–300
soft curing temperature	°C	190–210
glass transition temperature	°C	>350
polymerization level, precursor	–	40% (dry etch), 50% (photosensitive)
Young's modulus	GPa	$2.0 \pm 0.2$
tensile strength	MPa	$85 \pm 9$
coefficient of thermal expansion	ppm K <sup>-1</sup>	52 (20 °C), 36–61 (-50..175 °C)
elongation	–	$6 \pm 2.5$ %
density	kg·dm <sup>-3</sup>	1.05
moisture uptake (weight)		<0.12% (81% r.h.)
shrinkage during curing	–	<5%
planarization degree <sup>a</sup>	–	>90%
relative permittivity	–	2.65 (1–10 GHz)
power dissipation factor	–	0.0008–0.002 (1 MHz–10 GHz)
breakdown voltage	kV·cm <sup>-1</sup>	3000

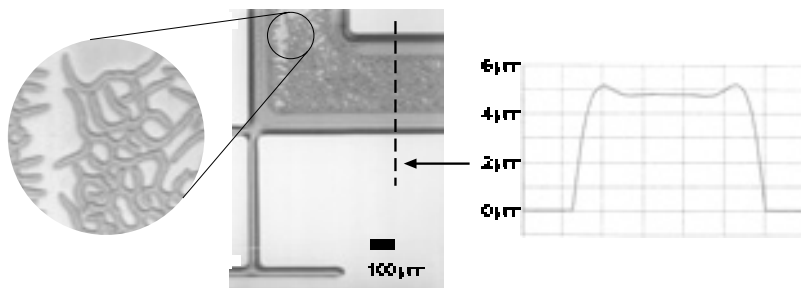
<sup>a</sup>the degree of planarization is defined as  $\left(1 - \frac{t_s}{t}\right) \cdot 100\%$ , where  $t$  is the thickness of the structures to be covered, and  $t_s$  is the resulting bump on the top surface of the polymer when spun onto the structures with a thickness of  $2 \cdot t$

the work reported in paper 1, and the bonding of patterned to unpatterned polyimide structures was published in [283].

The dry-etch BCB has to be cured before its patterning procedure in order to be chemically and physically resistant enough for the plasma etching. If fully cross-linked, the material is no longer "sticky" enough for the subsequent bonding. Thus, the BCB should be partly cross-linked in a so-called "soft-cure" process step to about 50% of

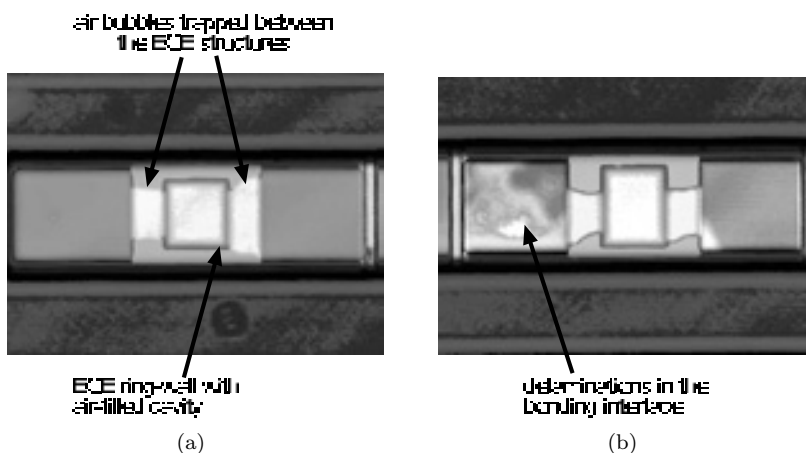


**Figure 34.** Shape retention of bonded dry-etch BCB patterns depending on the soft-curing parameters: (a) well bonded pattern, soft-cured at 200 °C; (b) shape reflow due to an insufficient soft-curing temperature of 160 °C.



**Figure 35.** Incomplete bond due to surface non-uniformity of the patterned structures, soft-cured at 190 °C, bonded at a pressure of 2.5 bar.

polymerization, to ensure both sufficient consistency and chemical resistance for the patterning processes and enough adhesion for the bonding. Paper 1 investigates the optimal processing parameters for the soft-curing and the bonding of dry-etch and photosensitive BCB. For dry-etch BCB, the most suitable soft-curing temperature range was found to be 190–210 °C for 40 minutes. The optimal bonding pressure depends on the mask pattern. However, a pressure of 2.0–2.5 bar on a 4 inch chuck was found to be a suitable start value for masks with a remaining BCB area of 40–70%. The BCB is fully cross-linked during the bonding when the bond chucks are heated above 250 °C for about 1 hour. The bonding temperature can be raised to decrease the bonding time, but BCB becomes thermally unstable above 300 °C and the bond strength decreases drastically [277]. Figure 34(a) shows a well bonded BCB pattern and Figure 34(b) illustrates the effect of insufficient soft-curing of the dry-etch BCB before the bonding procedure. The same mask was used in both experiments and the BCB thickness was about 5  $\mu\text{m}$ . The bonding pressure was in both cases 2.5 bar applied for 1 hour at a temperature of 250 °C.



**Figure 36.** Scanning acoustic microscope (SAM) pictures of (a) a homogeneous bond and (b) delamination in the bond interface, using a transducer with a frequency of 50 MHz.

Since the BCB is relatively rigid after the soft-curing, the bond quality depends strongly on the surface uniformity of the BCB pattern. For a BCB thickness of about  $5\text{ }\mu\text{m}$  the material is no longer soft enough to compensate for non-uniformities of about  $0.1\text{--}0.2\text{ }\mu\text{m}$ , as shown in Figure 35.

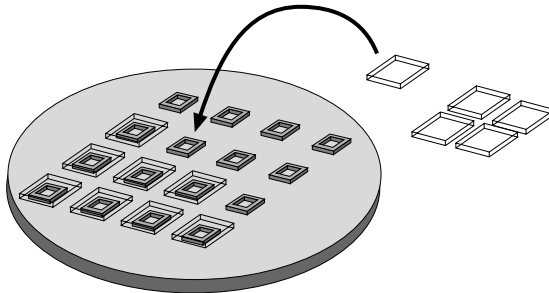
Delamination in the bonding interface can be detected visually by looking through a microscope when using a transparent top wafer, as shown in Figures 34 and 35, or by scanning acoustic microscopy (SAM), as illustrated in Figure 36.

## 4.2 Glass lid encapsulation by patterned adhesive wafer bonding

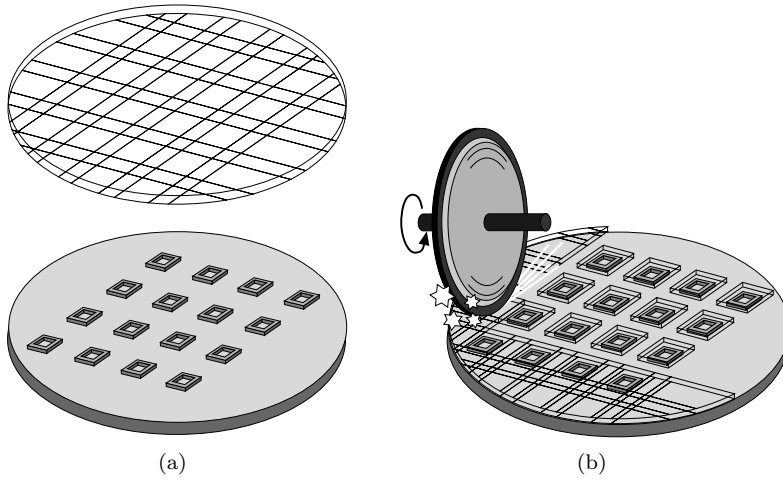
The conventional wafer-level encapsulation procedure by transferring individual sealing caps to the substrate wafer using flip-chip-like pick-and-place techniques is illustrated in Figure 37. Solder-bonding of thin silicon microcaps was investigated for this purpose [284] and B-stage epoxy was used as bonding material in [234]. As suggested in the previous section, patterned BCB is a very interesting material to encapsulate cavities in this way. Photo-sensitive BCB rings on the target wafer were found to be suitable for individual glass lid packaging [255].

Paper 5 introduces a further parallelization of this wafer-level packaging technique, where the encapsulation procedure is done by full-wafer process steps (so-called wafer-scale packaging): a covering glass wafer is bonded to the substrate wafer containing the BCB patterns, and the glass caps are separated after the bonding by dicing only the top glass wafer, as illustrated in Figure 38. The dicing of the top wafer without harming the bottom substrate allows access to electrical contact pads on the bottom substrate. This was not possible in a reported packaging procedure using AuSn eutectic bonding, where the capping wafer and the substrate wafer were diced together [285]. The selective dicing is possible if the BCB pattern is taller than the vertical accuracy of the die saw, which might not be the case for many practical applications. Therefore, paper 5 also presents a technique to solve this problem. The outlines of the final glass lids are sliced in the front-side of the glass wafer before the bonding, which allows a larger vertical accuracy of the dicing process after the bonding, as shown in the fabrication sketch in Figure 39.

It is also possible to separate the encapsulation lids by etching the top wafer



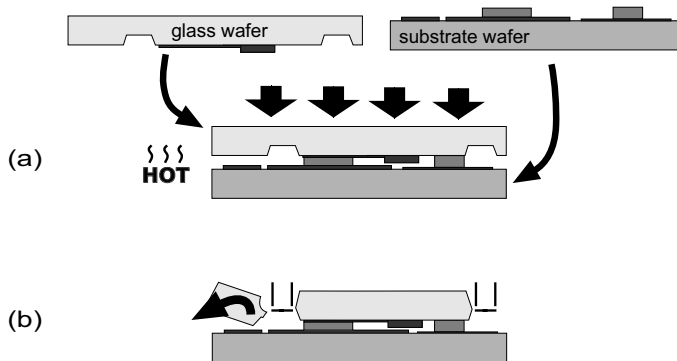
**Figure 37.** Glass lid encapsulation on wafer-level by pick-and-place bonding of individual caps.



**Figure 38.** Glass lid encapsulation by (a) full-wafer bonding with a patterned adhesive layer, and (b) dicing the glass caps after the bonding procedure.

instead of dicing it. This technique, creating sealing caps from a full wafer by etching trenches in the top wafer after the adhesive bonding procedure, was proposed recently to package surface acoustic wave (SAW) filters [286]. The choice of the right method depends on the application and on the substrate material. However, the dicing method proposed in papers 5 and 6 has the following advantages:

- very low cost, since done outside clean-room facilities with a very simple machine, especially compared to a through-wafer plasma etch process
- mask-less process
- very simple and robust



**Figure 39.** Illustration of the glass lid encapsulation process with pre-dicing of the front-side of the glass wafer, as described in the text. As shown in the drawing, the top glass wafer may also contain microsystem structures.

- no substrate material restrictions
- no special requirements on etch-stop layers

The main disadvantage of the dicing technique compared to the etching technique is that the dicing pattern is restricted to complete lines over the full wafer.

Figure 40(a) shows an array of glass lids created by this technique after dicing the top glass wafer only, and Figure 40(b) is a photograph of a single encapsulated chip.

### 4.3 Vertical and in-plane interconnections

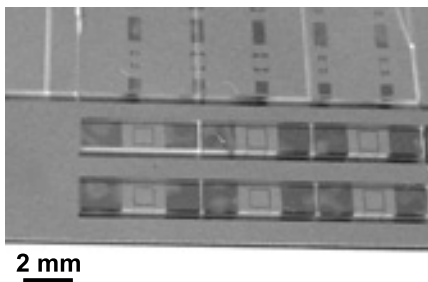
Through-wafer vias are often needed in packaging, when electrical components on the front-side of a chip have to be accessed from the back-side. Such vias are fabricated by etching holes through the substrate and filling them or by coating their sidewalls with an electrically conducting material.

The etching techniques for such vias in glass wafers presented so far are deep reactive ion etching (DRIE), adapted for pyrex glass [287], and electrochemical discharge drilling [288]. Almost in parallel to the work presented in this thesis, researchers at the Korea Institute of Science and Technology, in cooperation with the Samsung Advanced Institute of Technology (SAIT), published on sandblasted holes in pyrex glass wafers for through-wafer vias filled by gold, electroplated onto a Cr/Au seed layer on the sidewalls of the vias [234].

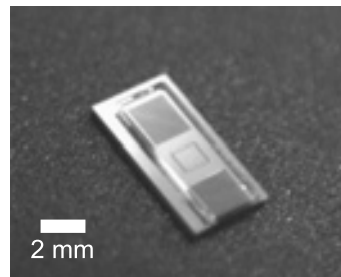
Paper 5 introduces a new technique to create through-wafer vias in glass substrates. The technique consists of two etch steps, where the first one is carried out from the backside "almost-through" the wafer and the second etch step finally completes the holes. Powder-blasting [289] is used as the first etch step, and the second etch step consists of a short wet etch in concentrated hydrofluoric acid (HF) to further etch the recesses down to the contact pads, as illustrated in Figure 41.

Splitting up the whole etch process into two steps has the following advantages:

- The main etch step from the back-side (here accomplished by powder-blasting) is carried out before any other process step on the glass wafer and can therefore be done outside clean-room facilities, lowering the fabrication costs.



(a)



(b)

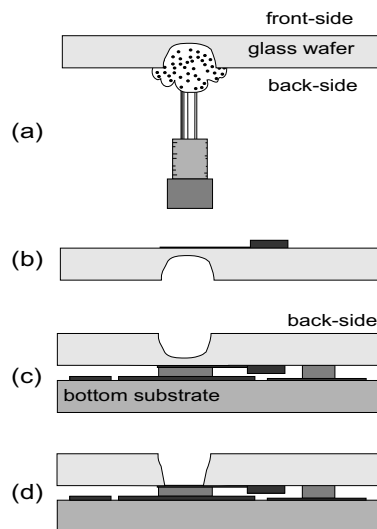
**Figure 40.** Photographs of glass lids created by patterned adhesive full-wafer bonding, (a) after dicing the top wafer and (b) a single chip after also dicing the substrate wafer.



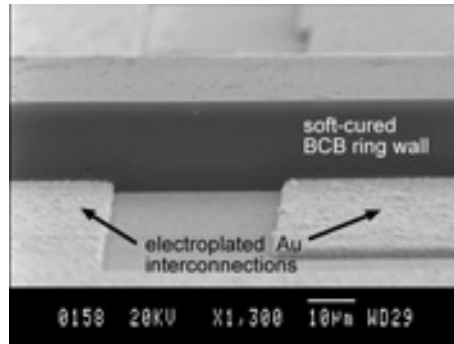
- Powder-blasting is a rather inexpensive process for relatively controlled deep etching in glass.
- The surface of the front-side is completely intact after etching the deep recesses by the first etching step; thus surface micromachining is basically unrestricted on the front-side.
- The metal contact pads to be accessed from the back-side act also as an etch-stop layer for the HF etch-step.
- Both etch processes are applicable on different types of glass wafers, even though Pyrex 7740 and Hoya SD-2 glass are the preferred substrates due to their high and uniform etch-rates in HF.
- The final wet-etch step is carried out after the bonding; thus, the structures on the front-side of the glass wafer are protected from the etching chemical.

Whenever adhesive bonding is used for the bonding, the polymer might not provide sufficient resistance against the chemical etch. In this case, the bond interface should not be exposed to the etch solution, which can be achieved by using a wafer holder with single side protection.

The two-step etch technique can basically also be applied on other substrates than glass using a different physical, chemical or plasma-etch combination. The main disadvantage of powder-blasting is the large feature size. It is difficult to achieve a height to width aspect ratio of over 2:1 and the sidewall slope is typically about



**Figure 41.** Overview of the fabrication procedure of through-wafer electrical interconnections in glass wafers, used as a packaging method together with patterned adhesive bonding: (a) powder-blasting of deep recesses in the back-side of the glass wafer; (b) surface-micromachining of structures and contact pads on the front side; (c) bonding; (d) short wet etch to complete the vias.



**Figure 42.** In-plane electrical interconnections of electroplated gold with a thickness of  $2\ \mu\text{m}$  penetrating through a  $18.2\ \mu\text{m}$  thick BCB wall.

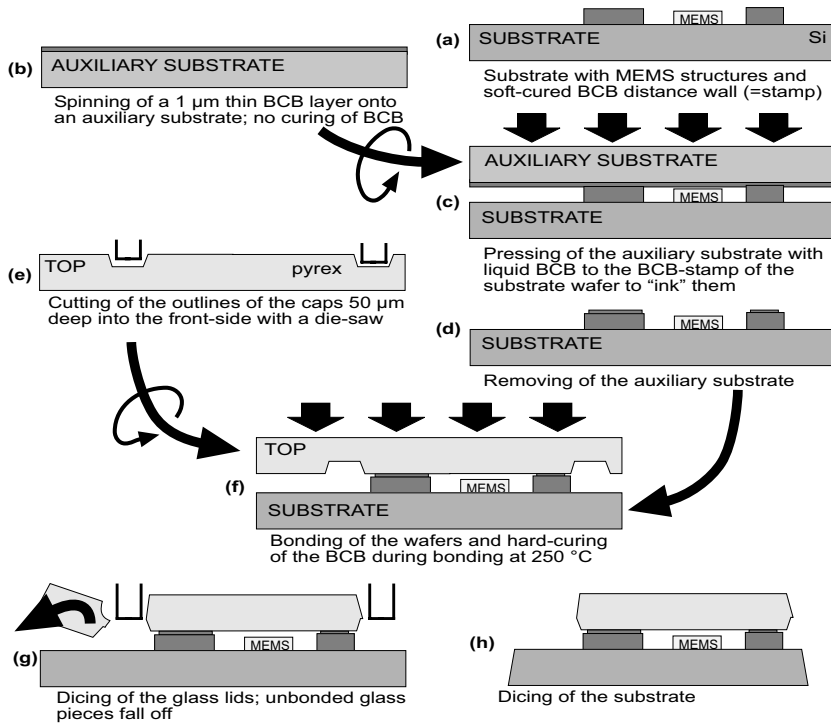
$70^\circ$  [289–291]. Thus, the minimum hole dimensions to etch about 90% deep into the wafer have to be of the order of the thickness of the wafer. Besides accessing the contact pads by wire bonding, as described in paper 5, the slope of the sidewalls offers the possibility to deposit a metal on the walls to electrically connect the contact pad to the back-side of the wafer (such a technique is used for the packaging of the OMRON switch [158]).

In-plane interconnections penetrating through the BCB wall can be made because of the excellent planarization properties of BCB [271], providing a sufficiently flat surface for the adhesive bonding. Even thick interconnection lines of  $2\ \mu\text{m}$  are very smoothly covered by a BCB film of a thickness of  $18.2\ \mu\text{m}$ , as illustrated in Figure 42.

#### 4.4 Bond strength improvement by bonding with a solid/liquid BCB pattern

The bond strength of a patterned BCB layer is lower than the bond strength of an unpatterned layer since the polymer has to be cross-linked to a certain degree before the bonding in order to be sufficiently chemically and physically resistant for the patterning procedure. Also, the soft-cured BCB is less elastic than an uncured and basically liquid BCB layer, which decreases the bonding yield in case of surface non-uniformities, as discussed in Section 4.1.

Paper 6 presents a simple method which combines the advantages of a patterned adhesive layer with the advantages of a liquid polymer phase before the bonding. The pattern in the adhesive layer is "inked" by pressing the substrate containing the pattern (bottom wafer) toward an auxiliary wafer with a thin and still uncured polymer layer. After removing the auxiliary wafer, a thin film is locally transferred to the pattern of the bottom wafer, which is finally bonded to the second wafer (top wafer). The procedure to fabricate glass lid encapsulations with this special bonding technique is illustrated in Figure 43. The fabrication is very robust and the BCB pattern on the bottom wafer can even be fully cured before the transfer of the film, which makes the final bonding result independent of crucial soft-cure parameters. Also, the pattern can basically be made with any other material, and other adhesives



**Figure 43.** Fabrication steps to create glass lid encapsulations by full-wafer bonding, using the bonding technique with solid/liquid BCB pattern for improved bond strength.

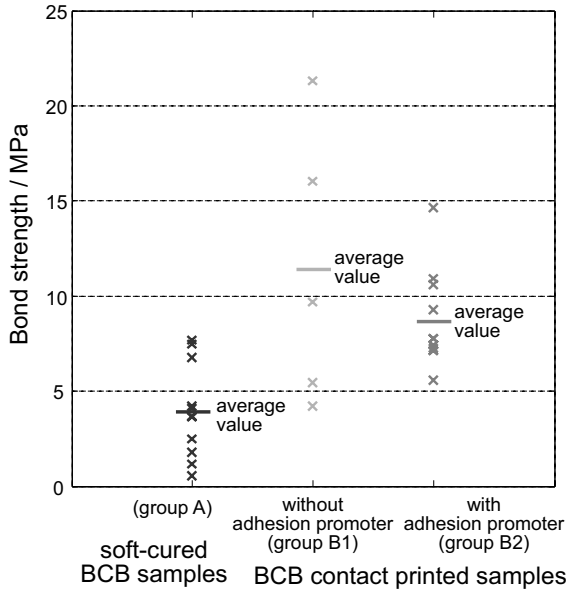
than BCB might be used to ink the structures.

Using BCB, features with a size as small as 10  $\mu\text{m}$  could successfully be inked and bonded. The bond strength improvement of this BCB contact printing technique compared to the conventional patterned BCB bonding, as described in Section 4.1, is illustrated in the diagram in Figure 44. The influence of adhesion promoter, the shape flow of the viscous BCB, the distortion of the negative imprint pattern in the BCB layer on the auxiliary wafer, and the failure mechanisms occurring during the tensile strength tests are discussed in detail in paper 6.

## 4.5 Hermeticity evaluation

The main disadvantage of polymers in packaging is that these organic materials are highly permeable to gases and moisture compared to ceramics and metals, as shown in Figure 45. Thus, adhesive bonding is in principle not suitable for hermetic packaging.

The hermeticity properties of BCB used in MEMS packaging were investigated in [255] and in papers 1 and 2 by helium leak tests following the procedure described in MIL-STD-833E, Method 1014.9 [292]. This method comprises a test procedure involving a gross and a fine leak test. For the gross leak test, the samples are put into a heated perfluorocarbon solution with low surface tension and a high boiling temperature of 140–200  $^{\circ}\text{C}$ , and defects in the sealing layer can be detected as gas

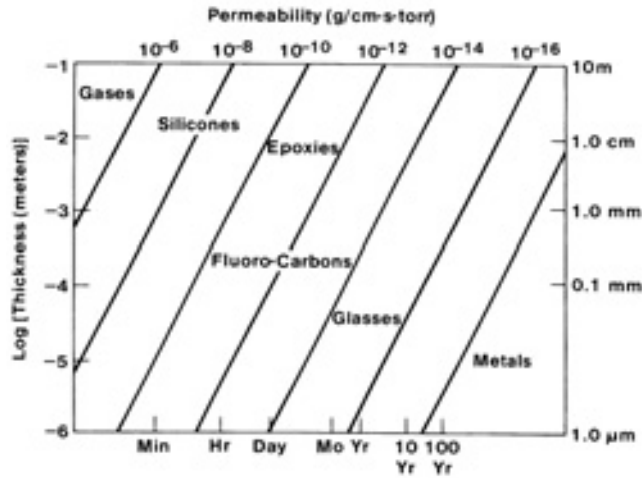


**Figure 44.** Tensile strength measurements of samples bonded with the BCB contact printing technique compared to the conventional soft-cured patterned BCB bonding technique.

bubbles coming out of the holes, since the gas trapped in the encapsulated cavity expands when heated by the solution. Small holes causing a leak rate of less than  $10^{-4}$  mbar·l·s $^{-1}$  cannot be found with the gross leak test because the vapor pressure can not overcome the surface tension of the liquid. The fine leak test consists of device exposure to helium with a pressure of 5.2 bar for 2 hours. During that dwell time, helium penetrates into the sealed cavity and the leak rate can be measured directly after the exposure with a helium leak detector with a typical sensitivity as low as  $10^{-11}$  mbar·l·s $^{-1}$  for an ALCATEL ASM 142 measurement device.

This test procedure was not found to be very suitable for encapsulated volumes below 1 mm $^3$  because the helium escapes too fast from a small cavity even for relatively small leaks. Only extremely small leaks can keep a stable helium flow for a few minutes after the exposure. Thus, the upper detection limit of the fine leak tests is lower for small cavities, and leaks of a certain size can neither be detected by the fine leak test nor by the gross leak test [255].

Therefore, relatively large samples were used for all of the tests presented in the papers 1 and 2, and the conclusions in paper 2 were drawn by comparing the measured leak rates rather than from the absolute measurement values. Furthermore, the influence of the time interval between the helium bombardment and each single measurement was investigated and considered, since the leak rate drops rapidly especially when measuring small devices. The specifications of the MIL standard only require the measurements to be taken within 60 minutes after the helium exposure. This inexact definition of the measurement step is only appropriate when the leak rate does not change rapidly over time, which is not the case for small test devices,



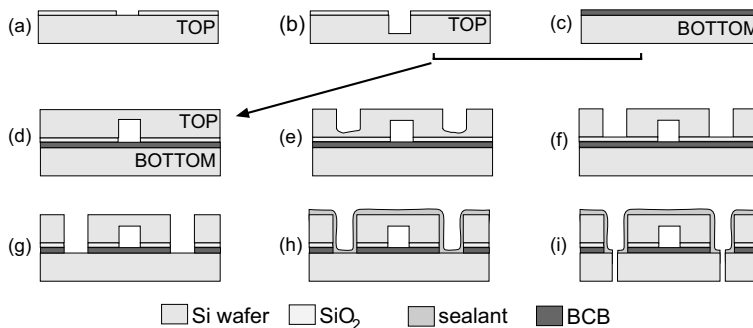
**Figure 45.** Effectiveness of sealant materials measured by the time for moisture to permeate various sealant materials in one defined geometry [190, Section 14.4.1].

where the leak rate changes drastically and is undetectable after only a few minutes.

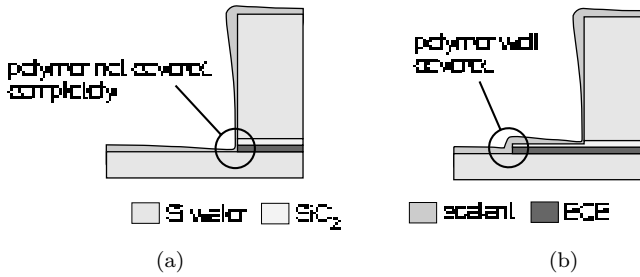
As a conclusion, helium leak tests are still very informative about the hermeticity of a MEMS package since the measurement devices are very sensitive. However, the test limits defined in the MIL standard are not applicable and the measurement results have to be interpreted carefully.

#### 4.6 Sealing by additional cladding layers

Paper 2 presents a mask-less hermetic encapsulation technique based on adhesive full-wafer bonding by cladding the adhesive with an additional diffusion barrier. Both PECVD silicon nitride and evaporated gold were investigated for that purpose. Silicon nitride is already widely used as a passivation and protective diffusion barrier in



**Figure 46.** Fabrication procedure of the adhesive bonded test structures covered with the additional diffusion barrier.

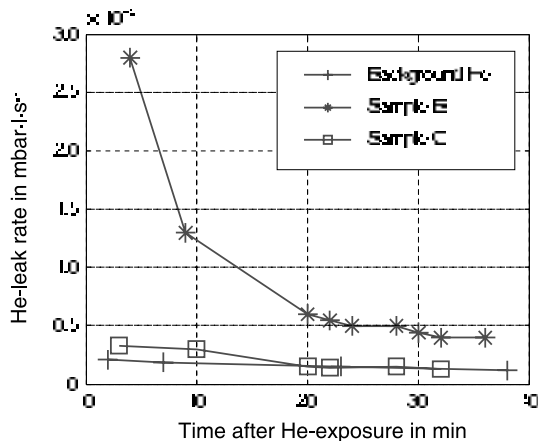


**Figure 47.** Step coverage of the polymer by the deposited diffusion barrier for (a) the fabrication technique as shown in Figure 46, and (b) the improved fabrication procedure as described in paper 2 and in the text.

IC processing to inhibit electro-oxidation (corrosion) and metal migration [293,294], which are attributed to the presence of moisture in the package [190].

Figure 46 shows the basic idea of the test device fabrication with the sealing on wafer-level: A top wafer containing the etched cavities to be encapsulated (a)–(b) is bonded to a bottom wafer with a spin-coated, 5  $\mu\text{m}$  thick layer of BCB (c)–(d). Then, grooves separating the samples are diced in the top wafer with a relatively thick saw blade (e). The grooves are etched further down to the bottom substrate by plasma-etching (f)–(g), and the devices are finally sealed by depositing the diffusion barrier (h). The devices are separated by a second dicing step with a thin saw blade (i).

This procedure is relatively simple but unfortunately did not work as intended. The deposited cladding material does not fully cover the concave corner directly underneath the 500  $\mu\text{m}$  thick top wafer as illustrated in Figure 47(a). The sealing method was improved by creating a step in the base of the device consisting of the polymer



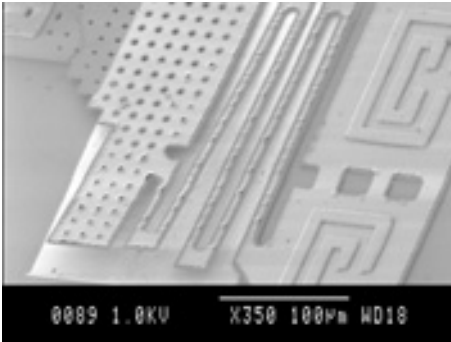
**Figure 48.** Helium leak rates over time for two samples fabricated with the straight-down etch technique (sample B) and with the improved cladding technique (sample C), compared to the background helium rate.

layer and covered by the silicon dioxide layer left from the hard mask which defined the cavity in the top wafer. This step is created by an additional plasma etch step widening the trenches by etching the sidewalls of the top wafer. The step in the basis of the device leads to full coverage of the unprotected polymer areas, as illustrated in Figure 47(b). The total fabrication procedure, slightly more complex than illustrated in Figure 46, is still mask-less since it does not involve any photolithography process, and is described in detail in paper 2.

Hermeticity tests carried out on samples using the straight-down etching technique and on samples using the modified fabrication procedure show definitively an improvement of the hermeticity of the sealed cavities, as illustrated in Figure 48.



*about some minor and major mistakes happening  
in the clean-room by doing the work for this thesis ...*



*Science must have originated in the  
feeling that something was wrong.*

*Thomas Carlyle  
1795–1881, Scottish Philosopher*

*Our greatest glory is not in never  
falling, but in rising every time we fall.*

*Confucius  
BC 551–479, Chinese Ethical Teacher*

*Sometimes the best gain is to lose.*

*George Herbert  
1593–1633, British Metaphysical Poet*

*Man needs difficulties; they are neces-  
sary for health.*

*Carl Gustav Jung  
1875–1961, Swiss Psychiatrist*

*Nothing is more intolerable than to have  
to admit to yourself your own errors.*

*Ludwig van Beethoven  
1770–1827, German Composer*

*A clever man commits no minor blun-  
ders.*

*Johann Wolfgang von Goethe  
1749–1832, German Poet*

*There is a crack in everything, that's  
how the light gets in.*

*Leonard Cohen  
1934–, Canadian Singer*

*I hear — I forget. I see — I remember.  
I do — I understand.*

*Confucius  
BC 551–479, Chinese Ethical Teacher*



*Success is not the key to happiness.  
Happiness is the key to success. If you  
love what you are doing, you will be suc-  
cessful.*

*Albert Schweitzer  
1875–1965, Medical Missionary,  
Theologian*

## 5 Summary of appended papers

### Quick summary

**Paper 1** investigates the process parameters for adhesive bonding with patterned dry-etch BCB layers as well as photosensitive BCB layers.

**Paper 2** reports on a method of improving the hermeticity of adhesive bonded cavities by using an additional passivation layer as diffusion barrier.

**Paper 3** introduces the S-shaped film actuator based switch with characterization results of the first prototypes.

**Paper 4** discusses design details and fabrication aspects of the switch.

**Paper 5** shows a method to create glass lid packages by full-wafer bonding with a patterned BCB layer, including thick in-plane interconnections and vertical through-wafer vias in glass substrates by a novel two-step etch technique.

**Paper 6** reports on a method to significantly improve the bond strength of patterned adhesive bonded wafers by combining a pre-cured patterned BCB layer with an uncured contact printed BCB film.

**Paper 7** discusses the design and initial measurement results of a 2-bit meander slot antenna intended to be reconfigurable by the film actuator based switch.

### Detailed summary

**Paper 1:** *Selective Wafer-Level Adhesive Bonding with Benzocyclobutene for Fabrication of Cavities*

This paper describes an adhesive wafer-bonding technique in which the adhesive material is structured prior to bonding. This technique can be used to create encapsulated cavities of different heights and sizes directly in the bonding layer. Benzocyclobutene (BCB) was used as the adhesive bonding material. The structuring of the BCB was done either by dry etching or by using photosensitive BCB. The process parameters needed to achieve a high bond quality while retaining the shapes of the structures in the bonding layer were investigated. Furthermore, helium leak tests were performed to investigate the suitability of selective adhesive bonding for applications requiring quasi-hermetic seals.

**Keywords:** adhesive bonding, BCB, helium leak tests, selective bonding, wafer level packaging.

**Paper 2:** *Sealing of Adhesive Bonded Devices on Wafer Level*

This paper presents a low temperature wafer-level encapsulation technique to hermetically seal adhesive bonded microsystem structures by cladding the adhesive with an additional diffusion barrier. Two wafers containing cavities for MEMS devices were bonded using benzocyclobutene (BCB). The devices were sealed by a combined dicing and self-aligning etching technique and by finally coating the structures with evaporated gold or PECVD silicon nitride. The sealing layer was inspected visually by SEM and helium leak tests were carried out. For the larger test devices, the helium leak test method was found to be suitable for comparative measurements, and the passivation of the devices with silicon nitride showed an improvement of the package hermeticity.

**Keywords:** adhesive bonding, BCB, hermetic sealing, wafer level packaging.

**Paper 3:** *Low-Voltage High-Isolation DC-to-RF MEMS Switch Based on an S-shaped Film Actuator*

This paper describes a novel electrostatically actuated microelectromechanical series switch for switching DC to RF signals. The device is based on a flexible S-shaped film moving between a top and a bottom electrode in touch-mode actuation. This concept, in contrast to most other MEMS switches, allows a design with a low actuation voltage independent of the off-state contact distance. Larger nominal switching contact areas are possible while maintaining high isolation in the off-state. Prototype switches were successfully fabricated and evaluated.

**Keywords:** electrostatic actuator, film actuator, MEMS switches, microwave switches, RF MEMS.

**Paper 4:** *Design and Fabrication Aspects of an S-shaped Film Actuator Based DC to RF MEMS Switch*

This paper reports on design and fabrication aspects of a new microelectromechanical series switch for switching DC to RF signals. The switch consists of a flexible S-shaped film with the switching contact, rolling between a top and a bottom electrode in electrostatic touch-mode actuation. This concept enables a low actuation voltage design independent of the contact distance in the off-state. The RF transmission line and the MEMS part of the switch are fabricated on separate wafers, allowing implementation of the switch with different RF substrates. The final assembly is done on device level for the first prototypes, even though the design provides the possibility of an assembly by full wafer bonding leading to a "ready-to-dice", near-hermetic packaged switch.

**Keywords:** film actuator, MEMS switches, RF MEMS, touch-mode actuation.

**Paper 5:** *Low-Cost Glass-Lid Packaging by Adhesive Full-Wafer Bonding with Two-Step Etched Electrical Feedthroughs*

This paper reports on a low-cost fabrication technique to create glass lid encapsulations for 0-level near-hermetic packaging of microsystem devices using adhesive full-wafer bonding with a previously patterned polymer layer of benzocyclobutene (BCB). Furthermore, a new two-step technique to create low-density feedthroughs in glass substrates for electrical interconnections through the wafer is introduced. The vias are fabricated by combining a mechanical etch step by powder-blasting and a subsequent short hydrofluoric acid wet etch step. Both techniques were successfully combined to fabricate glass lid encapsulations by adhesive full-wafer bonding with thick electrical interconnections penetrating through the patterned adhesive layer, and with vertical vias through the glass wafer to electrically access structures on the glass lids.

**Keywords:** adhesive bonding, glass lid encapsulation, powder blasting, through-wafer vias, wafer-level packaging.

**Paper 6:** *BCB Contact Printing for Patterned Adhesive Full-Wafer Bonded 0-Level Packages*

An adhesive bonding method with benzocyclobutene is presented which combines the advantages of patterned adhesive bonding with the advantages of bonding with an uncured polymer layer. The pre-cured pattern in the adhesive layer is "inked" with uncured polymer by pressing the substrate toward an auxiliary wafer with a thin liquid polymer layer. The substrate with the inked pattern is finally bonded to a glass wafer to create glass lid packages by full-wafer bonding. The bond interface was investigated by scanning acoustic microscopy and the failure mechanisms occurring during pull-tests were investigated. Tensile strength tests revealed a significantly improved bond strength compared to soft-cured patterned BCB bonding.

**Keywords:** adhesive bonding, BCB, bond strength, contact printing, MEMS packaging, wafer-level packaging.

**Paper 7:** *A 2-bit Reconfigurable Meander Slot Antenna with High-displacement Low-voltage RF MEMS Switches (manuscript)*

This paper presents a novel meander slot antenna which is intended to be re-configured by two RF MEMS switches of the S-shaped film actuator switch type. The active reconfiguration of the antenna allows four different operating frequencies with high frequency selectivity and good impedance matching. The upper part of the micromachined switch is intended to be placed directly on the antenna substrate by flip-chip-like mounting. The manuscript presents simulation and initial measurement results of the antenna without mounted switches and discusses the special implementation concept of the switches onto the antenna substrate.

**Keywords:** meander slot antenna, reconfigurable antenna, RF MEMS switch, switch integration.

### *about decisions ...*

*Decide promptly, but never give any reasons. Your decisions may be right, but your reasons are sure to be wrong.*

*Lord Mansfield  
1867–1915, British Artist, Author*

*Our lives are a sum total of the choices we have made.*

*Wayne Dyer  
1940–, American Psychotherapist*

*It is only in our decisions that we are important.*

*Jean-Paul Sartre  
1905–1980, French Writer*

*A weak man has doubts before a decision, a strong man has them afterwards.*

*Karl Kraus  
1874–1936, Austrian Satirist*

*We will either find a way, or make one.*

*Hannibal  
BC 247–182, Carthaginian General*

*Indecision is the seedling of fear.*

*Napoleon Hill  
1883–1970, American Speaker*

*The quality of decision is like the well-timed swoop of a falcon which enables it to strike and destroy its victim.*

*Sun Tzu  
BC 400–430, Chinese Statesman,  
in "The Art of War"*



*Satisfaction lies in the effort, not in the attainment, full effort is full victory.*

*Mahatma Gandhi  
1869–1948, Indian Political*

## 6 Conclusions

This thesis presents a novel RF MEMS switch based on an S-shaped film actuator. The switch consists of a thin silicon nitride membrane rolling between a top and a bottom electrode. This design provides the following main advantages:

- large contact distance for high signal isolation and large overlapping contact areas
- low actuation voltage independent of the contact distance
- active opening capability
- metal contacts for large bandwidth DC to RF signal switching
- electrostatic actuation for almost-zero power consumption
- minimum distortions of the RF signal with the film rolling longitudinal to the transmission line
- two-part assembly for integration with substrates and circuits incompatible with MEMS fabrication
- wafer-level assembly by individual flip-chip bonding or by full-wafer bonding, both providing a near-hermetic, ready-to-dice package

The thesis

- introduces and discusses the concept, comparing its advantages and disadvantages,
- provides information on the design and on the fabrication of the devices,
- presents the characterization results of the actuator, and
- reports on the measured and simulated RF performance of the fabricated prototypes.

The first devices of the switch were successfully fabricated and characterized, confirming the concept with most of the expected advantages, despite a few design-related problems of the prototypes. The S-shaped film actuator was found to be suitable for RF MEMS switching devices with demands on high signal isolation, low actuation voltage and almost-zero power consumption.

Furthermore, this thesis introduces novel ideas on packaging of (RF) MEMS devices by adhesive full-wafer bonding:

- The process parameters of patterned adhesive bonding were investigated, and this technique was successfully applied to create cavities for MEMS devices by full-wafer bonding.
- Hermeticity tests were carried out on such encapsulated cavities, and it was shown that the gas-tightness of the packages can be improved by additional passivation layers.
- Wafer-scale glass lid encapsulation by patterned adhesive full-wafer bonding with subsequent dicing of the glass wafer without harming structures on the substrate wafer was successfully demonstrated.
- The fabrication of electrical in-plane interconnections penetrating through the polymer sealing rings was investigated and a method to create vertical through-wafer interconnects in glass substrates, by using a novel two-step etch technique based on powder-blasting and wet-etching, was applied to the glass-lid encapsulation technique.
- The bond strength of patterned adhesive wafer bonding was substantially improved by a contact printing method, inking the pattern with a liquid polymer film, which combines the advantages of patterned adhesive bonding with the higher bond strength of uncured polymer bonding.
- The bond interfaces were qualitatively and quantitatively investigated by tensile strength measurements, visual inspection with optical microscopes, razor-blade tests, and scanning acoustic microscopy.

Patterned adhesive full-wafer bonding was found to be a simple, robust and highly parallel near-hermetic wafer-level packaging technique for MEMS devices.



*some inspiration from a very nice place in the south ...*

*I really lack the words to compliment myself today.*

Alberto Tomba  
1966–, Italian Alpine Ski Star

*Nebulat, ergo cogito.*

Umberto Eco  
1932–, Italian Semiologist, Novelist

*When the danger is past God is cheated.*

Italian Proverb of Unknown Origin

*On the piazza, the sun is shining for everyone.*

Italian Proverb of Unknown Origin

*It may be you fear more to deliver judgment upon me than I fear judgment.*

Giordano Bruno  
1548–1600, Italian Scientist

*Power tires those alone who don't have it.*

Giulio Andreotti  
1919–, Former Italian Prime Minister

*He who knows little knows enough if he knows how to hold his tongue.*

Italian Proverb of Unknown Origin

*I'm a pessimist because of intelligence, but an optimist because of will.*

Antonio Gramsci  
1891–1937, Italian Political Theorist

*Anyone who has obeyed nature by transmitting a piece of gossip experiences the explosive relief that accompanies the satisfying of a primary need.*

Primo Levi  
1919–1987, Italian Author

*He who is not impatient is not in love.*

Italian Proverb of Unknown Origin

*Adversity reveals genius, prosperity conceals it.*

Horace  
BC 65–8, Italian Poet

*See everything, overlook a great deal, correct a little.*

Pope John XXIII  
1881–1963, Italian Head of Roman Catholic Order

*One does not kill oneself for love of a woman, but because love — any love — reveals us in our nakedness, our misery, our vulnerability, our nothingness.*

Cesare Pavese  
1908–1950, Italian Novelist

*A little man often casts a long shadow.*

Italian Proverb of Unknown Origin

*Myth is nourished by silence as well as by words.*

Italo Calvino  
1923–1985, Italian Novelist



*When men speak of the future, the Gods laugh.*

*Chinese Proverb*

## 7 Outlook

The presented RF MEMS switch is a quite complex device and I am glad that I was able to fabricate and to characterize the first prototypes. However, the device evaluation has revealed some problems and a second generation design should address the following topics, as discussed in Section 3.3.2:

- improved RF design of the coplanar waveguide
- larger contact force for lower contact resistance
- surface contamination and stiction problems of the metal contacts
- miniaturization of the total device size by reducing the lateral dimensions, while maintaining the large contact distance
- batch assembly by full-wafer bonding in clean-room environment
- access of the electrodes on the top part of the switch from the bottom wafer by integrated vertical interconnections

The reliability of the switch, encapsulated in an inert atmosphere, should be fully investigated, in connection to the processing parameters of the silicon nitride film and variations of the actuator size, the contact distance and the contact area. It would also be interesting to evaluate the RF benefits of a capacitive shunt version of the switch.

It is a pleasure for me to see that the work on MEMS switches based on this concept will continue in our microsystem group in two different application fields. Capacitive shunt switches from the first design are going to be used for reconfigurable antennas in a joint project at our department, as reported in the manuscript in the appendix. The other application uses the combination of high signal isolation and electrostatic actuation of the film actuator for switching telecommunication networks.

Patterned adhesive full-wafer bonding is a valuable packaging tool which is already used in commercial applications. The ideas presented in this thesis, especially the contact-printing method to improve the bond strength, are hopefully valuable for the future development in the field. The main problem of packages fabricated by adhesive bonding is the insufficient hermeticity of the polymer bonding materials. Sealing by additional passivation layers, as presented in this thesis, might be an option for certain applications, but adhesives with improved hermeticity, probably non-organic materials such as the already widely used but difficult to pattern glass-frit, are necessary for a wider use of adhesive bonding in near-hermetic and hermetic packaging. The technology is already quite mature on a device and system level and seems just to be waiting for new impulses from the material science world. The MEMS engineer however is still challenged with design problems such as the vertical electrical interconnections between the adhesive bonded wafers, an issue which might attract increased attention of the research community in this field.

*So your work is to analyze this incomplete manuscript of life and to explore what happened to the manuscript here and there. It is not necessary for you to understand everything about your past. You should not torment yourself in trying to learn it. If you do not understand your past, you are not ignorant, you are merely innocent. You are here at a certain point in this present life. By understanding your present situation you can understand the future and the past. The manuscript of your life has been written by you.*

*Sri Swami Rama  
1925–1996, Indian Sage  
in "Path of Fire and Light: Vol. 2"*

*No duty is more urgent than that of returning thanks.*

*Saint Ambrose  
339–397, Bishop of Milan*

## Acknowledgments

Finally, the acknowledgments. For those of you who thought this to be the first page to read: you are right in the sense that these are definitively the most important pages of the book!

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This work would not have been possible without the assistance and the help of many people I would like to express my sincere gratitude to.

First of all, I would like to thank my supervisor Prof. Göran Stemme for his guidance, for being so patient with the demanding and sometimes a bit egoistic temper from the south, for giving me much freedom in many decisions but bringing me back to the right track whenever I got stuck in less relevant questions.

I am also very grateful to my present colleagues and friends in the Microsystem Technology group: to Kjell for his skillful mechanics for measurement setups and other more or less scientific projects, to Ulla for helping with so many little problems which seem to be unsolvable for a foreigner, to Frank for introducing me to the clean-room and for funny observations and comments on new countries and their people, to Sjoerd for his uncountable hints and tips in microsystem fabrication and evaluation, for his encouraging enthusiasm and for the funny and friendly atmosphere following him step by step, to Björn for sharing so many thoughts, to Thomas for being so patient with all my Macintosh related computer problems (and there are only Macintosh related computer problems, per definition), to Jessica for helping me with all my English language-related doubts despite some of my often politically incorrect comments, to Wouter for his inspiring mind and his positive attitude, to Stefan for showing interest in continuing my work, to Hans for sharing all his knowledge and experience and for taking care of so many official documents and meetings in the SUMMIT project, to Niclas for always having an open ear for technical questions, to Calle for never letting me forget the coffee break and the Wednesday-”bulle”, and, last but not least, to Aman, my Eritrean office-neighbor, for all the joyful moments and for making me lose so many senseless sport bets on Italian football.

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explode when you press the wrong button ... without your help, Cecilia, the switch had never been possible!

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It is not easy to live abroad for such a long time and I am very glad to be able to keep or to re-discover contact with so many people from former places I was living in or I was visiting: to Alex P. with Monika in Graz, Austria, Alex R. in Vienna, Austria, Hubert in Verona, Italy, Martin in Stuttgart, Germany, Reno with Caroline in Zurich, Switzerland, Erich in Sand, Italy, Philipp in Bozen, Italy, to Bernd, Washington, DC. To Ralf in Graz, Austria, for our unbelievable deep and unrestricted e-mail communication, to Thomas and Egmont especially for your administrative help in Graz, to Christa in Winterthur, Switzerland, for intensive conversations and for distracting me so nicely from my work on this book, and, finally, to Gerold in Sand, Italy, for our entertaining Thursday-evening almost-wireless two-dimensional distance Blitz-chess tournaments.

To my mother, my father, my sister, my brother and our lazy old cat: thank you very very much for being so patient during my long stays abroad, for your support and for the good thoughts you have been sending me throughout all these years I have been living so far away from home.

Stockholm, in the summer 2004, Joachim Oberhammer



## A Small Dictionary of Useful Research Phrases

Research Phrase	Translation
It has long been known...	I didn't look up the original reference.
A definite trend is evident...	These data are practically meaningless.
Of great theoretical and practical importance...	Interesting to me.
While it has not been possible to provide definite answers to these questions...	An unsuccessful experiment, but I still hope to get it published.
Three of the samples were chosen for detailed study...	The results of the others didn't make any sense.
Typical results are shown...	The best results are shown.
These results will be shown in a subsequent report...	I might get around to this sometime if I'm pushed.
The most reliable results are those obtained by Jones...	He was my graduate assistant.
It is believed that...	I think.
It is generally believed that...	A couple of other people think so too.
It is clear that much additional work will be required before a complete understanding of the phenomenon occurs...	I don't understand it.
Correct within an order of magnitude...	Wrong.
It is hoped that this study will stimulate further investigation in this field...	This is a lousy paper, but so are all the others on this miserable topic.
Thanks are due to Joe Blotz for assistance with the experiment and to George Frink for valuable discussions...	Blotz did the work and Frink explained to me what it meant.
A careful analysis of obtainable data...	Three pages of notes were obliterated when I knocked over a glass of beer.

*... and now please read this thesis again!*

*A civilization is built on what is required  
of men, not on that which is provided  
for them.*

*Antoine de Saint-Exupéry  
1900–1944, French Aviator, Writer*

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*Then the Lord said: If now, while they are one people, all speaking the same language, they have started to do this, nothing will later stop them from doing whatever they propose to do.*

*Genesis, XI, v.6,  
New American Bible*

## Nomenclature

AC	<u>a</u> lternating <u>c</u> urrent: electrical signal without any constant, i.e. DC component
ADC	<u>a</u> nalog to <u>d</u> igital converter: converts an analog input signal related to reference voltages and quantization steps to a binary coded (digital) signal
AICC	<u>a</u> utonomous <u>i</u> ntelligent <u>c</u> ruise <u>c</u> ontrol: autopilot-like system for cars, based on automotive radar and a cruise control management unit
ATE	<u>a</u> utomated <u>t</u> est <u>e</u> quipment: full automated test equipment for the end-control and tests of products; requires signal switching with very wide bandwidth (DC to RF) and very high signal quality
AuCuCd	gold with an alloying addition of copper and cadmium, used as a hard contact material in relays
AuNi5	gold with an alloying addition of 5% nickel, used as a hard contact material in relays
BCB	divinylsiloxane bis- <u>b</u> enzoc <u>y</u> clo <u>b</u> utene, an epoxy-based polymer suitable for adhesive wafer bonding
BiCMOS	<u>b</u> ipolar and $\rightarrow$ CMOS devices integrated on the same chip
C2W	0-level packaging by <u>c</u> ap to <u>w</u> afer bonding/encapsulation
CMOS	<u>c</u> omplementary <u>m</u> etal <u>o</u> xide <u>s</u> ilicon: fabrication technology enabling NMOS and PMOS $\rightarrow$ FETs in push-pull configuration on the same chip, for linear as well as digital $\rightarrow$ ICs
CMP	<u>c</u> hemical and <u>m</u> echanical <u>p</u> olishing: chemically assisted lapping of a wafer to reduce its thickness
CPW	<u>c</u> oplanar <u>w</u> ave <u>g</u> uide: microwave transmission line with ground-signal-ground lines in one plane; simple to fabricate and good mode-control over a large frequency range. Variations: finite ground coplanar waveguide (FGCPW) with limited lateral size of the ground lines; grounded CPW with a ground layer underneath the CPW.

CSP	<u>chip scale package</u> : package whose total area is not larger than 120% of the chip size
CTE	<u>coefficient of thermal expansion</u> : material property describing the relative elongation of a material with increasing temperature
CVD	<u>chemical vapor deposition</u> : deposition of silicon-containing layers in surface micromachining, based on convection, diffusion and absorption of reactants on the target surface followed by chemical reactions, desorption and transport of byproducts by diffusion and forced convection in the deposition tool. The most common deposition processes are APCVD (atmospheric pressure CVD), $\rightarrow$ LPCVD and $\rightarrow$ PECVD.
CW	<u>continuous wave</u>
DC	<u>direct current</u> : electrical signal with constant, i.e. time invariant level
DIP	<u>dual-in-line package</u> : industry standard plastic or ceramic package for through-hole mounting
DLP	<u>digital light processing device</u> : a trademark of Texas Instruments for MEMS digital micromirror arrays used in projectors
DRIE	<u>deep reactive ion etching</u> : $\rightarrow$ ICP
EMI	<u>electro-magnetic influence</u> : externally or internally created electrical disturbances within or between devices, due to capacitive, inductive, ohmic or radiation coupling of electrical signals
EMR	<u>electro-mechanical relay</u> : conventional macro-world relay assembled with traditional mechanical engineering fabrication techniques
FBAR	(thin) <u>film bulk acoustic resonators</u> : passive electrical components similar to $\rightarrow$ SAW, based on propagating transversal bulk acoustic waves (BAW) in a thin freestanding film
FET	<u>field effect transistor</u> : semiconductor device operating as a voltage controllable resistor in a drain-gate-source configuration
FGCPW	$\rightarrow$ CPW
GaAs	gallium arsenide: III-V semiconductor material with very good RF properties compared to high resistivity silicon
HF	hydrofluoric acid
HRSS	<u>high resistivity silicon substrate</u> : in contrast to standard silicon substrate for MEMS fabrication (1–10 $\Omega$ ·cm), substrate with a resistivity of over 1000 $\Omega$ ·cm
IC	<u>integrated circuit</u> : highly integrated semiconductor device, nowadays consisting of up to a few hundred million individual components fabricated on one single chip

ICP	<u>i</u> nductive <u>c</u> oupled <u>p</u> lasma process to etch deep structures with good control over the steepness of the side-walls
IP3	<u>i</u> ntercept <u>p</u> oint of <u>3</u> rd order: characterization of the nonlinear distortions of an electronic device by a two-tone intermodulation test. The IP3 is the fictive intersection of the linear portion of the output signal with the portion of the 3rd order intermodulation product
KOH	potassium (kalium) hydroxide, a crystal-orientation dependent etchant of silicon
LCP	<u>l</u> iquid <u>c</u> rystal <u>p</u> olymer: thermoplastic polymer made of aligned molecule chains with crystal-like spatial regularity, exhibiting unique electrical, physical and chemical properties
LPCVD	<u>l</u> ow <u>p</u> ressure <u>c</u> hemical <u>v</u> apor <u>d</u> eposition: a chemical deposition process ( $\rightarrow$ CVD) in surface micromachining to deposit high-density silicon-containing layers at temperatures above 600 °C
LTCC	<u>l</u> ow temperature <u>c</u> o-fired <u>c</u> eramic: material for multichip module (MCM) substrates and for ceramic packages with very good mechanical and electrical properties
LTO	<u>l</u> ow <u>t</u> emperature <u>o</u> xide, deposited silicon dioxide with a combination of phosphorus and boron providing excellent low-temperature reflow properties
MEMS	<u>m</u> icro <u>e</u> lectro <u>m</u> echanical <u>s</u> ystems: traditionally, electromechanical devices fabricated in standard IC manufacturing facilities. MEMS is nowadays also used for a broader group of $\rightarrow$ microsystem devices.
MIMO	<u>m</u> ultiple <u>i</u> nput <u>m</u> ultiple <u>o</u> utput
MST	<u>m</u> icro <u>s</u> ystem <u>t</u> echnology: a synonym for MEMS technology or for micromachining, especially often used in Europe
NiFe	nickel ferrum (typical permalloy composition of 81% Fe and 19% Ni): most common ferromagnetic material used in MEMS, providing the combination of relatively high saturation flux density, low hysteresis losses and near zero magnetostriction
OWR	<u>o</u> bstacle <u>w</u> arning <u>r</u> adar: automotive radar for driving and parking aid (front and side collision control)
PCB	<u>p</u> rinted <u>c</u> ircuit <u>b</u> oard: multi-layer laminate with embedded metal lines and vertical vias used in electronics industries to mechanically mount and interconnect $\rightarrow$ ICs and passive components
PDMS	<u>p</u> oly- <u>d</u> i <u>m</u> ethyl <u>s</u> iloxane: biocompatible polymer material suitable for micromolding plastic replication with very simple processing since it is casted over the master structures and fully cured at 65 °C

PECVD	<u>p</u> lasma <u>e</u> nhanced <u>c</u> hemical <u>v</u> apor <u>d</u> eposition: a chemical deposition process ( $\rightarrow$ CVD) in surface micromachining to deposit silicon-containing layers at low temperatures
PIN diode	semiconductor device operating as a variable resistor at RF and microwave frequencies, controlled by a bias current. The name stems from an intrinsic layer embedded between the p and the n-doped layers of a semiconductor diode.
PMMA	<u>p</u> oly <u>m</u> ethyl <u>m</u> eth <u>a</u> crylate: thermoplastic material, usually structured by conventional or modified plastic replication techniques such as injection molding or hot embossing.
PZT	lead ( <u>P</u> b) <u>z</u> irconate <u>t</u> itanate: along with lead titanate (PT) and lead metaniobate ( $\text{PbNb}_2\text{O}_6$ ), one of today's most used ceramic piezoelectric materials
RF	<u>r</u> adio <u>f</u> requency
RF MEMS	<u>r</u> adio <u>f</u> requency <u>m</u> icro <u>e</u> lectro <u>m</u> echanical systems
RIE	<u>r</u> eactive <u>i</u> on <u>e</u> tching: plasma process to etch thin films
RT	<u>r</u> oom <u>t</u> emperature
SAM (1)	<u>s</u> canning <u>a</u> coustic <u>m</u> icroscope: inspection tool detecting the change of acoustic impedance at material interfaces; especially to investigate failure mechanisms such as dilaminations and voids
SAM (2)	<u>s</u> elf <u>a</u> ssembled <u>m</u> onolayer: in RF MEMS, mainly used for anti-stiction coating
SAW	<u>s</u> urface <u>a</u> coustic <u>w</u> ave filters: passive electrical component for signal processing (filtering, delaying) based on the interference/wave propagation of transversal or shear acoustic waves on the surface of a piezoelectric crystal substrate
SEM	<u>s</u> canning <u>e</u> lectron <u>m</u> icroscope: high resolution microscope based on an electron beam scanning the sample surface; in MEMS used to examine micromachined structures
SiN	silicon nitride
SiO <sub>2</sub>	silicon dioxide
SMD	<u>s</u> urface <u>m</u> ounted <u>d</u> evices: technology for surface mounting of devices in standard packages on printed circuit boards
SMR	<u>s</u> olidly <u>m</u> ounted <u>r</u> esonator: thin film transversal-mode bulk acoustic wave (BAW) resonator similar to $\rightarrow$ FBAR, does not need a freestanding film, since it is based on reflecting films acoustically isolating from the substrate

SOC	<u>s</u> ystem <u>o</u> n a <u>c</u> hip: microsystem components of different types (e.g. MEMS and CMOS circuits) are integrated on a chip, requires compatibility in the fabrication processes and a comparable yield, but leads to a more compact system and simpler first-level packaging than the $\rightarrow$ SiP approach.
SOI	<u>s</u> ilicon <u>o</u> n <u>i</u> nsulator: wafer consisting of a thick bulk silicon, a thin silicon dioxide isolation layer and a silicon device layer; used for RF CMOS and for MEMS fabrication
SOP (1)	<u>s</u> ystem <u>o</u> n a <u>p</u> ackage: microsystem components of different types (RF circuits, MEMS, CMOS logic) are fabricated on different wafers, diced and the chips integrated in a single package. This approach, in contrast to $\rightarrow$ SOC, is not restricted in terms of process compatibility and allows the optimum available technology for each sub-system, but involves interconnection problems and leads to a larger, more complicated package. The term SiP (system in package) is used synonymously and is more correct than the circumscription SOP, but the latter is more commonly used.
SOP (2)	<u>s</u> mall <u>o</u> utline <u>p</u> ackage: industry standard package for surface mounted devices (SMD) on printed circuit boards ( $\rightarrow$ PCB)
SPDT	<u>s</u> ingle <u>p</u> ole <u>d</u> ouble <u>t</u> hrough: relay type characterized by its signal switching path: a single input is switched to one of the two outputs
SPST	<u>s</u> ingle <u>p</u> ole <u>s</u> ingle <u>t</u> hrough: relay type characterized by its signal switching path: a single input is switched to a single output
SRF	<u>s</u> elf <u>r</u> esonance <u>f</u> requency, measured to characterize the high frequency behavior of passive components
SSR	<u>s</u> olid <u>s</u> tate <u>r</u> elay: electronic relay for high current and high isolation voltage applications, based on $\rightarrow$ FET or bipolar transistors
TMAH	<u>t</u> etramethyl <u>a</u> mmonium <u>h</u> ydroxide, a crystal-orientation dependent etchant of silicon typically allowing for smoother etch walls than $\rightarrow$ KOH
VCO	<u>v</u> oltage <u>c</u> ontrolled <u>o</u> scillator
W2W	0-level packaging by <u>w</u> afer to <u>w</u> afer bonding/encapsulation
<i>S-parameters</i> :	n-by-n matrix whose elements fully characterize a linear electrical network with n ports
<i>0-level packaging</i> :	synonym to $\rightarrow$ wafer-level packaging
<i>all-metal switch</i> :	electrostatic switch design without isolation layers between the electrodes; eliminates dielectric charge trapping causing failure due to electrostatic stiction; distance keeping bumps are used instead of a full covering isolation layer

- cold-switching*: term used in the lifetime (switching cycles) determination of MEMS metal-contact switches: the switching operation is carried out without electrical load; opposite: →hot-switching
- dry etching*: etching process based on a plasma with reactive chemical and/or ionic components, in contrast to chemical solutions used in wet etching.
- gyro*: rotational sensor based on an angular vibrating structure converting an external tilting force perpendicular to the rotation axis to a precession moment in the plane perpendicular to both the rotation plane and to the tilting force plane
- hot-switching*: term used in the lifetime (switching cycles) determination of MEMS metal-contact switches: the switching operation is carried out with applied electrical load; opposite: →cold-switching
- in-line switch*: cantilever based series switch with the signal line leading over the cantilever and also used as an electrode for the switch actuation; the electrical potentials of the actuation mechanism and the signal path are not isolated and the signal line is only interrupted by one gap in the off-state
- kapton*: flexible polyimide material with especially good performance in applications for very high (400 °C) and very low (-269 °C) temperature extremes. Kapton has been used for about 35 years in a wide variety of applications such as substrates for flexible printed circuits, transformer and capacitor insulation and bar code labels
- microsystem*: small scale, integrated system consisting of different components with data processing and interface functions, such as microelectronic circuits, electrical interfaces, transducers, MEMS actuators
- plasma etching*: →dry etching
- polysilicon*: silicon made up of regions, so-called grains, of crystalline material, with each grain being oriented differently from the neighboring grain. The interface between the regions is called a grain boundary. The main deposition process is →CVD
- pyrex*: glass type used as microsystem substrate with special etching properties and matched →CTE to silicon
- quasi-static modeling*: time-dependent description of a system by discrete time steps with the system being stable at each single step calculated by finding an equilibrium of the different kinds of potential energy in the system, not taking into account the dynamic behavior caused by the inertia of the involved energy stores
- step coverage*: many thin-film deposition techniques result in a very uniform layer on flat, horizontal surfaces. However, slopes or vertical walls are not

covered very well. The ratio between the thickness of the sidewall deposited layer to the thickness of the layer on the horizontal surface is called the step coverage factor.

*thermoplastic polymer*: polymer material which can be remelted many times when heated up above the melting temperature and returning to the rigid state, when cooled down below the melting temperature

*thermoset polymer*: polymer material which does not remelt, once fully cured

*wafer-level packaging*: packaging of the devices on the wafer before separation by dicing e.g.

*wafer-scale packaging*: packaging on wafer-level by full-wafer processes such as wafer bonding





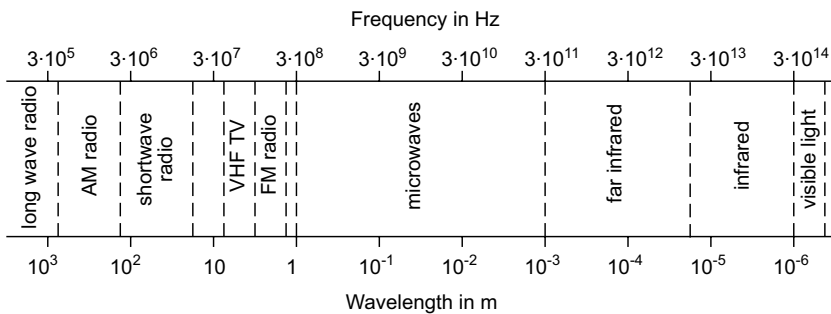
## Appendix: Standardized frequency bands

**Table 16.** Standardized frequency bands [295].

<i>From 30 Hz up to 300 GHz</i>			
Band designation	abbr.	Frequency	Wavelength
extremely low-frequency	ELF	30 Hz...300 Hz	10 Mm...1 Mm
voice-frequency	VF	300 Hz...3 kHz	1 Mm...100 km
very low frequency	VLF	3 kHz...30 kHz	100 km...10 km
low-frequency	LF	30 kHz...300 kHz	10 km...1 km
medium-frequency	MF	300 kHz...3 MHz	1 km...100 m
high-frequency	HF	3 MHz...30 MHz	100 m...10 m
very high-frequency	VHF	30 MHz...300 MHz	10 m...1 m
ultra high-frequency	UHF	300 MHz...3 GHz	1 m...10 cm
super high-frequency	SHF	3 GHz...30 GHz	10 cm...1 cm
extremely high-frequency	EHF	30 GHz...300 GHz	1 cm...1 mm

<i>At 1 GHz and above</i>		
Band designation	Frequency	Wavelength
L band	1 GHz...2 GHz	30 cm...15 cm
S band	2 GHz...4 GHz	15 cm...7.5 cm
C band	4 GHz...8 GHz	7.5 cm...3.75 cm
X band	8 GHz...12 GHz	3.75 cm...2.5 cm
Ku band	12 GHz...18 GHz	2.5 cm...1.67 cm
K band	18 GHz...26.5 GHz	1.67 cm...1.13 cm
Ka band	26.5 GHz...40 GHz	1.13 cm...7.5 mm
Q band	32 GHz...50 GHz	9.38 mm...6 mm
U band	40 GHz...60 GHz	7.5 mm...5 mm
V band	50 GHz...75 GHz	6 mm...4 mm
W band	75 GHz...100 GHz	4 mm...3.33 mm



**Figure 49.** Typical frequency bands.

